

# DICAS DE COMPONENTES DO PROFESSOR BAIRROS



08/05/2024



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## INTRODUÇÃO

Esse pdf mostra uma série de dicas que vão facilitar a escolha dos componentes para o seu projeto, identificação de componentes, e seleção do componente.

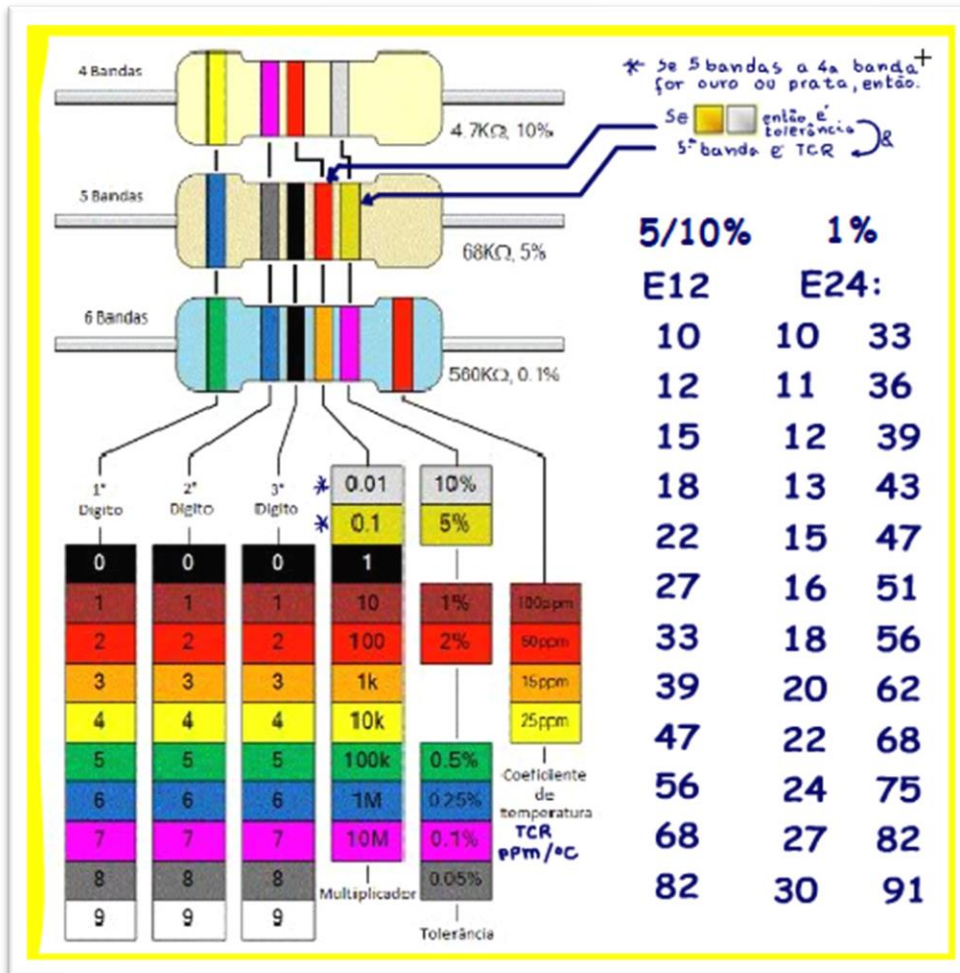
Em cada página você encontra uma lista rápida, o diagrama e pinagem quando necessário e uma descrição dos principais parâmetros e tabelas com os valores máximos.

Bom proveito.

DICAS DE COMPONENTES do Professor Bairros.

27/04/2024

CÓDIGO DE CORES DAS RESISTÊNCIAS VALORES E12/E24



Esse é o código de cores das resistências ou resistores.

E12 é a série de valores mais comum no mercado.

E24 é a série mais comum do mercado com tolerância de 1%.

Se na resistência de 5 bandas a 4ª banda for outro ou prata ela significa tolerância e a 5ª banda mais afastada significa o TCR.

Resistores com 6 bandas são geralmente para resistores de alta precisão que possuem uma banda adicional para especificar o coeficiente de temperatura (ppm/°C = ppm/K). A cor mais comum para a sexta banda é marrom (100 ppm/°C). Isso significa que para uma mudança de temperatura de 10 °C, o valor da resistência pode mudar 100ppm/10°C = 0,1%. Para o exemplo de resistor de 6 bandas mostrado acima: laranja (3), vermelho (2), marrom (1), marrom (x10), verde (1%), vermelho (50 ppm/°C) representa um resistor de 3,21 kΩ com um 1% de tolerância e um coeficiente de temperatura de 50 ppm/°C.

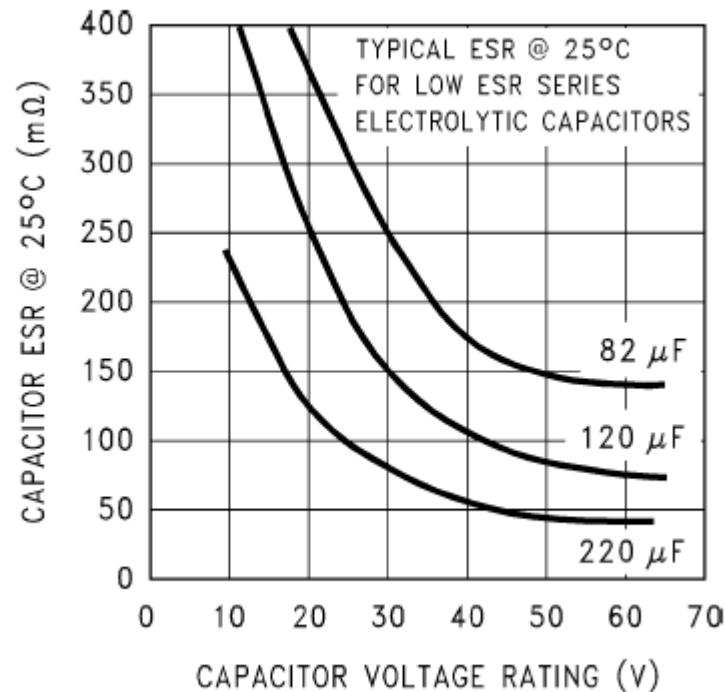
CAPACITOR

## TENSÃO E VALORES NOMINAIS DOS CAPACITORES/ELETROLÍTICOS.

ELETROLÍTICO (PTH)	POLIÉSTER	CERÂMICO
1uF (1mF)	1uF	22pF
2.2uF (2.2mF)	1nF (1K)	24pF
4.7uF (4.7mF)	10nF (10K)	27pF
10uF (10mF)	33nF (33K)	33pF
22uF (22mF)	100nF (100K)	47pF
33uF (33mF)	150nF (150K)	100pF
47uF (47mF)	4,7nF (4K7)	270pF
100uF (100mF)	47nF (47K)	330pF
120uF (120mF)	220nF (220K)	1.500pF
150uF (150mF)	330nF (330K)	1nF (1K)
220uF (220mF)	470nF (470K)	3.3nF (3K3)
330uF (330mF)		47nF (47K)
470uF (470mF)		68nF (68K)
680uf (680mF)		100nF (100K)
1.000uF (1.000mF)		
4.700uF (4.700mF)		
2.200uF (2.200mF)		

## ESR TÍPICO PARA CAPACITOR ELETROLÍTICO

A tabela abaixo mostra os valores típicos do ESR, esse é um parâmetro muito importante para os capacitores usados nas fontes chaveadas.



Uma regra de outro é usar capacitores de tensões de no mínimo 100VDC, quanto mais alta a tensão menor o ESR para a mesma capacitância.

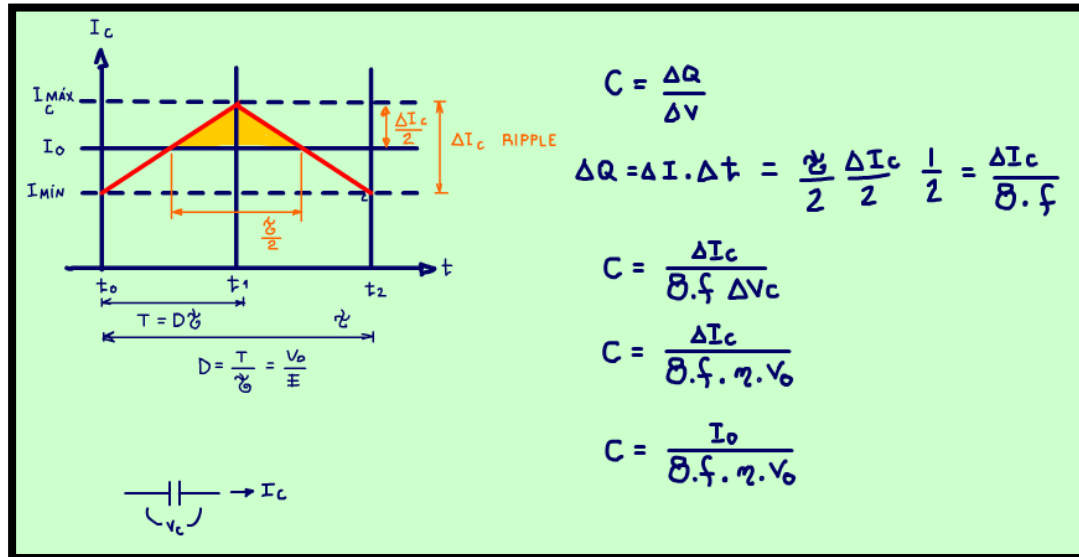
Outra dica é usar capacitores de alto valor de capacitância, quanto maior o valor da capacitância menor o capacitor.

Você também pode diminuir o ESR colocando capacitores em paralelo, é como colocar resistências em paralelo.

Essas dicas foram tiradas do manual do LM2596



COMO CALCULAR O CAPACITOR DE FILTRO.



Essa é uma das melhores equações porque leva em conta a carga que deve ser reposta pelo capacitor a cada semiciclo.

O "f" é de frequência da onda retificada, se for um retificador de meia onda deverá ser uma vez a frequência da rede, na frequência da rede de 60 Hz será 60, para onda completa será duas vezes a frequência da rede, para 60 Hz será 120.

O "Ic" é a corrente máxima na carga, "vo" é a tensão contínua na carga, aquele "n" é porcentagem do ripple desejada.

A unidade é o Farad.

Veja o exemplo da figura para uma corrente de 1A, veja que para 2A o capacitor dobra valor, a tensão de saída é 12Vcc e o ripple desejado é de 10 %, valor típico.

$$C = \frac{I_o}{8 \cdot f \cdot \eta \cdot V_o}$$

$V_o = 12V$   
 $I_o = 1A$   
 $\eta = 0,1 (10\%)$   
 $f = 120 (Onda Completa)$

$$C = \frac{1 A}{8 \cdot 120 \cdot 0,1 \cdot 12} = 868 \mu F$$

$$C = \frac{2 A}{8 \cdot 120 \cdot 0,1 \cdot 12} = 1736 \mu F$$

de

## FORMA SIMPLIFICADA DE CALCULAR O CAPACITOR DE FILTRO.

Existe uma forma rápida, lépida e rasteira de calcular o capacitor de filtro, basta fazer o capacitor proporcional a corrente máxima da fonte:

Para cada 1A use um capacitor de capacitância igual ou maior do que 1000 uF para retificador de onda completa.

No exemplo anterior para a corrente de 1A você colocaria 1000 uF ou mais, para 2 A você colocaria 2000 uF ou mais.

Para cada 2A use um capacitor de capacitância igual ou maior do que 2000 uF para retificador de meia completa.

No exemplo anterior para a corrente de 1A você colocaria 2000uF ou mais, para 2 A você colocaria 4000 uF ou mais.

Você também pode pensar em termos de miliamperes, então para retificador de onda completa para cada 1000 mA use um capacitor de 1000 uF ou mais, para onda completa.

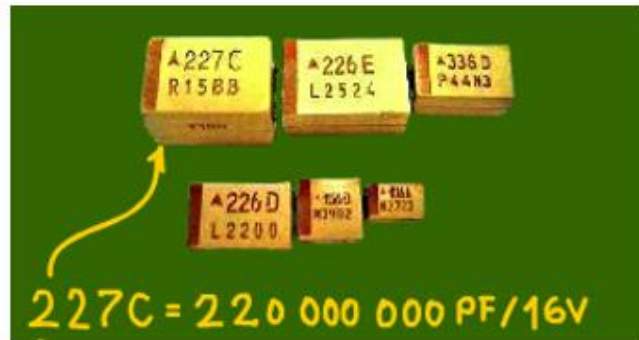
Exemplo, para uma corrente de 400 mA, use um capacitor de 400 uF ou mais, pode ser um de 470 uF.

Para meia onda dobre o valor do capacitor, simples assim.



## COMO LER CAPACITOR DE TÂNTALO SMD

Sopa de letrinhas do capacitor SMD de tântalo #shorts



Letra	Tensão
e	2.5
G	4
J	6.3
A	10
C	16
D	20
E	25
V	35
H	50

Hoje vou mostrar a sopa de letrinhas dos capacitores smd de tântalo.

O primeiro detalhe é que aquele traço branco não é o catodo, é o positivo do capacitor.

O valor do capacitor é descrito na forma de três dígitos, o primeiro e o segundo dígito descrevem o valor numérico e o terceiro dígito é o número de zeros, ou multiplicador, e a unidade é o pico Farad, não esqueça.

A tensão de trabalho é dada por uma letra conforme a tabela da figura.

27C significa 220uF tensão de 16V, 226E, 22uF tensão de trabalho de 25V.



Letra Indicativa da tolerância	tolerância
B	+/- 0,1 %
C	+/- 0,25 %
D	+/- 0,5 %
E	+/- 0,5 %
F	+/- 1%
G	+/- 2%
H	+/- 3%
J	+/- 5%
K	+/- 10%
M	+/- 20%
N	+/- 0,05%
Z	-20% a 80%

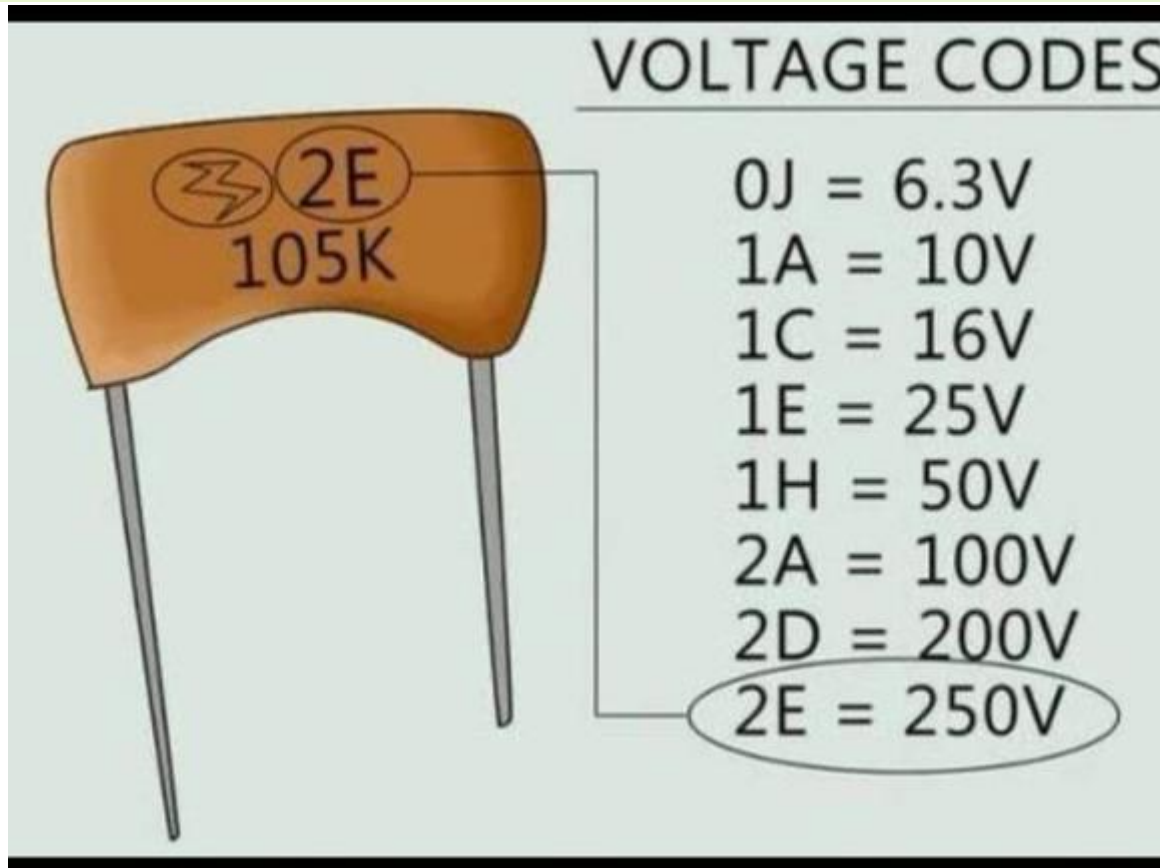
As vezes a capacitância é dada por outra sopa de letrinhas, está na tabela também.

No capacitor da figura a capacitância é igual a 100uF (A8) e o jota tensão de 6,3V.

Cuidado para não se engasgar.

Essa última tabela mostra tolerância e vale para qualquer tipo de capacitor.

## TENSÃO DOS CAPACITORES DE CERÂMICA

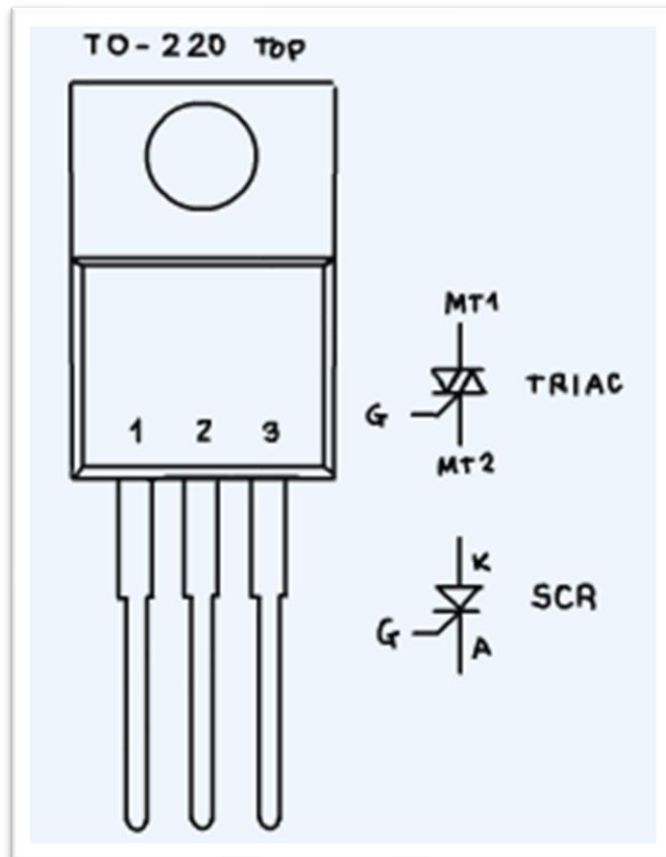


Se não disser nada 50V.

TIRISTORES/SCR/TRIACS/DIACS/MOC

## TRIAC TIC206, TIC216

TIC 206/TIC216 /MAC16

**TIC 206: 4A**

TRIAC para corrente de 4A tensão de trabalho: A=100V, B=200V, D=400V, M=600V, S=700V, N=800V. Pinos: 1=MT1, 2= MT2, 3=G.

**TIC 216: 8A**

TRIAC para corrente de 8A tensão de trabalho: D=400V, M=600V, S=700V, N=800V.

Pinos: 1=MT1, 2= MT2, 3=G.

**MAC16:**

TRIAC para corrente de 16A tensão de trabalho: D=400V, M=600V, N=800V,

Pinos: 1=MT1, 2= MT2, 3=G.



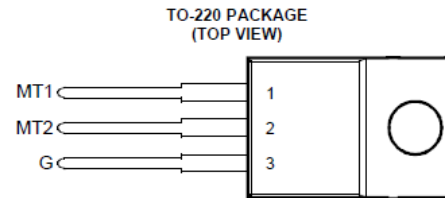
TRIAC TIC246 HIGH CURRENT 16A

TIC246 SERIES  
SILICON TRIACS

Copyright © 2000, Power Innovations Limited, UK

DECEMBER 1971 - REVISED JUNE 2000

- High Current Triacs
- 16 A RMS
- Glass Passivated Wafer
- 400 V to 800 V Off-State Voltage
- 125 A Peak Current
- Max  $I_{GT}$  of 50 mA (Quadrants 1 - 3)



Pin 2 is in electrical contact with the mounting base.

MDC2ACA

absolute maximum ratings over operating case temperature (unless otherwise noted)

RATING		SYMBOL	VALUE	UNIT
Repetitive peak off-state voltage (see Note 1)	TIC246D	$V_{DRM}$	400	V
	TIC246M		600	
	TIC246S		700	
	TIC246N		800	
Full-cycle RMS on-state current at (or below) 70°C case temperature (see Note 2)		$I_{T(RMS)}$	16	A
Peak on-state surge current full-sine-wave at (or below) 25°C case temperature (see Note 3)		$I_{TSM}$	125	A
Peak gate current		$I_{GM}$	±1	A
Operating case temperature range		$T_C$	-40 to +110	°C
Storage temperature range		$T_{stg}$	-40 to +125	°C
Lead temperature 1.6 mm from case for 10 seconds		$T_L$	230	°C

TRIAC BTA40 40A



**BTA40, BTA41 and BTB41 Series**

STANDARD

40A TRIACs

**Table 1: Main Features**

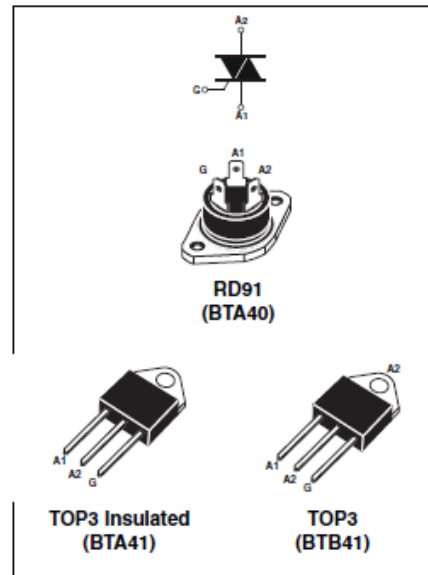
Symbol	Value	Unit
$I_{T(RMS)}$	40	A
$V_{DRM}/V_{RRM}$	600 and 800	V
$I_{GT}(Q_i)$	50	mA

**DESCRIPTION**

Available in high power packages, the **BTA/BTB40-41** series is suitable for general purpose AC switching. They can be used as an ON/OFF function in applications such as static relays, heating regulation, induction motor starting circuits... or for phase control operation in light dimmers, motor speed controllers, ...

Thanks to their clip assembly technique, they provide a superior performance in surge current handling capabilities.

By using an internal ceramic pad, the BTA series provides voltage insulated tab (rated at 2500V<sub>RMS</sub>) complying with UL standards (File ref.: E81734).

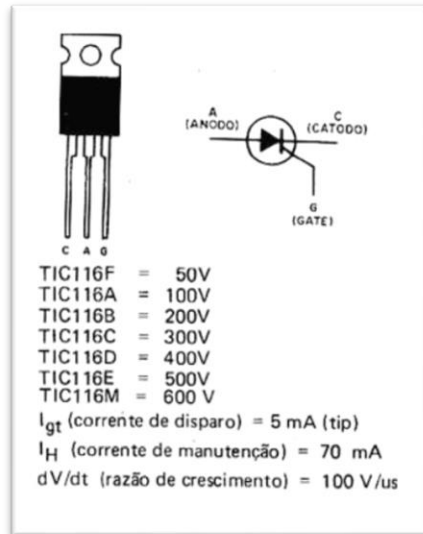


**Table 2: Order Codes**

Part Number	Marking
BTA40-xxxB	
BTA41-xxxBRG	See table 8 on page 6
BTB41-xxxBRG	

## SCR TIC 106 E FAMÍLIA

## TIC106/MCR100



**TIC 106:**SCR para corrente de 5A tensão de trabalho: D=400V, M=600V, S=700V, N=800V. Pinos: 1=K, 2=A, 3=G.

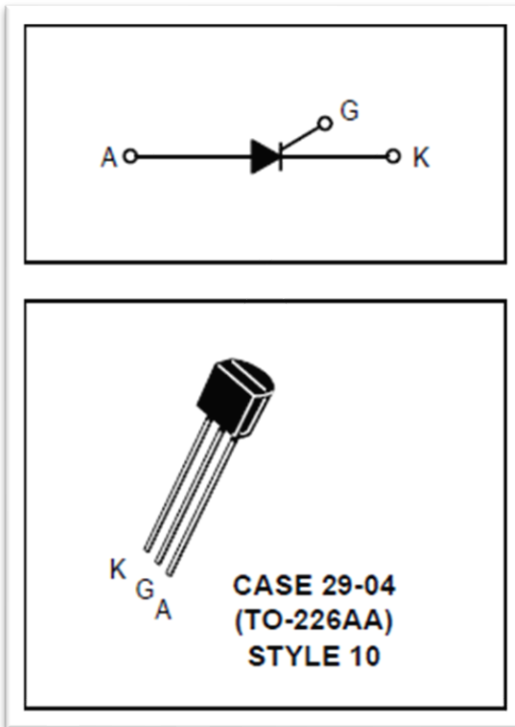
SCR corrente de 0,8A, corrente de Gate bem baixa 200uA e baixa tensão de retenção (Holding current) 5mA.

TIC 116: 6A

para corrente de 8A tensão de trabalho: D=400V, M=600V, S=700V, N=800V. Pinos: 1=K, 2= A, 3=G.

**R100:** MCR100-3 para 100V,

## MCR100 SCR PEQUENO MCR100 E FAMÍLIA



MCR100-4 para 200V, MCR100-5 para 400V, MCR100-6 para 600V.

Sensitive Gate Trigger Current – 200 mA Maximum

Low Reverse and Forward Blocking Current – 100 mA Maximum, TC = 125°C

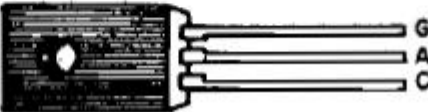
Low Holding Current – 5 mA Maximum

Glass-Passivated Surface for Reliability and Uniformity

## MCR406 SCR PARA 4A

**MCR406**

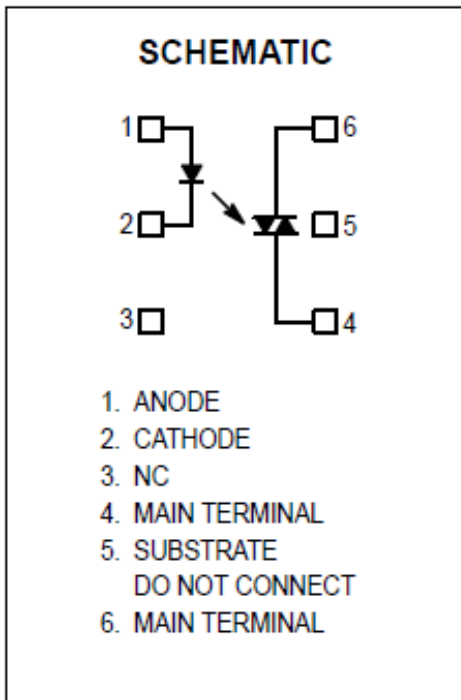
SCR para 4 A.



$V_{RRM}$  Sufixo: 1 - 30 V  
" 2 - 50 V  
" 3 - 100 V  
" 4 - 200 V

$I_{TSM} = 20 \text{ A}$   
 $V_{TM} (\text{max}) = 2 \text{ V}$   
 $I_T = 4 \text{ A}$   
 $I_{GT} = 200 \mu\text{A}$

MOC3021/3022/3023 TRIAC DRIVER



**MAXIMUM RATINGS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Rating	Symbol	Value	Unit
--------	--------	-------	------

**INFRARED EMITTING DIODE**

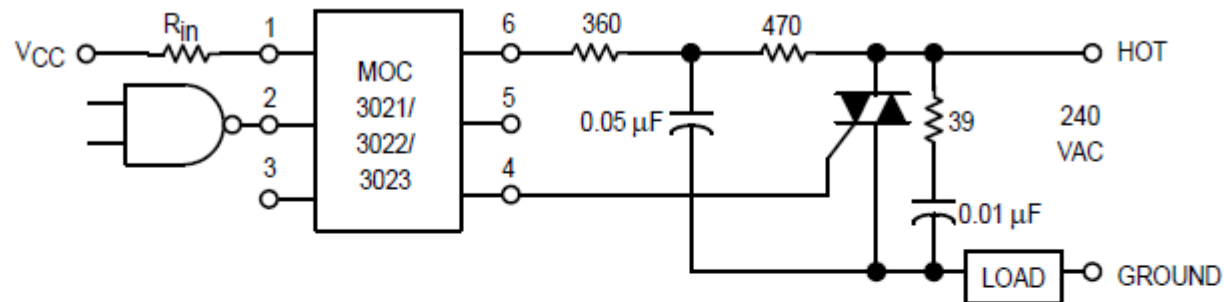
Reverse Voltage	$V_R$	3	Volts
Forward Current — Continuous	$I_F$	60	mA
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Negligible Power in Triac Driver Derate above $25^\circ\text{C}$	$P_D$	100	mW
		1.33	mW/ $^\circ\text{C}$

**OUTPUT DRIVER**

Off-State Output Terminal Voltage	$V_{DRM}$	400	Volts
Peak Repetitive Surge Current (PW = 1 ms, 120 pps)	$I_{TSM}$	1	A
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	300	mW
		4	mW/ $^\circ\text{C}$

**TOTAL DEVICE**

Isolation Surge Voltage <sup>(1)</sup> (Peak ac Voltage, 60 Hz, 1 Second Duration)	$V_{ISO}$	7500	Vac(pk)
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	330	mW
		4.4	mW/ $^\circ\text{C}$
Junction Temperature Range	$T_J$	-40 to +100	$^\circ\text{C}$
Ambient Operating Temperature Range <sup>(2)</sup>	$T_A$	-40 to +85	$^\circ\text{C}$
Storage Temperature Range <sup>(2)</sup>	$T_{stg}$	-40 to +150	$^\circ\text{C}$
Soldering Temperature (10 s)	$T_L$	260	$^\circ\text{C}$

**MOC3021 MOC3022 MOC3023**

\* This optoisolator should not be used to drive a load directly. It is intended to be a trigger device only.

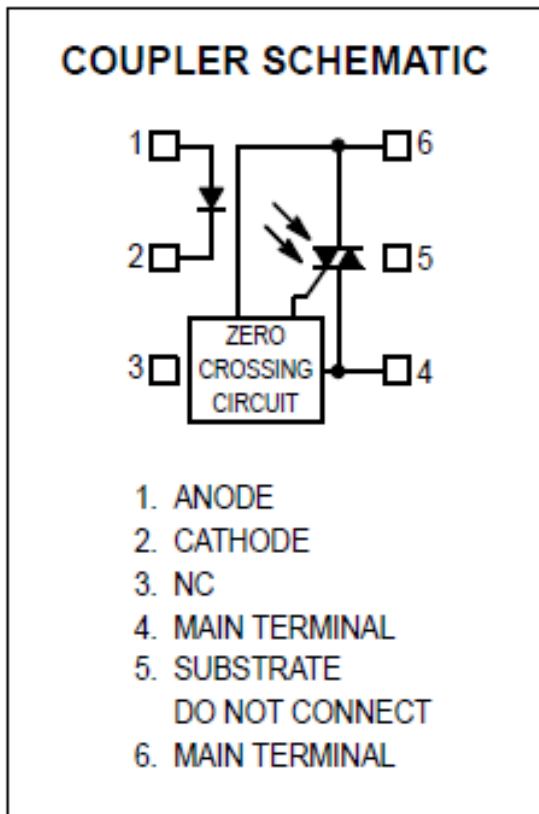
Additional information on the use of optically coupled triac drivers is available in Application Note AN-780A.

In this circuit the “hot” side of the line is switched and the load connected to the cold or ground side.

The 39 ohm resistor and 0.01  $\mu\text{F}$  capacitor are for snubbing of the triac, and the 470 ohm resistor and 0.05  $\mu\text{F}$  capacitor are for snubbing the coupler. These components may or may not be necessary depending upon the particular triac and load used.

**Figure 8. Typical Application Circuit**

MOC 3041 / 3042 / 3043 O MOC COM DETECÇÃO DA PASSAGEM POR ZERO



**MAXIMUM RATINGS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Rating	Symbol	Value	Unit
--------	--------	-------	------

**INFRARED EMITTING DIODE**

Reverse Voltage	$V_R$	6	Volts
Forward Current — Continuous	$I_F$	60	mA
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Negligible Power in Output Driver Derate above $25^\circ\text{C}$	$P_D$	120	mW
		1.41	mW/ $^\circ\text{C}$

**OUTPUT DRIVER**

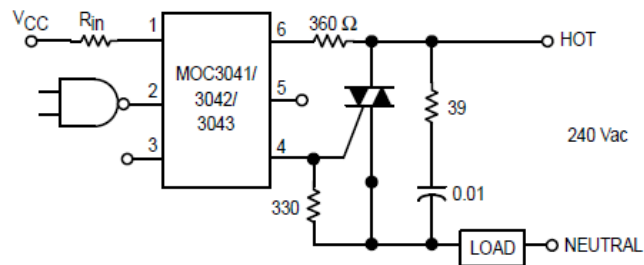
Off-State Output Terminal Voltage	$V_{DRM}$	400	Volts
Peak Repetitive Surge Current (PW = 100 $\mu\text{s}$ , 120 pps)	$I_{TSM}$	1	A
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	150	mW
		1.76	mW/ $^\circ\text{C}$

**TOTAL DEVICE**

Isolation Surge Voltage <sup>(1)</sup> (Peak ac Voltage, 60 Hz, 1 Second Duration)	$V_{ISO}$	7500	Vac(pk)
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	250	mW
		2.94	mW/ $^\circ\text{C}$
Junction Temperature Range	$T_J$	-40 to +100	$^\circ\text{C}$
Ambient Operating Temperature Range <sup>(2)</sup>	$T_A$	-40 to +85	$^\circ\text{C}$
Storage Temperature Range <sup>(2)</sup>	$T_{stg}$	-40 to +150	$^\circ\text{C}$
Soldering Temperature (10 s)	$T_L$	260	$^\circ\text{C}$



**MOC3041 MOC3042 MOC3043**

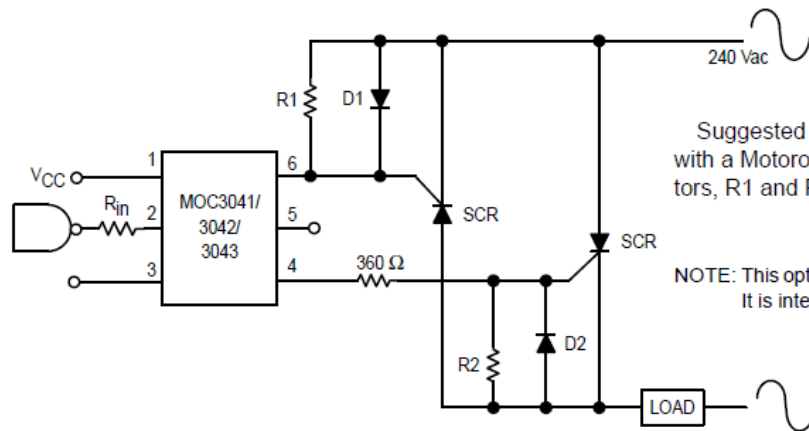


Typical circuit for use when hot line switching is required. In this circuit the "hot" side of the line is switched and the load connected to the cold or neutral side. The load may be connected to either the neutral or hot line.

$R_{in}$  is calculated so that  $I_F$  is equal to the rated  $I_{FT}$  of the part, 5 mA for the MOC3043, 10 mA for the MOC3042, or 15 mA for the MOC3041. The 39 ohm resistor and 0.01  $\mu$ F capacitor are for snubbing of the triac and may or may not be necessary depending upon the particular triac and load used.

\* For highly inductive loads (power factor < 0.5), change this value to 360 ohms.

Figure 8. Hot-Line Switching Application Circuit



Suggested method of firing two, back-to-back SCR's, with a Motorola triac driver. Diodes can be 1N4001; resistors, R1 and R2, are optional 330 ohms.

NOTE: This optoisolator should not be used to drive a load directly. It is intended to be a trigger device only.

Figure 9. Inverse-Parallel SCR Driver Circuit

## DIAC DB3/DB4

TIRISTOR BIDIRECIONAL.

**ABSOLUTE RATINGS**

Parameters	Symbols	DB3,DB4	UNITS
Power dissipation on printed $T_A=50^\circ\text{C}$ circuit (L=10mm)	$P_c$	150.0	mW
Repetitive peak on-state current $t_p=20\ \mu\text{S}$ $f=120\text{Hz}$	$I_{TRM}$	2.0	A
Operating junction temperature	$T_J$	-40--- +125	$^\circ\text{C}$
Storage temperature	$T_{STG}$	-40--- +125	$^\circ\text{C}$

**ELECTRICAL CHARACTERISTICS**

Parameters		Test Conditions	DB3	DB4	UNITS	
Breakover voltage (NOTE 1)	$V_{BO}$	C=22nf(NOTE 2) See FIG.1	Min	28	35	V
			Typ	32	40	
			Max	36	45	
Breakover voltage symmetry	$I+V_{BO} I-$ $I-V_{BO} I$	C=22nf(NOTE 2) See FIG.1	Max	$\pm 3.0$		V
Dynamic breakover voltage (NOTE 1)	$I \pm \Delta V$	$\Delta I=(I_{BO} \text{ to } I_F=10\text{mA})$ See FIG.1	Min	5.0		V
Output voltage (NOTE 1)	$V_o$	See FIG.2	Min	5.0		V
Breakover current (NOTE 1)	$I_{BO}$	C=22nf(NOTE 2)	Max	100.0		$\mu A$
Rise time (NOTE 1)	$t_r$	See FIG.3	Typ	1.5		$\mu S$
Leakage current (NOTE 1)	$I_R$	$V_R=0.5 V_{BO}$ See FIG.1	Max	10.0		$\mu A$

NOTE: 1.Electrical characteristics applicable in both forward and reverse directions.

[www.diode.co.kr](http://www.diode.co.kr)

2.Connected in parallel with the devices

## PUT BRY39P

O mais popular, mas pouco usado atualmente.

$V_{GA}$ .....70 V  
 $I_A$ .....175 mA  
 $I_{ARM}$ .....2,5 A  
 $di_A/dt$ .....20 A/ $\mu$ s  
 $I_p$  (max).....5 mA  
 $I_V$  (min).....25 mA  
 $t_r$  (max).....80 ns  
 (Características medidas a  $R_g = 10$  k)

a = ANODO  
 ag = PORTA DE ANODO  
 K = CATODO  
 Kg = PORTA DE CATODO

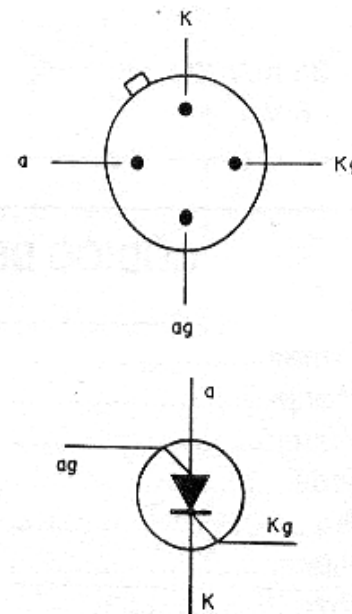


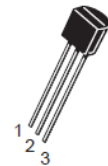
Figura 13 – O BRY39, PUT de uso geral

PUT 2N6027 2N6028

ON Semiconductor

<http://onsemi.com>

PUTs  
40 VOLTS  
300 mW



TO-92 (TO-226AA)  
CASE 029  
STYLE 16

PIN ASSIGNMENT

1	Anode
2	Gate
3	Cathode

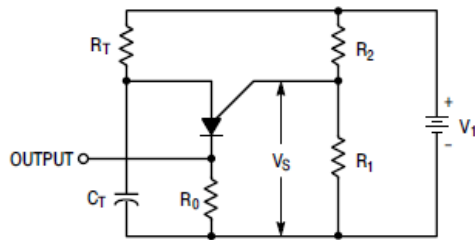


Figure 3.29(a). Typical Oscillator Circuit

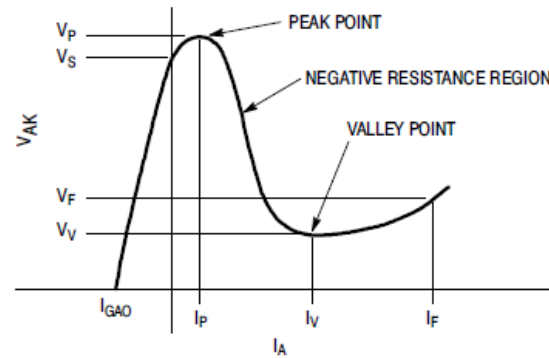


Figure 3.29(b). Static Characteristics

Table 3.1. Typical PUT Characteristics

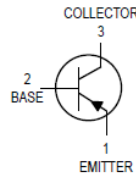
Symbol	Test Circuit Figure	Test Conditions	2N6027	2N6028	Unit
$I_P$	3.30	$R_G = 1\text{ m}\Omega$ $R_G = 10\text{ k}\Omega$	1.25 4	0.08 0.70	$\mu\text{A}$ $\mu\text{A}$
$I_V$	3.30	$R_G = 1\text{ M}\Omega$ $R_G = 10\text{ k}\Omega$	18 150	18 150	$\mu\text{A}$ $\mu\text{A}$
$V_{AG}$		(See Figure 3.31)			
$I_{GAO}$		$V_S = 40\text{ V}$	(See Figure 3.32)		
$I_{GKS}$		$V_S = 40\text{ V}$	5	5	nA
$V_F$	Curve Tracer Used	$I_F = 50\text{ mA}$	0.8	0.8	V
$V_O$	3.33		11	11	V
$t_r$	3.34		40	40	ns

TRANSISTOR

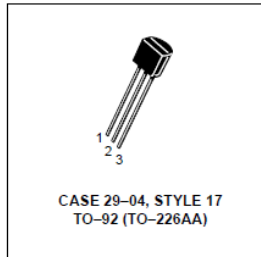
TRANSISTOR DE USO GERAL

BC212 13 14 PNP 0,1A

**Amplifier Transistors**  
PNP Silicon



**BC212,B**  
**BC213**  
**BC214**



**MAXIMUM RATINGS**

Rating	Symbol	BC 212	BC 213	BC 214	Unit
Collector–Emitter Voltage	V <sub>CEO</sub>	-50	-30	-30	Vdc
Collector–Base Voltage	V <sub>CBO</sub>	-60	-45	-45	Vdc
Emitter–Base Voltage	V <sub>EBO</sub>	-5.0			Vdc
Collector Current — Continuous	I <sub>C</sub>	-100			mAdc
Total Device Dissipation @ T <sub>A</sub> = 25°C Derate above 25°C	P <sub>D</sub>	350	2.8		mW mW/°C
Total Device Dissipation @ T <sub>C</sub> = 25°C Derate above 25°C	P <sub>D</sub>	1.0	8.0		Watts mW/°C
Operating and Storage Junction Temperature Range	T <sub>J</sub> , T <sub>stg</sub>	-55 to +150			°C

**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Ambient	R <sub>θJA</sub>	357	°C/W
Thermal Resistance, Junction to Case	R <sub>θJC</sub>	125	°C/W

**ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = 25°C unless otherwise noted)

Characteristic		Symbol	Min	Typ	Max	Unit
Collector–Emitter Breakdown Voltage (I <sub>C</sub> = -2.0 mAdc, I <sub>B</sub> = 0)	BC212 BC213 BC214	V <sub>(BR)CEO</sub>	-50 -30 -30	— — —	— — —	Vdc
Collector–Base Breakdown Voltage (I <sub>C</sub> = -10 μA, I <sub>E</sub> = 0)	BC212 BC213 BC214	V <sub>(BR)CBO</sub>	-60 -45 -45	— — —	— — —	Vdc
Emitter–Base Breakdown Voltage (I <sub>E</sub> = -10 μAdc, I <sub>C</sub> = 0)	BC212 BC213 BC214	V <sub>(BR)EBO</sub>	-5 -5 -5	— — —	— — —	Vdc
Collector–Emitter Leakage Current (V <sub>CB</sub> = -30 V)	BC212 BC213 BC214	I <sub>CBO</sub>	— — —	— — —	-15 -15 -15	nAdc
Emitter–Base Leakage Current (V <sub>EB</sub> = -4.0 V, I <sub>C</sub> = 0)	BC212 BC213 BC214	I <sub>EBO</sub>	— — —	— — —	-15 -15 -15	nAdc

**ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = 25°C unless otherwise noted) (Continued)

Characteristic		Symbol	Min	Typ	Max	Unit
<b>ON CHARACTERISTICS</b>						
DC Current Gain (I <sub>C</sub> = -10 μAdc, V <sub>CE</sub> = -5.0 Vdc)	BC212 BC213 BC214	h <sub>FE</sub>	40 40 100	— — —	— — —	—
(I <sub>C</sub> = -2.0 mAdc, V <sub>CE</sub> = -5.0 Vdc)	BC212 BC213 BC214		60 80 140	— — —	— — 600	
(I <sub>C</sub> = -100 mAdc, V <sub>CE</sub> = -5.0 Vdc) <sup>(1)</sup>	BC212, BC214 BC213		— —	120 140	— —	
Collector–Emitter Saturation Voltage (I <sub>C</sub> = -10 mAdc, I <sub>B</sub> = -0.5 mAdc) (I <sub>C</sub> = -100 mAdc, I <sub>B</sub> = -5.0 mAdc) <sup>(1)</sup>		V <sub>CE(sat)</sub>	— —	-0.10 -0.25	— -0.6	Vdc
Base–Emitter Saturation Voltage (I <sub>C</sub> = -100 mAdc, I <sub>B</sub> = -5.0 mAdc)		V <sub>BE(sat)</sub>	—	-1.0	-1.4	Vdc
Base–Emitter On Voltage (I <sub>C</sub> = -2.0 mAdc, V <sub>CE</sub> = -5.0 Vdc)		V <sub>BE(on)</sub>	-0.6	-0.62	-0.72	Vdc



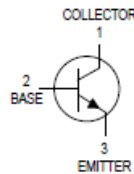
**DYNAMIC CHARACTERISTICS**

Current–Gain — Bandwidth Product ( $I_C = -10 \text{ mA dc}$ , $V_{CE} = -5.0 \text{ V dc}$ , $f = 100 \text{ MHz}$ )	BC212 BC214 BC213	$f_T$	— — —	280 320 360	— — —	MHz
Common–Base Output Capacitance ( $V_{CB} = -10 \text{ V dc}$ , $I_C = 0$ , $f = 1.0 \text{ MHz}$ )		$C_{ob}$	—	—	6.0	pF
Noise Figure ( $I_C = -0.2 \text{ mA dc}$ , $V_{CE} = -5.0 \text{ V dc}$ , $R_S = 2.0 \text{ k}\Omega$ , $f = 1.0 \text{ kHz}$ )	BC214	NF	—	—	2	dB
( $I_C = -0.2 \text{ mA dc}$ , $V_{CE} = -5.0 \text{ V dc}$ , $R_S = 2.0 \text{ k}\Omega$ , $f = 1.0 \text{ kHz}$ , $f = 200 \text{ Hz}$ )	BC212, BC213		—	—	10	
Small–Signal Current Gain ( $I_C = -2.0 \text{ mA dc}$ , $V_{CE} = -5.0 \text{ V dc}$ , $f = 1.0 \text{ kHz}$ )	BC212 BC213 BC214 BC212B	$h_{fe}$	60 80 140 200	— — — —	— — — 400	—

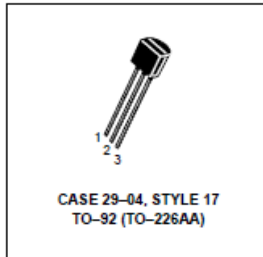
1. Pulse Test:  $T_p$  300 s, Duty Cycle 2.0%.

BC547 58 NPN 0,1A

**Amplifier Transistors**  
NPN Silicon



**BC546, B**  
**BC547, A, B, C**  
**BC548, A, B, C**



**MAXIMUM RATINGS**

Rating	Symbol	BC 546	BC 547	BC 548	Unit
Collector-Emitter Voltage	V <sub>CEO</sub>	65	45	30	V <sub>dC</sub>
Collector-Base Voltage	V <sub>CB0</sub>	80	50	30	V <sub>dC</sub>
Emitter-Base Voltage	V <sub>EBO</sub>	6.0			V <sub>dC</sub>
Collector Current — Continuous	I <sub>C</sub>	100			mA <sub>dC</sub>
Total Device Dissipation @ T <sub>A</sub> = 25°C Derate above 25°C	P <sub>D</sub>	625	5.0		mW mW/°C
Total Device Dissipation @ T <sub>C</sub> = 25°C Derate above 25°C	P <sub>D</sub>	1.5	12		Watt mW/°C
Operating and Storage Junction Temperature Range	T <sub>J</sub> , T <sub>stg</sub>	-55 to +150			°C

**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Ambient	R <sub>θJA</sub>	200	°C/W
Thermal Resistance, Junction to Case	R <sub>θJC</sub>	83.3	°C/W

**ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = 25°C unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Collector-Emitter Breakdown Voltage (I <sub>C</sub> = 1.0 mA, I <sub>B</sub> = 0)	V <sub>(BR)CEO</sub>	65	—	—	V
	BC547	45	—	—	
	BC548	30	—	—	
Collector-Base Breakdown Voltage (I <sub>C</sub> = 100 μA <sub>dC</sub> )	V <sub>(BR)CBO</sub>	80	—	—	V
	BC547	50	—	—	
	BC548	30	—	—	
Emitter-Base Breakdown Voltage (I <sub>E</sub> = 10 μA, I <sub>C</sub> = 0)	V <sub>(BR)EBO</sub>	6.0	—	—	V
	BC547	6.0	—	—	
	BC548	6.0	—	—	
Collector Cutoff Current (V <sub>CE</sub> = 70 V, V <sub>BE</sub> = 0)	I <sub>CES</sub>	—	0.2	15	nA
	BC547	—	0.2	15	
	BC548	—	0.2	15	
	BC546/547/548	—	—	4.0	μA

**ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = 25°C unless otherwise noted) (Continued)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>ON CHARACTERISTICS</b>					
DC Current Gain (I <sub>C</sub> = 10 μA, V <sub>CE</sub> = 5.0 V)	h <sub>FE</sub>	—	90	—	—
	BC547A/548A	—	150	—	
	BC546B/547B/548B	—	270	—	
	BC548C	—	—	—	
(I <sub>C</sub> = 2.0 mA, V <sub>CE</sub> = 5.0 V)	BC546	110	—	450	
	BC547	110	—	800	
	BC548	110	—	800	
	BC547A/548A	110	180	220	
	BC546B/547B/548B	200	290	450	
	BC547C/BC548C	420	520	800	
(I <sub>C</sub> = 100 mA, V <sub>CE</sub> = 5.0 V)	BC547A/548A	—	120	—	
	BC546B/547B/548B	—	180	—	
	BC548C	—	300	—	
Collector-Emitter Saturation Voltage (I <sub>C</sub> = 10 mA, I <sub>B</sub> = 0.5 mA)	V <sub>CE(sat)</sub>	—	0.09	0.25	V
(I <sub>C</sub> = 100 mA, I <sub>B</sub> = 5.0 mA)		—	0.2	0.6	
(I <sub>C</sub> = 10 mA, I <sub>B</sub> = See Note 1)		—	0.3	0.6	
Base-Emitter Saturation Voltage (I <sub>C</sub> = 10 mA, I <sub>B</sub> = 0.5 mA)	V <sub>BE(sat)</sub>	—	0.7	—	V
Base-Emitter On Voltage (I <sub>C</sub> = 2.0 mA, V <sub>CE</sub> = 5.0 V)	V <sub>BE(on)</sub>	0.55	—	0.7	V
(I <sub>C</sub> = 10 mA, V <sub>CE</sub> = 5.0 V)		—	—	0.77	

**SMALL-SIGNAL CHARACTERISTICS**

Current-Gain — Bandwidth Product ( $I_C = 10 \text{ mA}$ , $V_{CE} = 5.0 \text{ V}$ , $f = 100 \text{ MHz}$ )	BC546 BC547 BC548	$f_T$	150 150 150	300 300 300	— — —	MHz
Output Capacitance ( $V_{CB} = 10 \text{ V}$ , $I_C = 0$ , $f = 1.0 \text{ MHz}$ )		$C_{obo}$	—	1.7	4.5	pF
Input Capacitance ( $V_{EB} = 0.5 \text{ V}$ , $I_C = 0$ , $f = 1.0 \text{ MHz}$ )		$C_{ibo}$	—	10	—	pF
Small-Signal Current Gain ( $I_C = 2.0 \text{ mA}$ , $V_{CE} = 5.0 \text{ V}$ , $f = 1.0 \text{ kHz}$ )	BC546 BC547/548 BC547A/548A BC546B/547B/548B BC547C/548C	$h_{fe}$	125 125 125 240 450	— — 220 330 600	500 900 260 500 900	—
Noise Figure ( $I_C = 0.2 \text{ mA}$ , $V_{CE} = 5.0 \text{ V}$ , $R_S = 2 \text{ k}\Omega$ , $f = 1.0 \text{ kHz}$ , $\Delta f = 200 \text{ Hz}$ )	BC546 BC547 BC548	NF	— — —	2.0 2.0 2.0	10 10 10	dB

Note 1:  $I_B$  is value for which  $I_C = 11 \text{ mA}$  at  $V_{CE} = 1.0 \text{ V}$ .

BC549 BC550 ALTO GANHO

Philips  
Semiconductors



**PHILIPS**

Philips Semiconductors

Product specification

**NPN general purpose transistors**

**BC549; BC550**

**FEATURES**

- Low current (max. 100 mA)
- Low voltage (max. 45 V).

**APPLICATIONS**

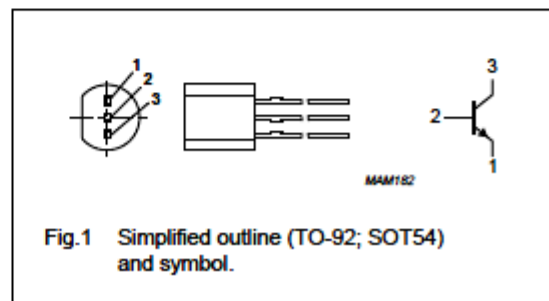
- Low noise stages in audio frequency equipment.

**DESCRIPTION**

NPN transistor in a TO-92; SOT54 plastic package.  
PNP complements: BC559 and BC560.

**PINNING**

PIN	DESCRIPTION
1	emitter
2	base
3	collector



## NPN general purpose transistors

BC549; BC550

## THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
$R_{th(j-a)}$	thermal resistance from junction to ambient	note 1	250	K/W

## Note

1. Transistor mounted on an FR4 printed-circuit board.

## CHARACTERISTICS

$T_j = 25\text{ }^\circ\text{C}$  unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{CBO}$	collector cut-off current	$I_E = 0; V_{CB} = 30\text{ V}$	–	–	15	nA
		$I_E = 0; V_{CB} = 30\text{ V}; T_j = 150\text{ }^\circ\text{C}$	–	–	5	$\mu\text{A}$
$I_{EBO}$	emitter cut-off current	$I_C = 0; V_{EB} = 5\text{ V}$	–	–	100	nA
$h_{FE}$	DC current gain BC549C; BC550C	$I_C = 10\text{ }\mu\text{A}; V_{CE} = 5\text{ V};$ see Fig.2	–	270	–	
		$I_C = 2\text{ mA}; V_{CE} = 5\text{ V};$ see Fig.2	420	520	800	
$V_{CEsat}$	collector-emitter saturation voltage	$I_C = 10\text{ mA}; I_B = 0.5\text{ mA}$	–	90	250	mV
		$I_C = 100\text{ mA}; I_B = 5\text{ mA}$	–	200	600	mV
$V_{BEsat}$	base-emitter saturation voltage	$I_C = 10\text{ mA}; I_B = 0.5\text{ mA};$ note 1	–	700	–	mV
		$I_C = 100\text{ mA}; I_B = 5\text{ mA};$ note 1	–	900	–	mV
$V_{BE}$	base-emitter voltage	$I_C = 2\text{ mA}; V_{CE} = 5\text{ V};$ note 2	580	660	700	mV
		$I_C = 10\text{ mA}; V_{CE} = 5\text{ V};$ note 2	–	–	770	mV
$C_C$	collector capacitance	$I_E = I_C = 0; V_{CB} = 10\text{ V}; f = 1\text{ MHz}$	–	1.5	–	pF
$C_E$	emitter capacitance	$I_C = I_E = 0; V_{EB} = 0.5\text{ V}; f = 1\text{ MHz}$	–	11	–	pF
$f_T$	transition frequency	$I_C = 10\text{ mA}; V_{CE} = 5\text{ V};$ $f = 100\text{ MHz}$	100	–	–	MHz
F	noise figure	$I_C = 200\text{ }\mu\text{A}; V_{CE} = 5\text{ V};$ $R_S = 2\text{ k}\Omega; f = 10\text{ Hz to }15.7\text{ kHz}$	–	–	4	dB
		$I_C = 200\text{ }\mu\text{A}; V_{CE} = 5\text{ V};$ $R_S = 2\text{ k}\Omega; f = 1\text{ kHz}; B = 200\text{ Hz}$	–	–	4	dB

## Notes

1.  $V_{BEsat}$  decreases by about 1.7 mV/K with increasing temperature.
2.  $V_{BE}$  decreases by about 2 mV/K with increasing temperature.

BC337 38 NPN 0,8A

**BC337, BC337-25,  
BC337-40**

**Amplifier Transistors**

NPN Silicon

Features

- These are Pb-Free Devices

MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Collector - Emitter Voltage	$V_{CE0}$	45	Vdc
Collector - Base Voltage	$V_{CB0}$	50	Vdc
Emitter - Base Voltage	$V_{EB0}$	5.0	Vdc
Collector Current - Continuous	$I_C$	800	mA dc
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	625 5.0	mW mW/ $^\circ\text{C}$
Total Device Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	1.5 12	W mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-55 to +150	$^\circ\text{C}$

THERMAL CHARACTERISTICS

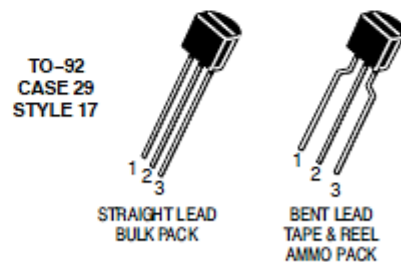
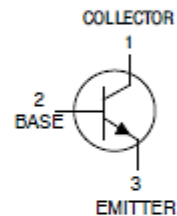
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction-to-Ambient	$R_{\theta JA}$	200	$^\circ\text{C}/\text{W}$
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$	83.3	$^\circ\text{C}/\text{W}$

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.



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MARKING DIAGRAM

**ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Collector–Emitter Breakdown Voltage ( $I_C = 10\text{ mA}$ , $I_B = 0$ )	$V_{(BR)CEO}$	45	–	–	Vdc
Collector–Emitter Breakdown Voltage ( $I_C = 100\ \mu\text{A}$ , $I_E = 0$ )	$V_{(BR)CES}$	50	–	–	Vdc
Emitter–Base Breakdown Voltage ( $I_E = 10\ \mu\text{A}$ , $I_C = 0$ )	$V_{(BR)EBO}$	5.0	–	–	Vdc
Collector Cutoff Current ( $V_{CB} = 30\text{ V}$ , $I_E = 0$ )	$I_{CBO}$	–	–	100	nAdc
Collector Cutoff Current ( $V_{CE} = 45\text{ V}$ , $V_{BE} = 0$ )	$I_{CES}$	–	–	100	nAdc
Emitter Cutoff Current ( $V_{EB} = 4.0\text{ V}$ , $I_C = 0$ )	$I_{EBO}$	–	–	100	nAdc
<b>ON CHARACTERISTICS</b>					
DC Current Gain ( $I_C = 100\text{ mA}$ , $V_{CE} = 1.0\text{ V}$ )	$h_{FE}$	100	–	630	–
		160	–	400	
		250	–	630	
		60	–	–	
Base–Emitter On Voltage ( $I_C = 300\text{ mA}$ , $V_{CE} = 1.0\text{ V}$ )	$V_{BE(on)}$	–	–	1.2	Vdc
Collector–Emitter Saturation Voltage ( $I_C = 500\text{ mA}$ , $I_B = 50\text{ mA}$ )	$V_{CE(sat)}$	–	–	0.7	Vdc
<b>SMALL-SIGNAL CHARACTERISTICS</b>					
Output Capacitance ( $V_{CB} = 10\text{ V}$ , $I_E = 0$ , $f = 1.0\text{ MHz}$ )	$C_{ob}$	–	15	–	pF
Current–Gain – Bandwidth Product ( $I_C = 10\text{ mA}$ , $V_{CE} = 5.0\text{ V}$ , $f = 100\text{ MHz}$ )	$f_T$	–	210	–	MHz

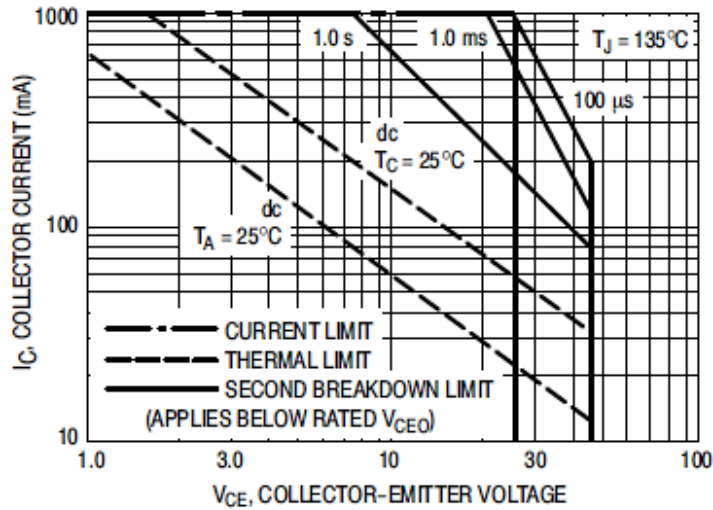


Figure 2. Active Region - Safe Operating Area

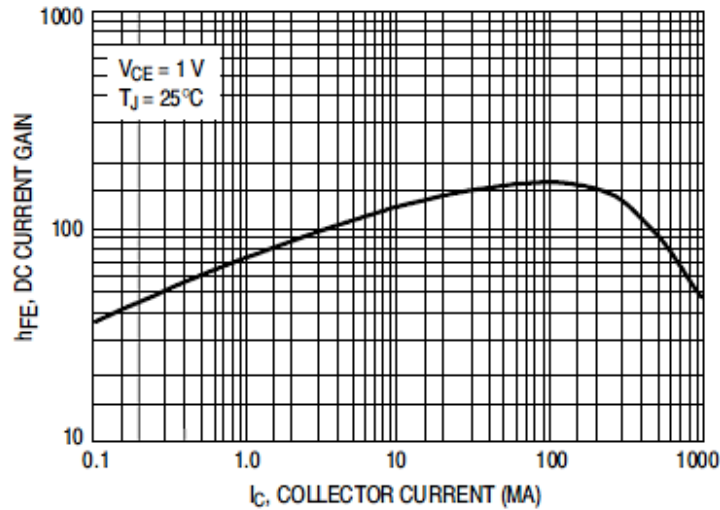


Figure 3. DC Current Gain

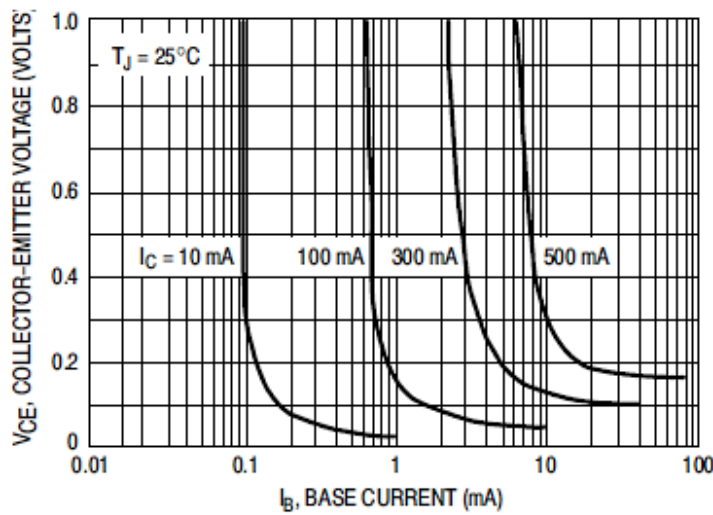


Figure 4. Saturation Region

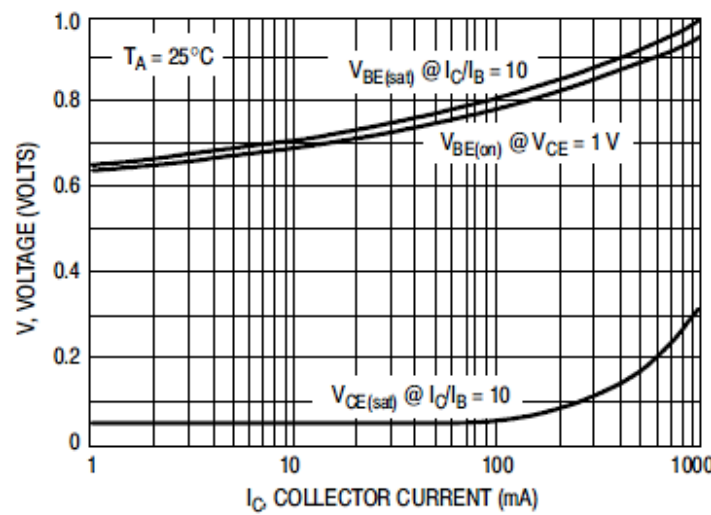


Figure 5. "On" Voltages



BC327 BC328 PNP 0,8A

**BC327, BC327-16,  
BC327-25, BC327-40**

**Amplifier Transistors**

PNP Silicon

**Features**

- Pb-Free Packages are Available\*

**MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Collector-Emmitter Voltage	$V_{CEO}$	-45	Vdc
Collector-Base Voltage	$V_{CES}$	-50	Vdc
Collector-Emmitter Voltage	$V_{EBO}$	-5.0	Vdc
Collector Current - Continuous	$I_C$	-800	mAdc
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $T_A = 25^\circ\text{C}$	$P_D$	625 5.0	mW mW/°C
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $T_A = 25^\circ\text{C}$	$P_D$	1.5 12	W mW/°C
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-55 to +150	°C

**THERMAL CHARACTERISTICS**

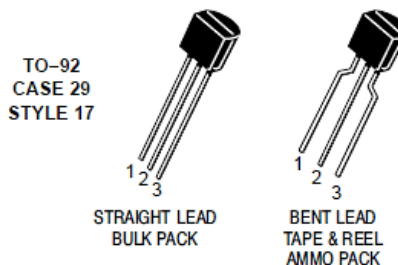
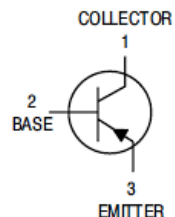
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction-to-Ambient	$R_{\theta JA}$	200	°C/W
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$	83.3	°C/W

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.



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**MARKING DIAGRAM**

**ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Collector–Emitter Breakdown Voltage ( $I_C = -10\text{ mA}$ , $I_B = 0$ )	$V_{(BR)CEO}$	-45	-	-	Vdc
Collector–Emitter Breakdown Voltage ( $I_C = -100\ \mu\text{A}$ , $I_E = 0$ )	$V_{(BR)CES}$	-50	-	-	Vdc
Emitter–Base Breakdown Voltage ( $I_E = -10\ \mu\text{A}$ , $I_C = 0$ )	$V_{(BR)EBO}$	-5.0	-	-	Vdc
Collector Cutoff Current ( $V_{CB} = -30\text{ V}$ , $I_E = 0$ )	$I_{CBO}$	-	-	-100	nAdc
Collector Cutoff Current ( $V_{CE} = -45\text{ V}$ , $V_{BE} = 0$ )	$I_{CES}$	-	-	-100	nAdc
Emitter Cutoff Current ( $V_{EB} = -4.0\text{ V}$ , $I_C = 0$ )	$I_{EBO}$	-	-	-100	nAdc
<b>ON CHARACTERISTICS</b>					
DC Current Gain ( $I_C = -100\text{ mA}$ , $V_{CE} = -1.0\text{ V}$ )	$h_{FE}$	100	-	630	-
		100	-	250	
		160	-	400	
		250	-	630	
( $I_C = -300\text{ mA}$ , $V_{CE} = -1.0\text{ V}$ )		40	-	-	
Base–Emitter On Voltage ( $I_C = -300\text{ mA}$ , $V_{CE} = -1.0\text{ V}$ )	$V_{BE(on)}$	-	-	-1.2	Vdc
Collector–Emitter Saturation Voltage ( $I_C = -500\text{ mA}$ , $I_B = -50\text{ mA}$ )	$V_{CE(sat)}$	-	-	-0.7	Vdc
<b>SMALL–SIGNAL CHARACTERISTICS</b>					
Output Capacitance ( $V_{CB} = -10\text{ V}$ , $I_E = 0$ , $f = 1.0\text{ MHz}$ )	$C_{ob}$	-	11	-	pF
Current–Gain – Bandwidth Product ( $I_C = -10\text{ mA}$ , $V_{CE} = -5.0\text{ V}$ , $f = 100\text{ MHz}$ )	$f_T$	-	260	-	MHz

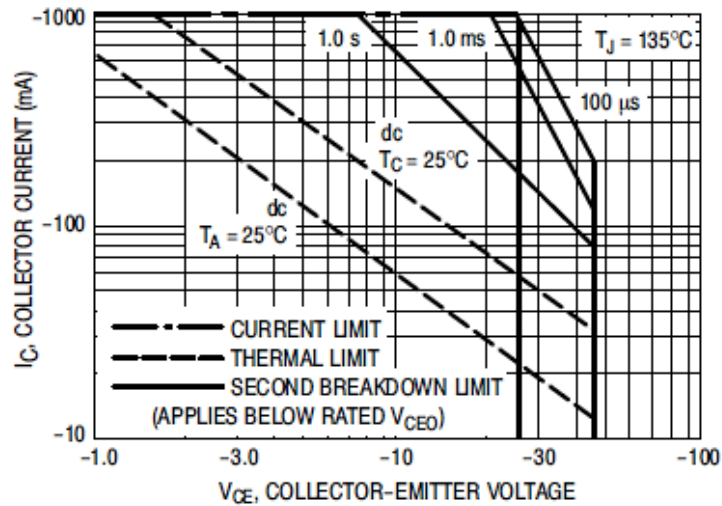


Figure 2. Active Region - Safe Operating Area

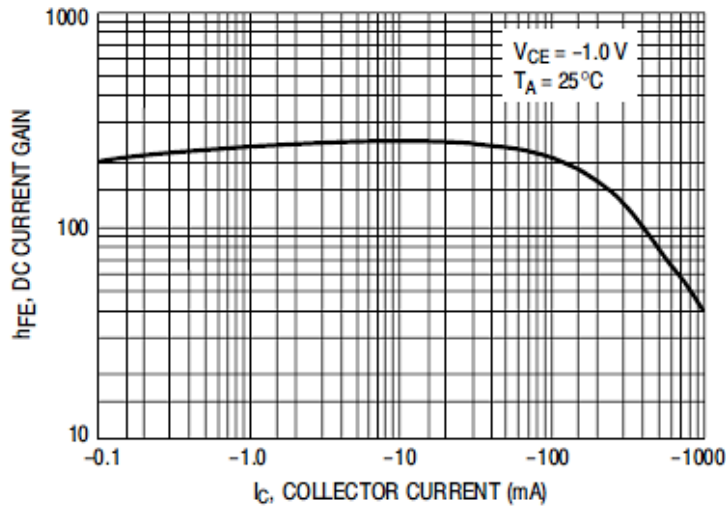


Figure 3. DC Current Gain

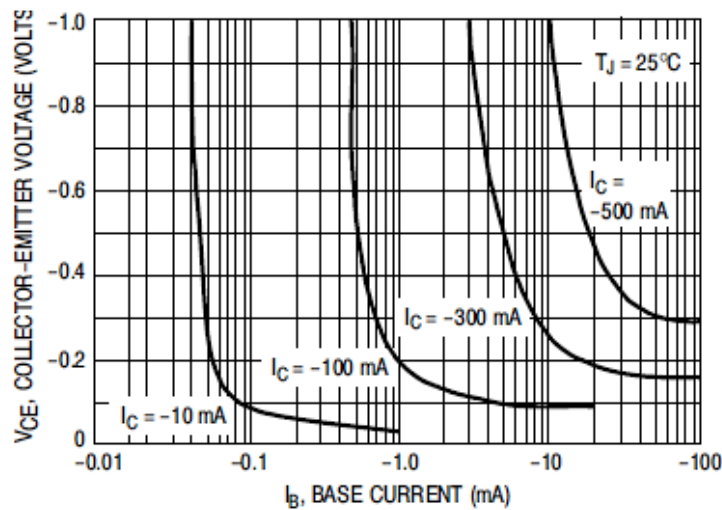


Figure 4. Saturation Region

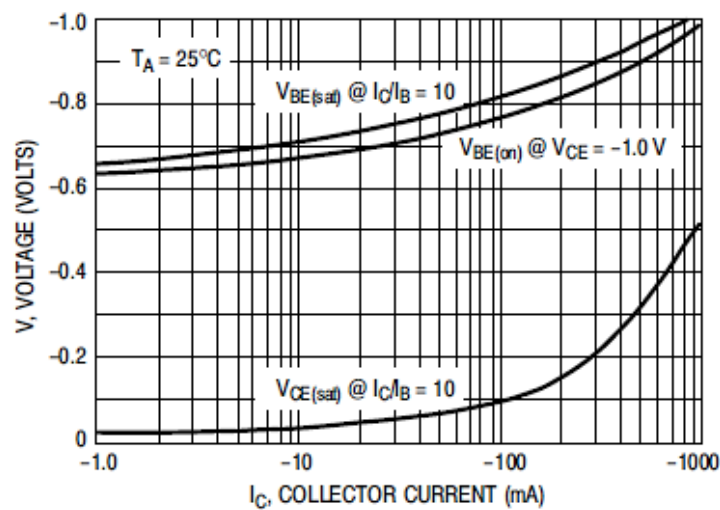


Figure 5. "On" Voltages

BC517 DARLINGTON 1A HFE 30000

**BC517**

**Darlington Transistors**

**NPN Silicon**

**Features**

- Pb-Free Packages are Available\*

**MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Collector - Emitter Voltage	$V_{CES}$	30	Vdc
Collector - Base Voltage	$V_{CB}$	40	Vdc
Collector - Emitter Voltage	$V_{EB}$	10	Vdc
Collector Current - Continuous	$I_C$	1.0	Adc
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $T_A = 25^\circ\text{C}$	$P_D$	625 12	mW mW/°C
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $T_A = 25^\circ\text{C}$	$P_D$	1.5 12	W mW/°C
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-55 to +150	°C

**THERMAL CHARACTERISTICS**

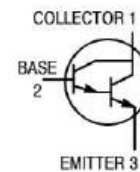
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction-to-Ambient	$R_{\theta JA}$	200	°C/W
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$	83.3	°C/W

Maximum ratings are those values beyond which device damage can occur. Maximum ratings applied to the device are individual stress limit values (not normal operating conditions) and are not valid simultaneously. If these limits are exceeded, device functional operation is not implied, damage may occur and reliability may be affected.



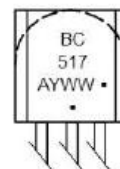
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TO-92  
CASE 29  
STYLE 17

**MARKING DIAGRAM**



**BC517****ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Collector - Emitter Breakdown Voltage ( $I_C = 2.0\text{ mAdc}$ , $I_{BE} = 0$ )	$V_{(BR)CES}$	30	-	-	Vdc
Collector - Base Breakdown Voltage ( $I_C = 10\text{ }\mu\text{Adc}$ , $I_E = 0$ )	$V_{(BR)CBO}$	40	-	-	Vdc
Emitter - Base Breakdown Voltage ( $I_E = 100\text{ }\mu\text{Adc}$ , $I_C = 0$ )	$V_{(BR)EBO}$	10	-	-	Vdc
Collector Cutoff Current ( $V_{CE} = 30\text{ Vdc}$ )	$I_{CES}$	-	-	500	nAdc
Collector Cutoff Current ( $V_{CB} = 30\text{ Vdc}$ , $I_E = 0$ )	$I_{CBO}$	-	-	100	nAdc
Emitter Cutoff Current ( $V_{CB} = 10\text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	-	-	100	nAdc
<b>ON CHARACTERISTICS</b> (Note 1)					
DC Current Gain ( $I_C = 20\text{ mAdc}$ , $V_{CE} = 2.0\text{ Vdc}$ )	$h_{FE}$	30,000	-	-	-
Collector - Emitter Saturation Voltage ( $I_C = 100\text{ mAdc}$ , $I_B = 0.1\text{ mAdc}$ )	$V_{CE(sat)}$	-	-	1.0	Vdc
Collector - Emitter Saturation Voltage ( $I_C = 10\text{ mAdc}$ , $V_{CE} = 5.0\text{ Vdc}$ )	$V_{BE(on)}$	-	-	1.4	Vdc
<b>SMALL-SIGNAL CHARACTERISTICS</b>					
Current-Gain - Bandwidth Product (Note 2) ( $I_C = 10\text{ mAdc}$ , $V_{CE} = 5.0\text{ Vdc}$ , $f = 100\text{ MHz}$ )	$f_T$	-	200	-	MHz

1. Pulse Test: Pulse Width = 300  $\mu\text{s}$ , Duty Cycle 2.0%.

2.  $f_T = |h_{re}| \cdot f_{test}$

O transistor BC517 pode ser substituído pelos seguintes transistores: BC875, BC879, BC517G, BC517P, GC195.

CA3083 MATRIZ TRANSISTOR 0,1A GENERAL PORPOUSE



CA3083

Data Sheet

September 1998

File Number 481.4

**General Purpose High Current NPN Transistor Array**

The CA3083 is a versatile array of five high current (to 100mA) NPN transistors on a common monolithic substrate. In addition, two of these transistors (Q<sub>1</sub> and Q<sub>2</sub>) are matched at low current (i.e., 1mA) for applications in which offset parameters are of special importance.

Independent connections for each transistor plus a separate terminal for the substrate permit maximum flexibility in circuit design.

**Ordering Information**

PART NUMBER (BRAND)	TEMP. RANGE (°C)	PACKAGE	PKG. NO.
CA3083	-55 to 125	16 Ld PDIP	E16.3
CA3083M (3083)	-55 to 125	16 Ld SOIC	M16.15
CA3083M96 (3083)	-55 to 125	16 Ld SOIC Tape and Reel	M16.15

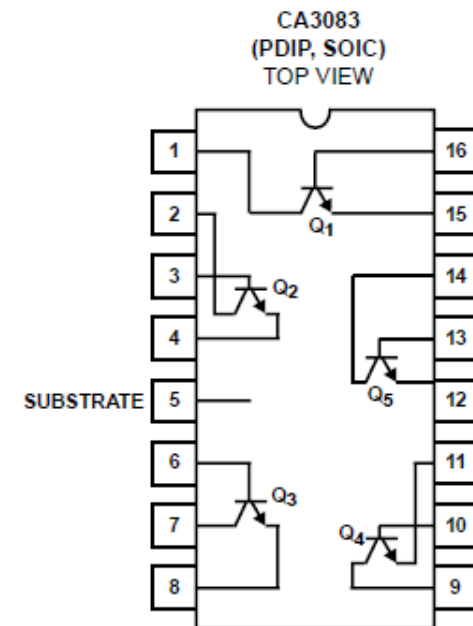
**Features**

- High I<sub>C</sub> ..... 100mA (Max)
- Low V<sub>CE sat</sub> (at 50mA) ..... 0.7V (Max)
- Matched Pair (Q<sub>1</sub> and Q<sub>2</sub>)
  - V<sub>IO</sub> (V<sub>BE Match</sub>) ..... ±5mV (Max)
  - I<sub>IO</sub> (at 1mA) ..... 2.5µA (Max)
- 5 Independent Transistors Plus Separate Substrate Connection

**Applications**

- Signal Processing and Switching Systems Operating from DC to VHF
- Lamp and Relay Driver
- Differential Amplifier
- Temperature Compensated Amplifier
- Thyristor Firing
- See Application Note AN5296 "Applications of the CA3018 Circuit Transistor Array" for Suggested Applications

**Pinout**



**CA3083****Absolute Maximum Ratings**

The following ratings apply for each transistor in the device:

Collector-to-Emitter Voltage, $V_{CE0}$ .....	15V
Collector-to-Base Voltage, $V_{CBO}$ .....	20V
Collector-to-Substrate Voltage, $V_{CIO}$ (Note 1) .....	20V
Emitter-to-Base Voltage, $V_{EBO}$ .....	5V
Collector Current ( $I_C$ ) .....	100mA
Base Current ( $I_B$ ) .....	20mA

**Operating Conditions**

Temperature Range .....

-55°C to 125°C

*CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.*

**NOTES:**

1. The collector of each transistor of the CA3083 is isolated from the substrate by an integral diode. The substrate must be connected to a voltage which is more negative than any collector voltage in order to maintain isolation between transistors and provide normal transistor action. To avoid undesired coupling between transistors, the substrate Terminal (5) should be maintained at either DC or signal (AC) ground. A suitable bypass capacitor can be used to establish a signal ground.
2.  $\theta_{JA}$  is measured with the component mounted on an evaluation PC board in free air.

**Thermal Information**

Thermal Resistance (Typical, Note 2)	$\theta_{JA}$ (°C/W)	$\theta_{JC}$ (°C/W)
PDIP Package .....	135	N/A
SOIC Package .....	200	N/A
Maximum Power Dissipation (Any One Transistor) .....		500mW
Maximum Junction Temperature (Plastic Package) .....		150°C
Maximum Storage Temperature Range .....		-65°C to 150°C
Maximum Lead Temperature (Soldering 10s) .....		300°C
(SOIC - Lead Tips Only)		





TRANSISTORES POTÊNCIA

## TABELA DE ESCOLHA COM OS PRINCIPAIS TIPOS DE TRANSISTORES DE POTÊNCIA

**TABELA 1**

<b>a) invólucro TO-220</b>						
Tipo	Pol.	Ptot (w)	Vceo (V)	Ic (A)	hfe	Vce/Ic (V/A)
2N63286	NPN	65	40	10	1k/20k	A
2N6387	NPN	65	60	10	1k/20k	3/3
BDX33	NPN	70	45	10	740-min	3/5
BDX33A	NPN	70	+60	10	750-min	3/4
BDX33B	NPN	70	80	10	750-min	3/4
BDX33C	NPN	70	100	10	750-min	3/3
BDX34	PNO	70	45	10	750-min	3/3
BDX34A	PNP	70	60	10	750-min	3/4
BDX34B	PNP	70	80	10	750-min	3/4
BDX34C	PNP	70	100	10	750-min	3/3
TIP100	NPN	80	60	15*	200-min	3/3
TIP101	NPN	80	80	15*	200-min	4/8
TIP102	NPN	80	100	15*	200-min	4/8
TIP105	PNP	80	60	15*	1k/20k	4/8
TIP106	PNP	80	80	15*	1k/20k	4/3
TIP107	PNP	80	100	15*	1k/20k	4/3
TIP110	NPN	50	60	4*	1k-min	4/3
TIP111	NPN	50	80	4*	1k-min	4/1
TIP112	NPN	50	100	4*	1k-min	4/1
TIP115	PNP	50	60	4*	1k-min	4/1
TIP116	PNP	50	80	4*	1k-min	4/1
TIP117	PNP	50	100	4*	1k-min	4/1
TIP120	NPN	65	60	8*	1k-min	4/1
TIP121	NPN	65	80	8*	1k-min	3/3
TIP122	NPN	65	100	8*	1k-min	3/3
TIP125	PNP	65	60	8*	1k-min	3/3
TIP126	PNP	65	80	8*	1k-min	3/3
TIP127	PNP	65	100	8*	1k-min	3/3
TIP130	NPN	70	60	12*	1k/15k	3/3
TIP131	NPN	70	80	12*	1k/15k	4/4
TIP132	NPN	70	100	12*	1k/15k	4/4
TIP135	PNP	70	60	12*	1k/15k	4/4
TIP136	PNP	70	80	12*	1k/15k	4/4
TIP137	PNP	70	100	12*	1k/15k	4/4
(*) Corrente de pico - Vceo						
<b>b) Invólucro TO-218</b>						
BU931RP	NPN	150	700	15	300	10/5
TIP140	NPN	125	60	10	1k-min	4/3
TIP141	NPN	125	80	10	1k-min	4/3
TIP142	NPN	125	100	10	1k-min	4/3
TIP145	PNP	125	60	10	1k-min	4/3
TIP146	PNP	125	80	10	1k-min	4/3
TIP147	PNP	125	100	10	1k-min	4/3

BD 135/137/139 NPN 1,5A

Philips Semiconductors

Product specification

**NPN power transistors**

**BD135; BD137; BD139**

**FEATURES**

- High current (max. 1.5 A)
- Low voltage (max. 80 V).

**APPLICATIONS**

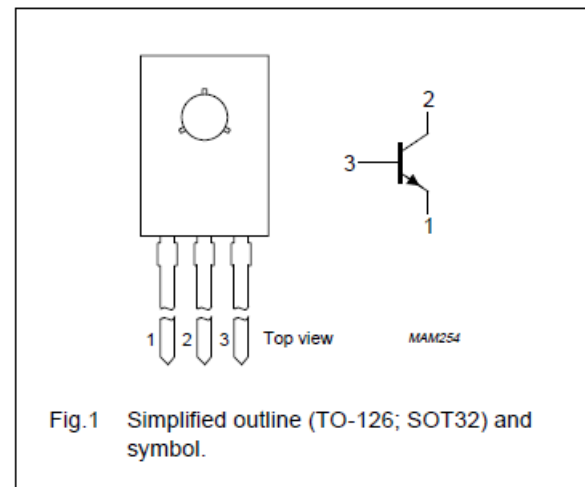
- Driver stages in hi-fi amplifiers and television circuits.

**DESCRIPTION**

NPN power transistor in a TO-126; SOT32 plastic package. PNP complements: BD136, BD138 and BD140.

**PINNING**

PIN	DESCRIPTION
1	emitter
2	collector, connected to metal part of mounting surface
3	base



**LIMITING VALUES**

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V <sub>CBO</sub>	collector-base voltage	open emitter			
	BD135		–	45	V
	BD137		–	60	V
	BD139	–	100	V	
V <sub>CEO</sub>	collector-emitter voltage	open base			
	BD135		–	45	V
	BD137		–	60	V
	BD139	–	80	V	
V <sub>EBO</sub>	emitter-base voltage	open collector	–	5	V
I <sub>C</sub>	collector current (DC)		–	1.5	A
I <sub>CM</sub>	peak collector current		–	2	A
I <sub>BM</sub>	peak base current		–	1	A
P <sub>tot</sub>	total power dissipation	T <sub>mb</sub> ≤ 70 °C	–	8	W
T <sub>stg</sub>	storage temperature		–65	+150	°C
T <sub>j</sub>	junction temperature		–	150	°C
T <sub>amb</sub>	operating ambient temperature		–65	+150	°C

NPN power transistors

BD135; BD137; BD139

**THERMAL CHARACTERISTICS**

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
$R_{th\ j-a}$	thermal resistance from junction to ambient	note 1	100	K/W
$R_{th\ j-mb}$	thermal resistance from junction to mounting base		10	K/W

**Note**

1. Refer to TO-126; SOT32 standard mounting conditions.

**CHARACTERISTICS**

$T_j = 25\text{ }^\circ\text{C}$  unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{CBO}$	collector cut-off current	$I_E = 0; V_{CB} = 30\text{ V}$	–	–	100	nA
		$I_E = 0; V_{CB} = 30\text{ V}; T_j = 125\text{ }^\circ\text{C}$	–	–	10	$\mu\text{A}$
$I_{EBO}$	emitter cut-off current	$I_C = 0; V_{EB} = 5\text{ V}$	–	–	100	nA
$h_{FE}$	DC current gain	$V_{CE} = 2\text{ V}$ ; (see Fig.2) $I_C = 5\text{ mA}$	40	–	–	
		$I_C = 150\text{ mA}$	63	–	250	
		$I_C = 500\text{ mA}$	25	–	–	
	DC current gain BD135-10; BD137-10; BD139-10 BD135-16; BD137-16; BD139-16	$I_C = 150\text{ mA}; V_{CE} = 2\text{ V}$ ; (see Fig.2)	63	–	160	
			100	–	250	
$V_{CEsat}$	collector-emitter saturation voltage	$I_C = 500\text{ mA}; I_B = 50\text{ mA}$	–	–	0.5	V
$V_{BE}$	base-emitter voltage	$I_C = 500\text{ mA}; V_{CE} = 2\text{ V}$	–	–	1	V
$f_T$	transition frequency	$I_C = 50\text{ mA}; V_{CE} = 5\text{ V}$ ; $f = 100\text{ MHz}$	–	190	–	MHZ
$\frac{h_{FE1}}{h_{FE2}}$	DC current gain ratio of the complementary pairs	$ I_C  = 150\text{ mA};  V_{CE}  = 2\text{ V}$	–	1.3	1.6	

BD 136/138/140 PNP 1,5A 8W

**PNP power transistors**

**BD136; BD138; BD140**

**FEATURES**

- High current (max. 1.5 A)
- Low voltage (max. 80 V).

**APPLICATIONS**

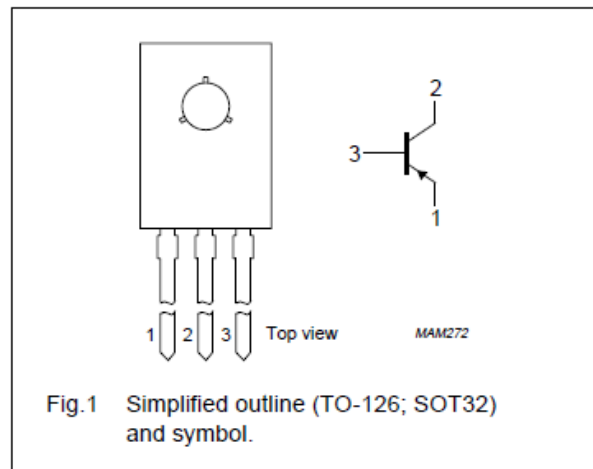
- General purpose power applications, e.g. driver stages in hi-fi amplifiers and television circuits.

**DESCRIPTION**

PNP power transistor in a TO-126; SOT32 plastic package. NPN complements: BD135, BD137 and BD139.

**PINNING**

PIN	DESCRIPTION
1	emitter
2	collector, connected to metal part of mounting surface
3	base



**LIMITING VALUES**

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V <sub>CBO</sub>	collector-base voltage	open emitter			
	BD136		–	–45	V
	BD138		–	–60	V
	BD140		–	–100	V
V <sub>CEO</sub>	collector-emitter voltage	open base			
	BD136		–	–45	V
	BD138		–	–60	V
	BD140		–	–80	V
V <sub>EBO</sub>	emitter-base voltage	open collector	–	–5	V
I <sub>C</sub>	collector current (DC)		–	–1.5	A
I <sub>CM</sub>	peak collector current		–	–2	A
I <sub>BM</sub>	peak base current		–	–1	A
P <sub>tot</sub>	total power dissipation	T <sub>mb</sub> ≤ 70 °C	–	8	W
T <sub>stg</sub>	storage temperature		–65	+150	°C
T <sub>j</sub>	junction temperature		–	150	°C
T <sub>amb</sub>	operating ambient temperature		–65	+150	°C

**THERMAL CHARACTERISTICS**

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
$R_{th\ j-a}$	thermal resistance from junction to ambient	note 1	100	K/W
$R_{th\ j-mb}$	thermal resistance from junction to mounting base		10	K/W

**Note**

1. Refer to TO-126 (SOT32) standard mounting conditions.

**CHARACTERISTICS**

$T_j = 25\text{ }^\circ\text{C}$  unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{CBO}$	collector cut-off current	$I_E = 0; V_{CB} = -30\text{ V}$	–	–	–100	nA
		$I_E = 0; V_{CB} = -30\text{ V}; T_j = 125\text{ }^\circ\text{C}$	–	–	–10	$\mu\text{A}$
$I_{EBO}$	emitter cut-off current	$I_C = 0; V_{EB} = -5\text{ V}$	–	–	–100	nA
$h_{FE}$	DC current gain	$V_{CE} = -2\text{ V};$ (see Fig.2)				
		$I_C = -5\text{ mA}$	40	–	–	
		$I_C = -150\text{ mA}$	63	–	250	
	$I_C = -500\text{ mA}$	25	–	–		
	DC current gain BD136-10; BD138-10; BD140-10 BD136-16; BD138-16; BD140-16	$I_C = -150\text{ mA}; V_{CE} = -2\text{ V};$ (see Fig.2)	63 100	– –	160 250	
$V_{CEsat}$	collector-emitter saturation voltage	$I_C = -500\text{ mA}; I_B = -50\text{ mA}$	–	–	–0.5	V
$V_{BE}$	base-emitter voltage	$I_C = -500\text{ mA}; V_{CE} = -2\text{ V}$	–	–	–1	V
$f_T$	transition frequency	$I_C = -50\text{ mA}; V_{CE} = -5\text{ V};$ $f = 100\text{ MHz}$	–	160	–	MHz
$\frac{h_{FE1}}{h_{FE2}}$	DC current gain ratio of the complementary pairs	$ I_C  = 150\text{ mA};  V_{CE}  = 2\text{ V}$	–	1.3	1.6	



TIP 29 NPN TIP 30 PNP 1A 30 W



**COMPLEMENTARY SILICON PLASTIC POWER TRANSISTORS**

... designed for use in general purpose power amplifier and switching applications.

**FEATURES:**

\* Collector-Emitter Sustaining Voltage -

$V_{CE(sus)}$  = 40V(Min)- TIP29,TIP30  
 60V(Min)- TIP29A,TIP30A  
 80V(Min)- TIP29B,TIP30B  
 100V(Min)-TIP29C,TIP30C

\* Collector-Emitter Saturation Voltage-  $V_{CE(sat)}$  = 0.7V(Max)@  $I_C$  = 1.0 A

\* Current Gain-Bandwidth Product  $f_T$  = 3.0 MHz (Min)@  $I_C$  = 200 mA

<b>NPN</b>	<b>PNP</b>
<b>TIP29</b>	<b>TIP30</b>
<b>TIP29A</b>	<b>TIP30A</b>
<b>TIP29B</b>	<b>TIP30B</b>
<b>TIP29C</b>	<b>TIP30C</b>

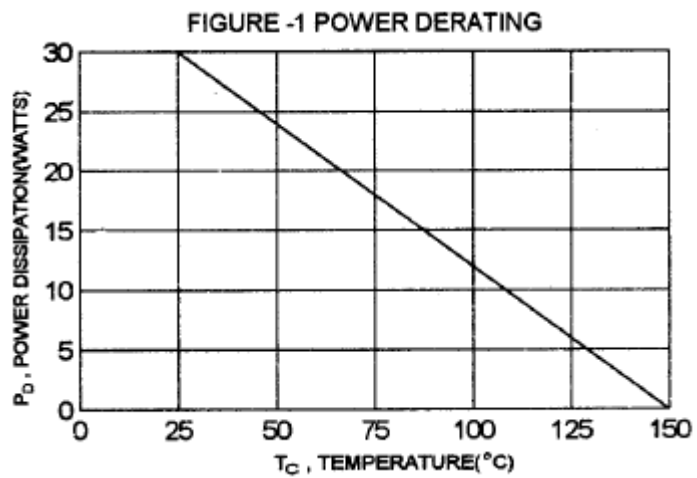
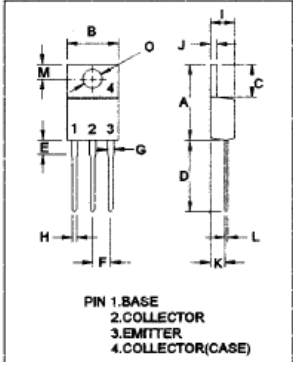
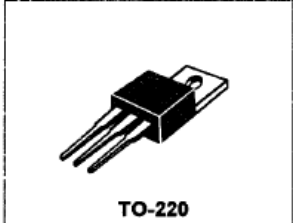
1.0 AMPERE  
 COMPLEMENTARY SILICON  
 POWER TRANSISTORS  
 40-100 VOLTS  
 30 WATTS

**MAXIMUM RATINGS**

Characteristic	Symbol	TIP29 TIP30	TIP29A TIP30A	TIP29B TIP30B	TIP29C TIP30C	Unit
Collector-Emitter Voltage	$V_{CEO}$	40	60	80	100	V
Collector-Base Voltage	$V_{CBO}$	40	60	80	100	V
Emitter-Base Voltage	$V_{EBO}$	5.0				V
Collector Current - Continuous - Peak	$I_C$	1.0 3.0				A
Base Current	$I_B$	0.4				A
Total Power Dissipation@ $T_C = 25^\circ C$ Derate above $25^\circ C$	$P_D$	30 0.24				W W/ $^\circ C$
Operating and Storage Junction Temperature Range	$T_J, T_{STG}$	-65 to +150				$^\circ C$

**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance Junction to Case	$R_{\theta jc}$	4.167	$^\circ C/W$



**TIP29, TIP29A, TIP29B, TIP29C NPN / TIP30, TIP30A, TIP30B, TIP30C PNP**

**ELECTRICAL CHARACTERISTICS (  $T_c = 25^\circ\text{C}$  unless otherwise noted )**

Characteristic	Symbol	Min	Max	Unit
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**OFF CHARACTERISTICS**

Collector-Emitter Sustaining Voltage(1) ( $I_c = 30 \text{ mA}$ , $I_B = 0$ )	TIP29, TIP30 TIP29A, TIP30A TIP29B, TIP30B TIP29C, TIP30C	$V_{CEO(sus)}$	40 60 80 100	V
Collector Cutoff Current ( $V_{CE} = 30 \text{ V}$ , $I_B = 0$ ) ( $V_{CE} = 60 \text{ V}$ , $I_B = 0$ )	TIP29, TIP30, TIP29A, TIP30A ← TIP29B, TIP30B, TIP29C, TIP30C	$I_{CEO}$	0.3 0.3	mA
Collector Cutoff Current ( $V_{CE} = 40 \text{ V}$ , $V_{EB} = 0$ ) ( $V_{CE} = 60 \text{ V}$ , $V_{EB} = 0$ ) ( $V_{CE} = 80 \text{ V}$ , $V_{EB} = 0$ ) ( $V_{CE} = 100 \text{ V}$ , $V_{EB} = 0$ )	TIP29, TIP30 TIP29A, TIP30A TIP29B, TIP30B TIP29C, TIP30C	$I_{CES}$	0.2 0.2 0.2 0.2	mA
Emitter Cutoff Current ( $V_{EB} = 5.0 \text{ V}$ , $I_c = 0$ )		$I_{EBO}$	1.0	mA

**ON CHARACTERISTICS (1)**

DC Current Gain ( $I_C = 0.2 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ ) ( $I_C = 1.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ )	$h_{FE}$	40 15	75	
Collector-Emitter Saturation Voltage ( $I_C = 1.0 \text{ A}$ , $I_B = 125 \text{ mA}$ )	$V_{CE(sat)}$		0.7	V
Base-Emitter On Voltage ( $I_C = 1.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ )	$V_{BE(on)}$		1.3	V

**DYNAMIC CHARACTERISTICS**

Current Gain - Bandwidth Product (2) ( $I_C = 200 \text{ mA}$ , $V_{CE} = 10 \text{ V}$ , $f = 1 \text{ MHz}$ )	$f_T$	3.0		MHz
Small Signal Current Gain ( $I_C = 200 \text{ mA}$ , $V_{CE} = 10 \text{ V}$ , $f = 1 \text{ kHz}$ )	$h_{fe}$	20		

(1) Pulse Test: Pulse width  $\leq 300 \mu\text{s}$ , Duty Cycle  $\leq 2.0 \%$

(2)  $f_T = |h_{fe}| \cdot f_{TEST}$

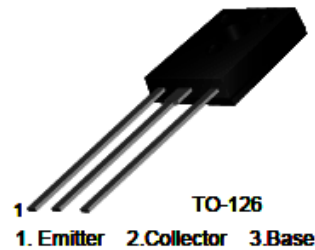
BD233 BD235 BD237 NPN 2A



## BD233/235/237

### Medium Power Linear and Switching Applications

- Complement to BD 234/236/238 respectively



BD233/235/237

### NPN Epitaxial Silicon Transistor

#### Absolute Maximum Ratings $T_C=25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Value	Units	
$V_{CBO}$	Collector-Base Voltage	: BD233	45	V
		: BD235	60	V
		: BD237	100	V
$V_{CEO}$	Collector-Emitter Voltage	: BD233	45	V
		: BD235	60	V
		: BD237	80	V
$V_{CER}$	Collector-Emitter Voltage	: BD233	45	V
		: BD235	60	V
		: BD237	100	V
$V_{EBO}$	Emitter-Base Voltage	5	V	
$I_C$	Collector Current (DC)	2	A	
$I_{CP}$	*Collector Current (Pulse)	6	A	
$P_C$	Collector Dissipation ( $T_C=25^\circ\text{C}$ )	25	W	
$T_J$	Junction Temperature	150	$^\circ\text{C}$	
$T_{STG}$	Storage Temperature	- 65 ~ 150	$^\circ\text{C}$	

**Electrical Characteristics**  $T_C=25^\circ\text{C}$  unless otherwise noted

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Units
$V_{CEO(sus)}$	* Collector-Emitter Sustaining Voltage : BD233 : BD235 : BD237	$I_C = 100\text{mA}, I_B = 0$	45 60 80			V V V
$I_{CBO}$	Collector Cut-off Current : BD233 : BD235 : BD237	$V_{CB} = 45\text{V}, I_E = 0$ $V_{CB} = 60\text{V}, I_E = 0$ $V_{CB} = 100\text{V}, I_E = 0$			100 100 100	$\mu\text{A}$ $\mu\text{A}$ $\mu\text{A}$
$I_{EBO}$	Emitter Cut-off Current	$V_{EB} = 5\text{V}, I_C = 0$			1	mA
$h_{FE}$	* DC Current Gain	$V_{CE} = 2\text{V}, I_C = 150\text{mA}$ $V_{CE} = 2\text{V}, I_C = 1\text{A}$	40 25			
$V_{CE(sat)}$	* Collector-Emitter Saturation Voltage	$I_C = 1\text{A}, I_B = 0.1\text{A}$			0.6	V
$V_{BE(on)}$	* Base-Emitter ON Voltage	$V_{CE} = 2\text{V}, I_C = 1\text{A}$			1.3	V
$f_T$	Current Gain Bandwidth Product	$V_{CE} = 10\text{V}, I_C = 250\text{mA}$	3			MHz

\* Pulse Test: PW=300 $\mu\text{s}$ , duty Cycle=1.5% Pulsed

Typical Characteristics

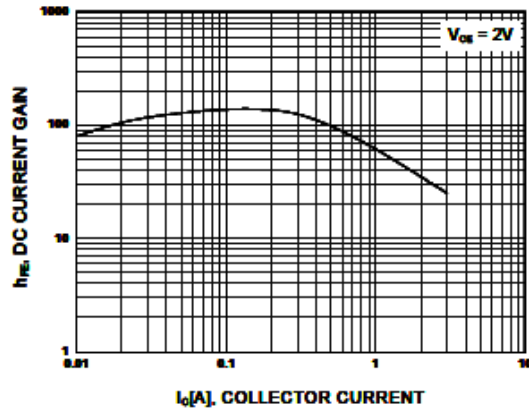


Figure 1. DC current Gain

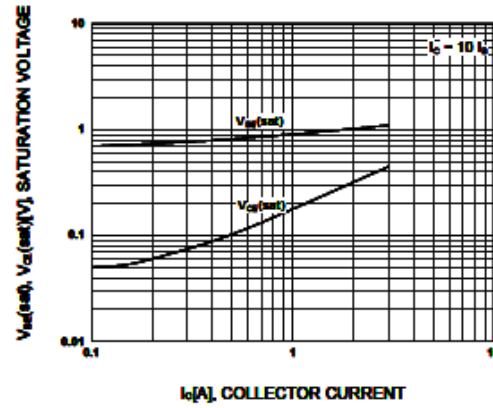


Figure 2. Base-Emitter Saturation Voltage  
Collector-Emitter Saturation Voltage

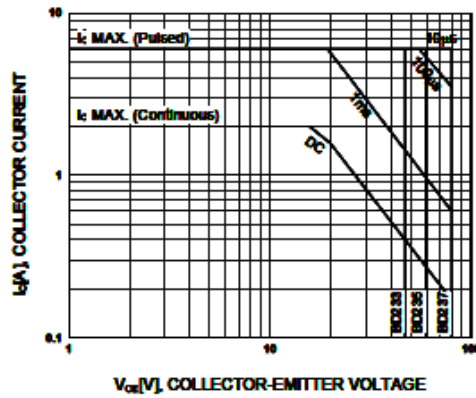


Figure 3. Safe Operating Area

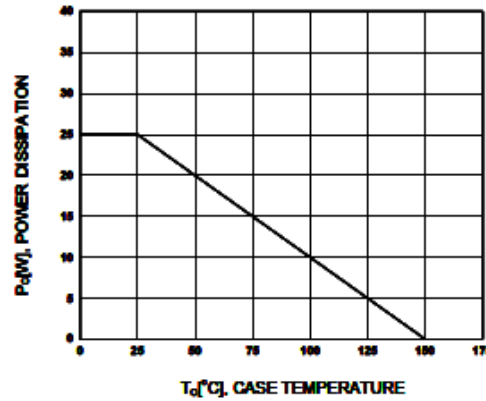


Figure 4. Power Derating

BD233/235/237

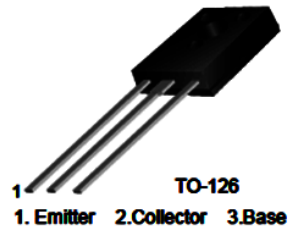
BD234 BD236 BD238 PNP 2A



### BD234/236/238

#### Medium Power Linear and Switching Applications

- Complement to BD 233/235/237 respectively



BD234/236/238

#### PNP Epitaxial Silicon Transistor

#### Absolute Maximum Ratings $T_C=25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Value	Units
$V_{CB0}$	Collector-Base Voltage		
	: BD234	- 45	V
	: BD236	- 60	V
	: BD238	- 100	V
$V_{CE0}$	Collector-Emitter Voltage		
	: BD234	- 45	V
	: BD236	- 60	V
	: BD238	- 80	V
$V_{CER}$	Collector-Emitter Voltage		
	: BD234	- 45	V
	: BD236	- 60	V
	: BD238	- 100	V
$V_{EBO}$	Emitter-Base Voltage	- 5	V
$I_C$	Collector Current (DC)	- 2	A
$I_{CP}$	*Collector Current (Pulse)	- 6	A
$P_C$	Collector Dissipation ( $T_C=25^\circ\text{C}$ )	25	W
$T_J$	Junction Temperature	150	$^\circ\text{C}$
$T_{STG}$	Storage Temperature	- 65 ~ 150	$^\circ\text{C}$

### Typical Characteristics

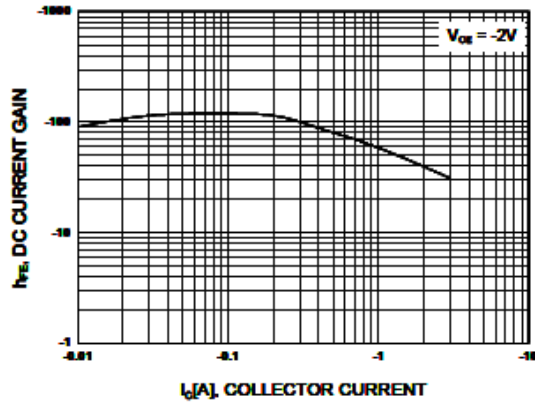


Figure 1. DC current Gain

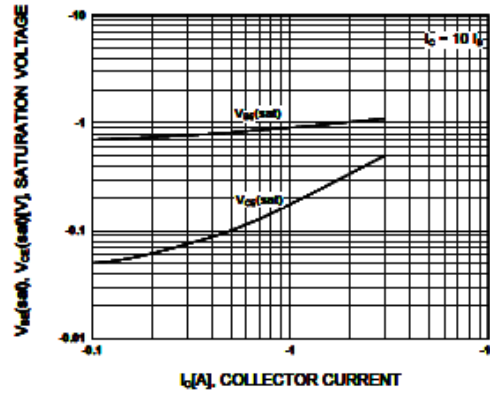


Figure 2. Base-Emitter Saturation Voltage  
Collector-Emitter Saturation Voltage

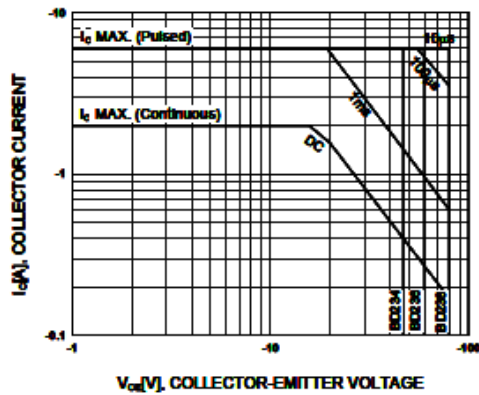


Figure 3. Safe Operating Area

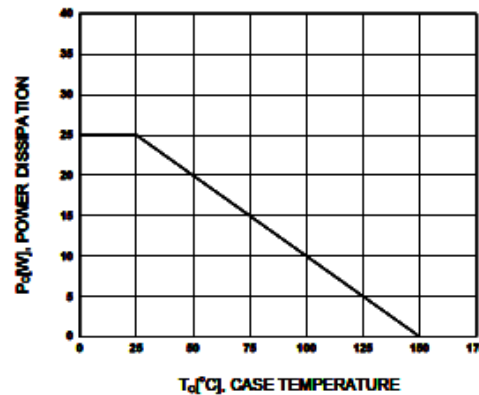


Figure 4. Power Derating

BD234/236/238



TIP31 NPN TIP32 PNP 3A GANHO BAIXO 10 40W



**COMPLEMENTARY SILICON PLASTIC  
POWER TRANSISTORS**

... designed for use in general purpose power amplifier and switching applications.

**FEATURES:**

- \* Collector-Emitter Sustaining Voltage -  $V_{CE(sus)}$  = 40V(Min)- TIP31,TIP32  
60V(Min)- TIP31A,TIP32A  
80V(Min)- TIP31B,TIP32B  
100V(Min)-TIP31C, TIP32C
- \* Collector-Emitter Saturation Voltage-  $V_{CE(sat)}$  = 1.2V(Max)@ $I_C = 3.0 A$
- \* Current Gain-Bandwidth Product  $f_T = 3.0 \text{ MHz (Min)} @ I_C = 500 \text{ mA}$

<b>NPN</b>	<b>PNP</b>
<b>TIP31</b>	<b>TIP32</b>
<b>TIP31A</b>	<b>TIP32A</b>
<b>TIP31B</b>	<b>TIP32B</b>
<b>TIP31C</b>	<b>TIP32C</b>

**3 AMPERE  
COMPLEMENTARY SILICON  
POWER TRANSISTORS  
40 -100 VOLTS  
40 WATTS**

**MAXIMUM RATINGS**

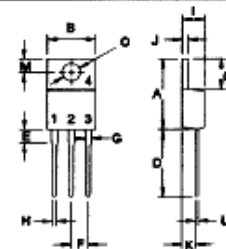
Characteristic	Symbol	TIP31 TIP32	TIP31A TIP32A	TIP31B TIP32B	TIP31C TIP32C	Unit
Collector-Emitter Voltage	$V_{CE0}$	40	60	80	100	V
Collector-Base Voltage	$V_{CB0}$	40	60	80	100	V
Emitter-Base Voltage	$V_{EB0}$	5.0				V
Collector Current - Continuous - Peak	$I_C$	3.0 5.0				A
Base Current	$I_B$	1.0				A
Total Power Dissipation@ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	40 0.32				W W/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{sto}$	-65 to +150				$^\circ\text{C}$

**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance Junction to Case	$R_{\theta jc}$	3.125	$^\circ\text{C/W}$

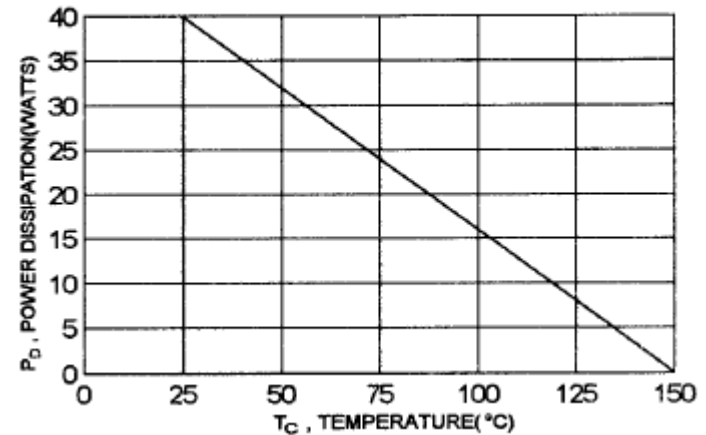


**TO-220**



PIN 1.BASE  
2.COLLECTOR  
3.EMITTER  
4.COLLECTOR(CASE)

**FIGURE -1 POWER DERATING**



TIP31, TIP31A, TIP31B, TIP31C NPN / TIP32, TIP32A, TIP32B, TIP32C PNP

**ELECTRICAL CHARACTERISTICS** (  $T_c = 25^\circ\text{C}$  unless otherwise noted )

Characteristic	Symbol	Min	Max	Unit
----------------	--------	-----	-----	------

**OFF CHARACTERISTICS**

Collector-Emitter Sustaining Voltage(1) ( $I_c = 30 \text{ mA}$ , $I_B = 0$ )	TIP31, TIP32 TIP31A, TIP32A TIP31B, TIP32B TIP31C, TIP32C	$V_{CEO(sus)}$	40 60 80 100	V
Collector Cutoff Current ( $V_{CE} = 30 \text{ V}$ , $I_B = 0$ ) ( $V_{CE} = 60 \text{ V}$ , $I_B = 0$ )	TIP31, TIP32, TIP31A, TIP32A TIP31B, TIP32B, TIP31C, TIP32C	$I_{CEO}$		0.3 0.3 mA
Collector Cutoff Current ( $V_{CE} = 40 \text{ V}$ , $V_{EB} = 0$ ) ( $V_{CE} = 60 \text{ V}$ , $V_{EB} = 0$ ) ( $V_{CE} = 80 \text{ V}$ , $V_{EB} = 0$ ) ( $V_{CE} = 100 \text{ V}$ , $V_{EB} = 0$ )	TIP31, TIP32 TIP31A, TIP32A TIP31B, TIP32B TIP31C, TIP32C	$I_{CES}$		0.2 0.2 0.2 0.2 mA
Emitter Cutoff Current ( $V_{EB} = 5.0 \text{ V}$ , $I_C = 0$ )		$I_{EBO}$		1.0 mA

**ON CHARACTERISTICS (1)**

DC Current Gain ( $I_C = 1.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ ) ( $I_C = 3.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ )	$h_{FE}$	25 10	50	
Collector-Emitter Saturation Voltage ( $I_C = 3.0 \text{ A}$ , $I_B = 375 \text{ mA}$ )	$V_{CE(sat)}$		1.2	V
Base-Emitter On Voltage ( $I_C = 3.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ )	$V_{BE(on)}$		1.8	V

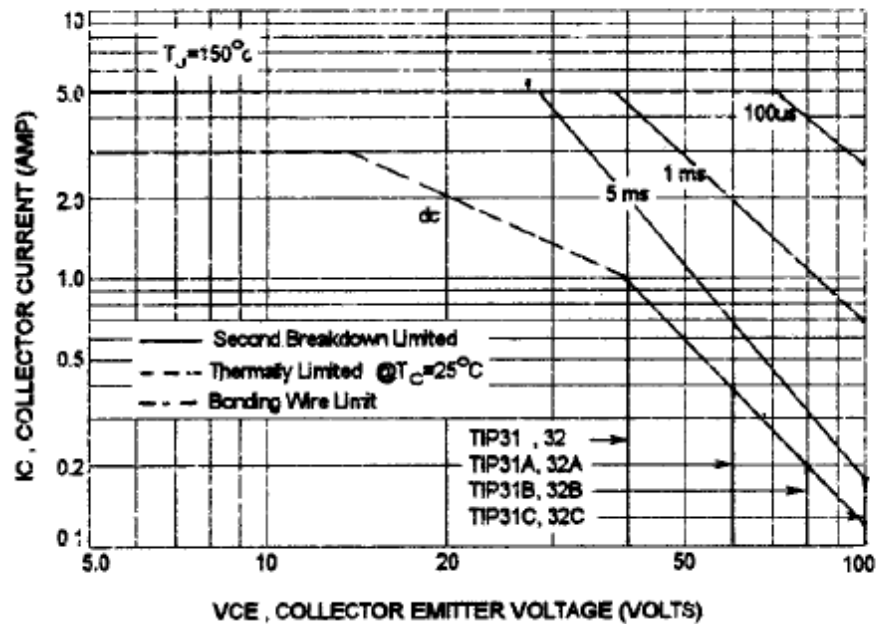
**DYNAMIC CHARACTERISTICS**

Current Gain - Bandwidth Product (2) ( $I_C = 500 \text{ mA}$ , $V_{CE} = 10 \text{ V}$ , $f_{TEST} = 1 \text{ MHz}$ )	$f_T$	3.0		MHz
Small Signal Current Gain ( $I_C = 500 \text{ mA}$ , $V_{CE} = 10 \text{ V}$ , $f = 1 \text{ kHz}$ )	$h_{ie}$	20		

(1) Pulse Test: Pulse width  $\leq 300 \mu\text{s}$ , Duty Cycle  $\leq 2.0 \%$

(2)  $f_T = |h_{ie}| \cdot f_{TEST}$

FIG-6 ACTIVE REGION SAFE OPERATING AREA



There are two limitation on the power handling ability of a transistor: average junction temperature and second breakdown safe operating area curves indicate  $I_C$ - $V_{CE}$  limits of the transistor that must be observed for reliable operation i.e., the transistor must not be subjected to greater dissipation than curves indicate.

The data of FIG-6 curve is base on  $T_{J(PK)} = 150^{\circ}\text{C}$ ;  $T_C$  is variable depending on power level. second breakdown pulse limits are valid for duty cycles to 10% provided  $T_{J(PK)} \leq 150^{\circ}\text{C}$ . At high case temperatures, thermal limitation will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

TIP41 NPN /TIP42 PNP 6A 65W



## TIP41 (NPN) Series TIP42 (PNP) Series



Nell High Power Products

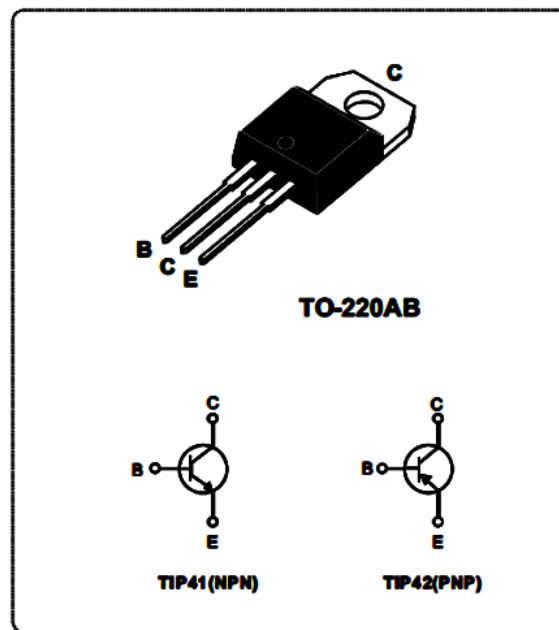
### Complementary Silicon Power Transistor 6A/40~100V/65W

#### FEATURES

- Complementary NPN-PNP transistors
- Low collector-emitter saturation voltage
- Satisfactory linearity of forward current transfer ratio  $h_{FE}$
- TO-220AB package which can be installed to the heat sink with one screw
- Collector - Emitter Saturation Voltage:  
 $V_{CE(sat)} = 1.5V_{dc}$  (MAX.) @  $I_C = 6A$
- Collector - Emitter Saturation Voltage:  
 $V_{CEO(sus)} = 40V_{dc}$  (Min.) - TIP41, TIP42  
 $= 60V_{dc}$  (Min.) - TIP41A, TIP42A  
 $= 80V_{dc}$  (Min.) - TIP41B, TIP42B  
 $= 100V_{dc}$  (Min.) - TIP41C, TIP42C
- DC Current Gain  $h_{FE} = 30$  (Min.) @  $I_C = 0.3A$
- High Current Gain - Bandwidth product  
 $f_T = 3.0$  MHz (Min.) @  $I_C = 0.5A$

#### APPLICATIONS

- Audio amplifier
- General purpose switching and amplifier



<b>ABSOLUTE MAXIMUM RATINGS (T<sub>C</sub> = 25°C)</b>							
SYMBOL	PARAMETER		VALUE				UNIT
			TIP41 TIP42	TIP41A TIP42A	TIP41B TIP42B	TIP41C TIP42C	
V <sub>CB0</sub>	Collector to base voltage (I <sub>E</sub> = 0)		40	60	80	100	V
V <sub>CE0</sub>	Collector to emitter voltage (I <sub>B</sub> = 0)		40	60	80	100	
V <sub>EB0</sub>	Emitter to base voltage (I <sub>C</sub> = 0)		5				
I <sub>C</sub>	Collector current		6				A
I <sub>CM</sub>	Collector peak current (t <sub>p</sub> < 0.3mS)		10				
I <sub>B</sub>	Base current		2				
P <sub>C</sub>	Collector power dissipation (Derate above 25°C)		@T <sub>C</sub> = 25°C	65 (0.52)			W(W°C)
			@T <sub>A</sub> = 25°C	2.0 (0.016)			
T <sub>J</sub>	Junction temperature		150				°C
T <sub>stg</sub>	Storage temperature		-65 to 150				
E	Unclamped inductive load energy (Note 1)		62.5				mJ

Note: 1. This rating is based on the capability of the transistor to operate safely in a circuit of:

I<sub>C</sub> = 2.5A, L = 20mH, R<sub>BE</sub> = 100Ω, P.R.F. = 10 Hz, V<sub>CC</sub> = 10V

<b>THERMAL CHARACTERISTICS (T<sub>C</sub> = 25°C)</b>			
SYMBOL	PARAMETER	VALUE	UNIT
R <sub>th(j-c)</sub>	Maximum thermal resistance, junction to case	1.67	°C/W
R <sub>th(j-a)</sub>	Maximum thermal resistance, junction to ambient	57	°C/W

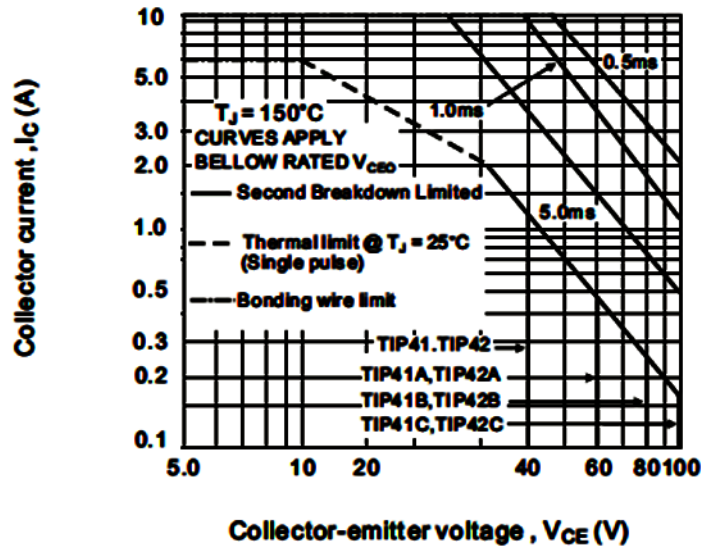
ELECTRICAL CHARACTERISTICS (T <sub>C</sub> = 25°C unless otherwise specified)					
SYMBOL	PARAMETER	CONDITIONS	MIN	MAX	UNIT
<b>Off Characteristics</b>					
V <sub>CE(sus)</sub>	Collector to emitter sustaining voltage (Note 1)	I <sub>C</sub> = 30mA, I <sub>B</sub> = 0	TIP41, TIP42	40	V
			TIP41A, TIP42A	60	
			TIP41B, TIP42B	80	
			TIP41C, TIP42C	100	
I <sub>CEO</sub>	Collector cutoff current	V <sub>CE</sub> = 30V, I <sub>B</sub> = 0	TIP41, TIP42 TIP41A, TIP42A	0.7	mA
		V <sub>CE</sub> = 60V, I <sub>B</sub> = 0	TIP41B, TIP42B TIP41C, TIP42C		
I <sub>EBO</sub>	Emitter cutoff current	V <sub>EB</sub> = 5V, I <sub>C</sub> = 0		1.0	
I <sub>CES</sub>	Collector cutoff current	V <sub>CE</sub> = 40V, V <sub>EB</sub> = 0	TIP41, TIP42	400	μA
		V <sub>CE</sub> = 60V, V <sub>EB</sub> = 0	TIP41A, TIP42A	400	
		V <sub>CE</sub> = 80V, V <sub>EB</sub> = 0	TIP41B, TIP42B	400	
		V <sub>CE</sub> = 100V, V <sub>EB</sub> = 0	TIP41C, TIP42C	400	
<b>On Characteristics</b>					
h <sub>FE</sub>	Forward current transfer ratio (DC current gain)	V <sub>CE</sub> = 4V, I <sub>C</sub> = 0.3A	30		
		V <sub>CE</sub> = 4V, I <sub>C</sub> = 3A	15	75	
V <sub>CE(sat)</sub>	Collector to emitter saturation voltage (Note 1)	I <sub>C</sub> = 6A, I <sub>B</sub> = 0.6A		1.5	V
V <sub>BE(on)</sub>	Base to emitter voltage (Note 1)	I <sub>C</sub> = 6A, V <sub>CE</sub> = 4V		2.0	
<b>Dynamic Characteristics</b>					
f <sub>T</sub>	Current gain - Bandwidth product (note 2)	I <sub>C</sub> = 0.5A, V <sub>CE</sub> = 10V, f <sub>test</sub> = 1MHz	3.0		MHz
h <sub>fe</sub>	Small signal current gain	I <sub>C</sub> = 0.5A, V <sub>CE</sub> = 10V, f = 1KHz	20		

Note 1. Pulsed : Pulse duration ≤ 300 μs, duty cycle ≤ 2.0%.

Note 2. f<sub>T</sub> = |h<sub>fe</sub>| · f<sub>TEST</sub>

Note 3. For PNP type voltage and current are negative.

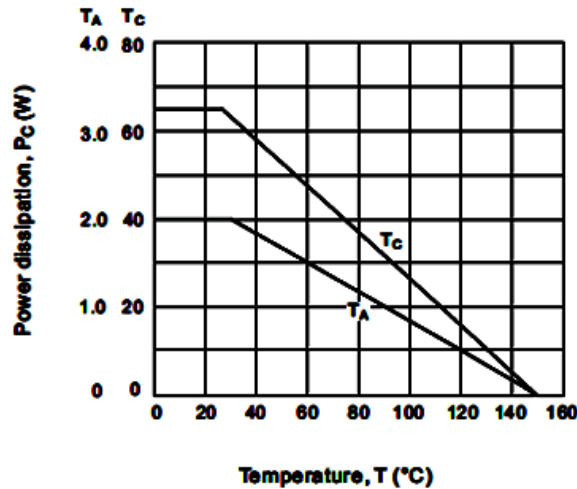
**Fig.3 Active region safe operating area**



There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate  $I_C$ - $V_{CE}$  limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curve indicate.

The data of fig.3 is based on  $T_{J(pk)} = 150^\circ\text{C}$ ;  $T_C$  is variable depending on conditions. Second breakdown pulse limits are valid for duty cycles to 10% provided  $T_{J(pk)} \leq 150^\circ\text{C}$ .  $T_{J(pk)}$  may be calculated from the data in Figure 13. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

**Fig.1 Power Derating**





TIP3055 15A 90W



COMPLEMENTARY SILICON POWER TRANSISTORS

...designed for use in general-purpose amplifier and switching applications

FEATURES:

- \* Power Dissipation -  $P_D = 90W @ T_C = 25^\circ C$
- \* DC Current Gain  $hFE = 20 \sim 100 @ I_C = 4.0 A$
- \*  $V_{CE(sat)} = 1.1 V (Max.) @ I_C = 4.0 A, I_B = 400 mA$

MAXIMUM RATINGS

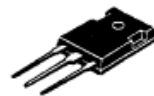
Characteristic	Symbol	Rating	Unit
Collector-Emitter Voltage	$V_{CEO}$	60	V
Collector-Emitter Voltage	$V_{CER}$	70	V
Collector-Base Voltage	$V_{CBO}$	100	V
Emitter-Base Voltage	$V_{EBO}$	7.0	V
Collector Current-Continuous	$I_C$	15	A
Base Current	$I_B$	7.0	A
Total Power Dissipation @ $T_C=25^\circ C$ Derate above $25^\circ C$	$P_D$	90 0.72	W W/ $^\circ C$
Operating and Storage Junction Temperature Range	$T_J, T_{STG}$	- 65 to +150	$^\circ C$

THERMAL CHARACTERISTICS

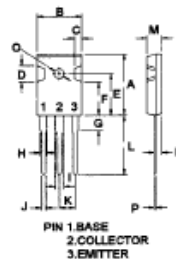
Characteristic	Symbol	Max	Unit
Thermal Resistance Junction to Case	$R_{\theta jc}$	1.39	$^\circ C/W$

NPN PNP  
TIP3055 TIP2955

15 AMPERE  
COMPLEMENTARY SILICON  
POWER TRANSISTORS  
60 VOLTS  
90 WATTS



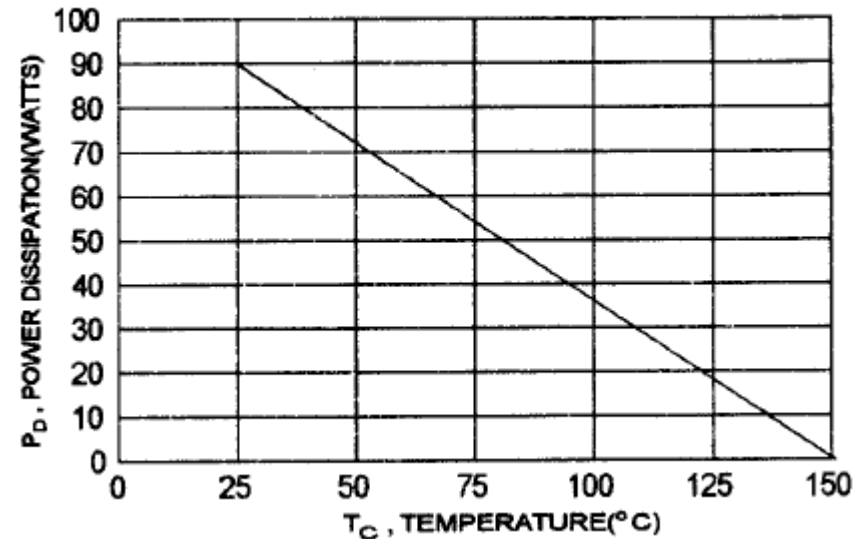
TO-247(3P)



PIN 1 BASE  
2. COLLECTOR  
3. EMITTER

DIM	MILLIMETERS
.....	.....

FIGURE -1 POWER DERATING



**TIP3055 NPN / TIP2955 PNP****ELECTRICAL CHARACTERISTICS (  $T_c = 25^\circ\text{C}$  unless otherwise noted )**

Characteristic	Symbol	Min	Max	Unit
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**OFF CHARACTERISTICS**

Collector - Emitter Sustaining Voltage (1) ( $I_C = 30 \text{ mA}$ , $I_B = 0$ )	$V_{CEO(SUS)}$	60		V
Collector Cutoff Current ( $V_{CE} = 70 \text{ V}$ , $R_{BE} = 100 \text{ ohm}$ )	$I_{CER}$		1.0	mA
Collector Cutoff Current ( $V_{CE} = 30 \text{ V}$ , $I_B = 0$ )	$I_{CEO}$		0.7	mA
Collector Cutoff Current ( $V_{CE} = 100 \text{ V}$ , $V_{BE(off)} = 1.5 \text{ V}$ )	$I_{CEV}$		5.0	mA
Emitter Cutoff Current ( $V_{EB} = 7.0 \text{ V}$ , $I_C = 0$ )	$I_{EBO}$		5.0	mA

**ON CHARACTERISTICS (1)**

DC Current Gain ( $I_C = 4.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ ) ( $I_C = 10 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ )	$h_{FE}$	20 5.0	100	
Collector - Emitter Saturation Voltage ( $I_C = 4.0 \text{ A}$ , $I_B = 0.4 \text{ A}$ ) ( $I_C = 10 \text{ A}$ , $I_B = 3.3 \text{ A}$ )	$V_{CE(sat)}$		1.1 3.0	V
Base - Emitter On Voltage ( $I_C = 4.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ )	$V_{BE(on)}$		1.8	V

**DYNAMIC CHARACTERISTICS**

Current Gain - Bandwidth Product ( $I_C = 500 \text{ mA}$ , $V_{CE} = 10 \text{ V}$ , $f = 1.0 \text{ MHz}$ )	$f_T$	2.5		MHz
Small-Signal Current Gain ( $I_C = 1.0 \text{ A}$ , $V_{CE} = 4.0 \text{ V}$ , $f = 1 \text{ KHz}$ )	$h_{FE}$	15		

(1) Pulse Test: Pulse width =  $300 \mu\text{s}$ , Duty Cycle  $\leq 2.0\%$

(2)  $f_T = |h_{fe}| \cdot f_{test}$

2N3055 TO-3 METÁLICO 15A 115W

**MOTOROLA**  
SEMICONDUCTOR TECHNICAL DATA

Order this document  
by 2N3055/D

### Complementary Silicon Power Transistors

... designed for general-purpose switching and amplifier applications.

- DC Current Gain —  $h_{FE} = 20-70 @ I_C = 4 \text{ Adc}$
- Collector-Emitter Saturation Voltage —  
 $V_{CE(sat)} = 1.1 \text{ Vdc (Max) @ } I_C = 4 \text{ Adc}$
- Excellent Safe Operating Area

**MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Collector-Emitter Voltage	$V_{CEO}$	60	Vdc
Collector-Emitter Voltage	$V_{CER}$	70	Vdc
Collector-Base Voltage	$V_{CB}$	100	Vdc
Emitter-Base Voltage	$V_{EB}$	7	Vdc
Collector Current — Continuous	$I_C$	15	A <sub>dc</sub>
Base Current	$I_B$	7	A <sub>dc</sub>
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	115 0.657	Watts W/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-65 to +200	$^\circ\text{C}$

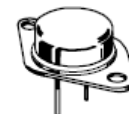
**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	1.52	$^\circ\text{C/W}$

**NPN**  
**2N3055 \***  
**PNP**  
**MJ2955 \***

\*Motorola Preferred Device

**15 AMPERE**  
**POWER TRANSISTORS**  
**COMPLEMENTARY**  
**SILICON**  
**60 VOLTS**  
**115 WATTS**



CASE 1-07  
TO-204AA  
(TO-3)

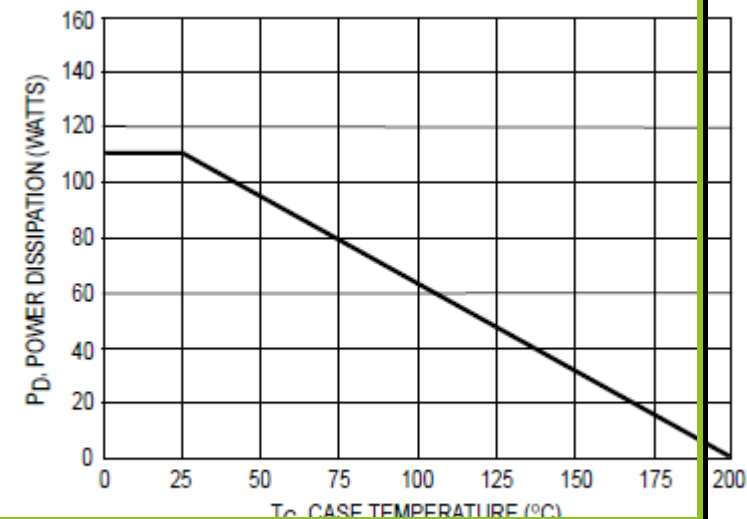


Figure 1. Power Derating

**2N3055 MJ2955****ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
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**\*OFF CHARACTERISTICS**

Collector–Emitter Sustaining Voltage (1) ( $I_C = 200 \text{ mAdc}$ , $I_B = 0$ )	$V_{CEO(sus)}$	60	—	Vdc
Collector–Emitter Sustaining Voltage (1) ( $I_C = 200 \text{ mAdc}$ , $R_{BE} = 100 \text{ Ohms}$ )	$V_{CER(sus)}$	70	—	Vdc
Collector Cutoff Current ( $V_{CE} = 30 \text{ Vdc}$ , $I_B = 0$ )	$I_{CEO}$	—	0.7	mAdc
Collector Cutoff Current ( $V_{CE} = 100 \text{ Vdc}$ , $V_{BE(off)} = 1.5 \text{ Vdc}$ ) ( $V_{CE} = 100 \text{ Vdc}$ , $V_{BE(off)} = 1.5 \text{ Vdc}$ , $T_C = 150^\circ\text{C}$ )	$I_{CEX}$	— —	1.0 5.0	mAdc
Emitter Cutoff Current ( $V_{BE} = 7.0 \text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	—	5.0	mAdc

**\*ON CHARACTERISTICS (1)**

DC Current Gain ( $I_C = 4.0 \text{ Adc}$ , $V_{CE} = 4.0 \text{ Vdc}$ ) ( $I_C = 10 \text{ Adc}$ , $V_{CE} = 4.0 \text{ Vdc}$ )	$h_{FE}$	20 5.0	70 —	—
Collector–Emitter Saturation Voltage ( $I_C = 4.0 \text{ Adc}$ , $I_B = 400 \text{ mAdc}$ ) ( $I_C = 10 \text{ Adc}$ , $I_B = 3.3 \text{ Adc}$ )	$V_{CE(sat)}$	—	1.1 3.0	Vdc
Base–Emitter On Voltage ( $I_C = 4.0 \text{ Adc}$ , $V_{CE} = 4.0 \text{ Vdc}$ )	$V_{BE(on)}$	—	1.5	Vdc

**SECOND BREAKDOWN**

Second Breakdown Collector Current with Base Forward Biased ( $V_{CE} = 40 \text{ Vdc}$ , $t = 1.0 \text{ s}$ , Nonrepetitive)	$I_{s/b}$	2.87	—	Adc
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**DYNAMIC CHARACTERISTICS**

Current Gain — Bandwidth Product ( $I_C = 0.5 \text{ Adc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ MHz}$ )	$f_T$	2.5	—	MHz
*Small–Signal Current Gain ( $I_C = 1.0 \text{ Adc}$ , $V_{CE} = 4.0 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	$h_{fe}$	15	120	—
*Small–Signal Current Gain Cutoff Frequency ( $V_{CE} = 4.0 \text{ Vdc}$ , $I_C = 1.0 \text{ Adc}$ , $f = 1.0 \text{ kHz}$ )	$f_{hfe}$	10	—	kHz

\* Indicates Within JEDEC Registration. (2N3055)

(1) Pulse Test: Pulse Width  $\leq 300 \mu\text{s}$ , Duty Cycle  $\leq 2.0\%$ .

2SB1184 2SB1243 PNP 3A 50V

# Power Transistor (50V, 3A)

## 2SD1760 / 2SD1864 / 2SD1762

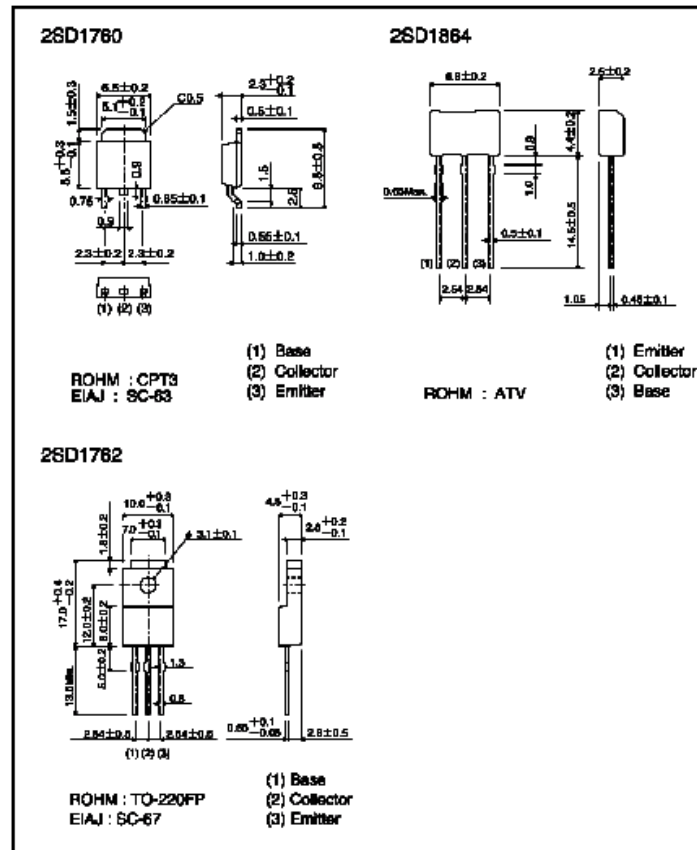
●Features

- 1) Low  $V_{CE(sat)}$ ,  
 $V_{CE(sat)} = 0.5V$  (Typ.)  
 $(I_C/I_B = 2A/0.2A)$
- 2) Complements the  
 2SB1184 / 2SB1243 / 2SB1185.

●Structure

Epitaxial planar type  
 NPN silicon transistor

●External dimensions (Units: mm)



**Transistors****2SD1760 / 2SD1864 / 2SD1762****●Absolute maximum ratings (Ta = 25°C)**

Parameter	Symbol	Limits	Unit
Collector-base voltage	V <sub>CB0</sub>	60	V
Collector-emitter voltage	V <sub>CE0</sub>	50	V
Emitter-base voltage	V <sub>EB0</sub>	5	V
Collector current	I <sub>C</sub>	3	A (DC)
		4.5	A (Pulse) *1
Collector power dissipation	2SD1760	15	W(Tc=25°C) *2
	2SD1864	1	W
	2SD1762	25	W(Tc=25°C) *2
Junction temperature	T <sub>J</sub>	150	°C
Storage temperature	T <sub>stg</sub>	-55~+150	°C

\*1 Single pulse, P<sub>w</sub>=10ms\*2 Printed circuit board, 1.7mm thick, collector copper plating 100mm<sup>2</sup> or larger.



●Electrical characteristics (Ta = 25°C)

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Collector-base breakdown voltage	BV <sub>CSO</sub>	60	—	—	V	I <sub>C</sub> =50 μA
Collector-emitter breakdown voltage	BV <sub>CEO</sub>	50	—	—	V	I <sub>C</sub> =1mA
Emitter-base breakdown voltage	BV <sub>ES0</sub>	5	—	—	V	I <sub>E</sub> =50 μA
Collector cutoff current	I <sub>CS0</sub>	—	—	1	μA	V <sub>CS</sub> =40V
Emitter cutoff current	I <sub>ES0</sub>	—	—	1	μA	V <sub>EB</sub> =4V
Collector-emitter saturation voltage	V <sub>CE(sat)</sub>	—	0.5	1	V	I <sub>C</sub> /I <sub>B</sub> =2A/0.2A
DC current transfer ratio	h <sub>FE</sub>	82	—	390	—	V <sub>CE</sub> =3V, I <sub>C</sub> =0.5A
		60	—	320		
Transition frequency	f <sub>r</sub>	—	90	—	MHz	V <sub>CE</sub> =5V, I <sub>E</sub> =-500mA, f=30MHz
Output capacitance	C <sub>ob</sub>	—	40	—	pF	V <sub>CS</sub> =10V, I <sub>E</sub> =0A, f=1MHz

\* Measured using pulse current.

●Packaging specifications and h<sub>FE</sub>

Type	h <sub>FE</sub>	Package	Taping		Bulk
		Code	TL	TV2	—
		Basic ordering unit (pieces)	2500	2500	200
2SD1760	PQR	○	—	—	
2SD1884	PQR	—	○	—	
2SD1762	DEF	—	—	○	

h<sub>FE</sub> values are classified as follows :

Item	P	Q	R
h <sub>FE</sub>	82~180	120~270	180~390

Item	D	E	F
h <sub>FE</sub>	60~120	100~200	160~320

Transistors

2SD1760 / 2SD1864 / 2SD1762

●Electrical characteristic curves

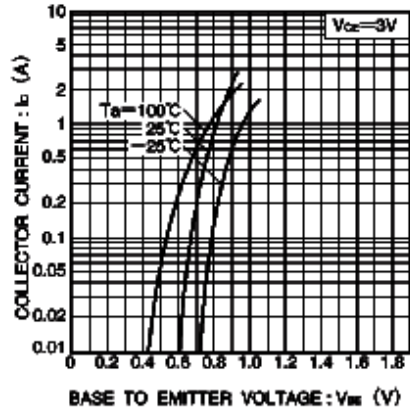


Fig.1 Grounded emitter propagation characteristics

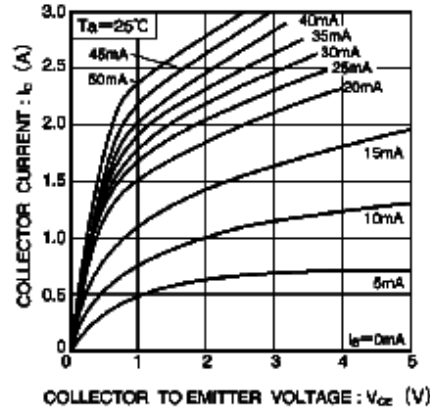


Fig.2 Grounded emitter output characteristics (I)

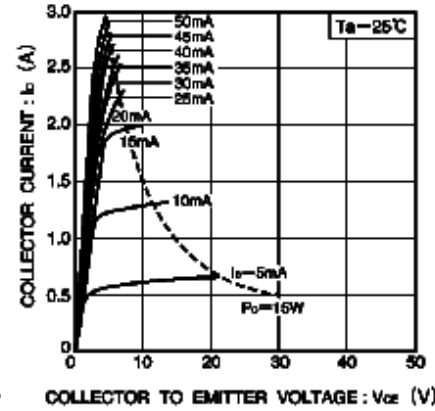


Fig.3 Grounded-emitter output characteristics (II)

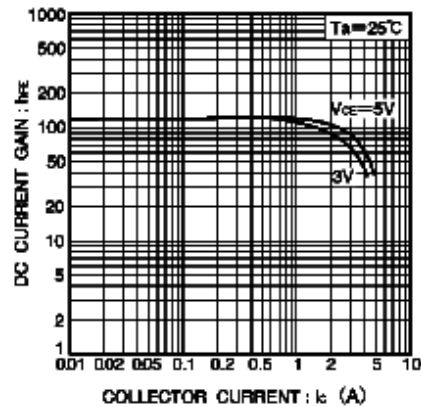


Fig.4 DC current gain vs. collector current (I)

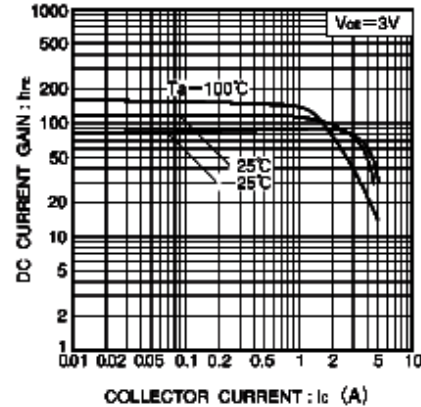


Fig.5 DC current gain vs. collector current (II)

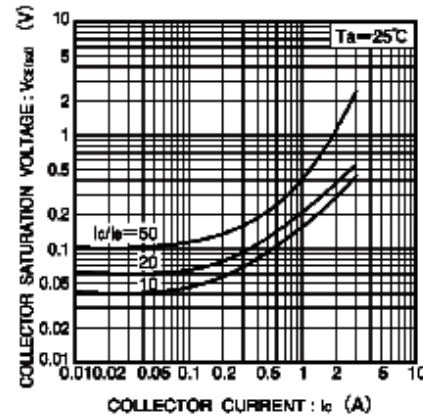


Fig.6 Collector-emitter saturation voltage vs. collector current

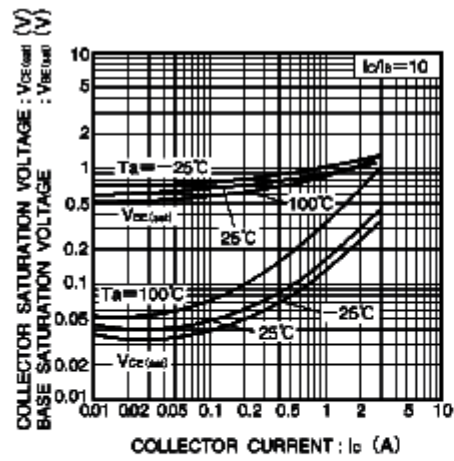


Fig.7 Collector-emitter saturation voltage vs. collector current  
Base-emitter saturation voltage vs. collector current

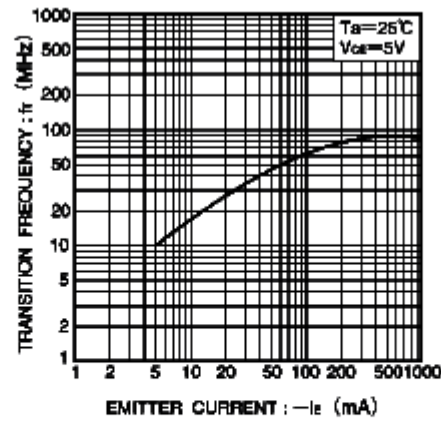


Fig.8 Gain bandwidth product vs. emitter current

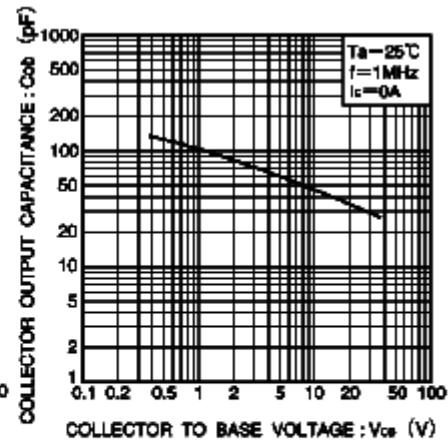


Fig.9 Collector output capacitance vs. collector-base voltage

Transistors

2SD1760 / 2SD1864 / 2SD1762

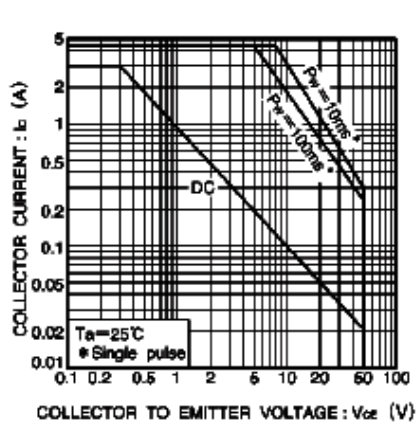


Fig.10 Safe operating area (2SD1760)

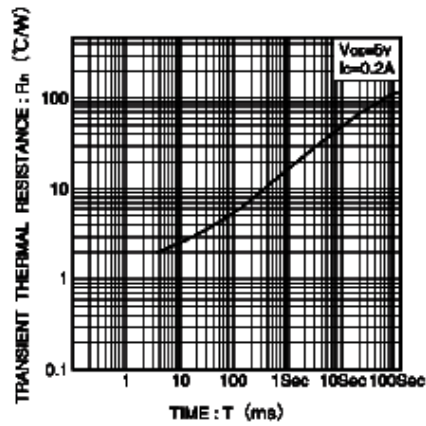


Fig.11 Transient thermal resistance (2SD1760)

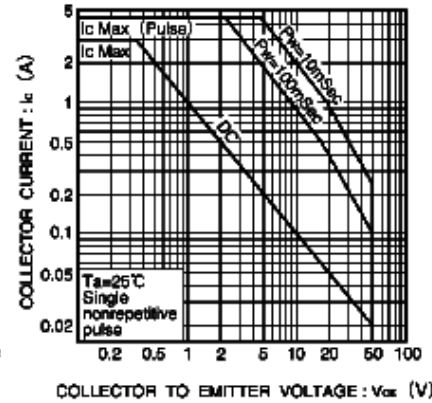


Fig.12 Safe operating area (2SD1864)

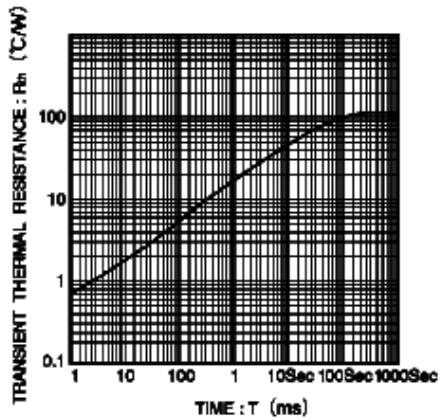


Fig.13 Transient thermal resistance (2SD1864)

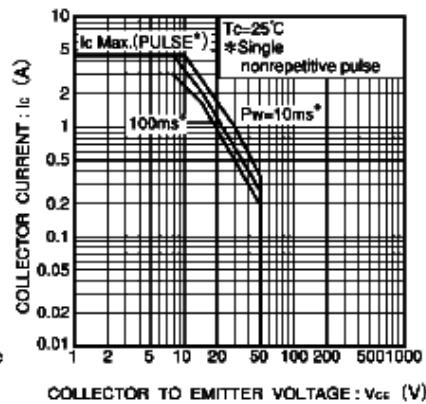


Fig.14 Safe operating area (2SD1762)

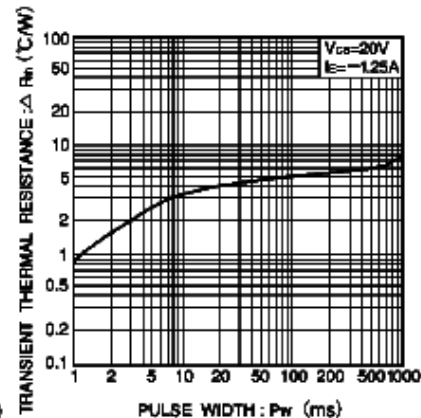


Fig.15 Transient thermal resistance (2SD1762)

2SB1184 2SB1243 PNP 3A 50V

2SB1184 / 2SB1243

Transistors

# Power Transistor (-60V, -3A)

## 2SB1184 / 2SB1243

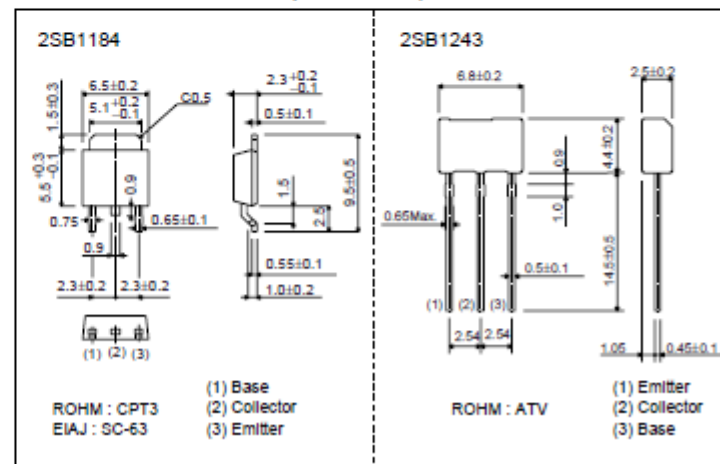
●Features

- 1) LOW  $V_{CE(sat)}$ .  
 $V_{CE(sat)} = -0.5V$  (Typ.)  
 $(I_c/I_B = -2A / -0.2A)$
- 2) Complements the 2SD1760 / 2SD1864.

●Structure

Epitaxial planar type  
 PNP silicon transistor

●External dimensions (Units : mm)



● **Absolute maximum ratings** ( $T_a=25^\circ\text{C}$ )

Parameter	Symbol	Limits	Unit
Collector-base voltage	$V_{CB0}$	-60	V
Collector-emitter voltage	$V_{CE0}$	-50	V
Emitter-base voltage	$V_{EB0}$	-5	V
Collector current	$I_c$	-3	A (DC)
	$I_{cP}$	-4.5	A (Pulse) *1
Collector power dissipation	2SB1184	1	W
		15	W ( $T_c=25^\circ\text{C}$ )
	2SB1243	1	W *2
Junction temperature	$T_j$	150	$^\circ\text{C}$
Storage temperature	$T_{stg}$	-55~+150	$^\circ\text{C}$

\*1 Single pulse,  $P_w=100\text{ms}$

\*2 Printed circuit board, 1.7mm thick, collector copper plating  $100\text{mm}^2$  or larger.

## 2SB1184 / 2SB1243

## Transistors

## ●Electrical characteristics (Ta=25°C)

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Collector-base breakdown voltage	$BV_{CBO}$	-60	-	-	V	$I_C = -50\mu A$
Collector-emitter breakdown voltage	$BV_{CEO}$	-50	-	-	V	$I_C = -1mA$
Emitter-base breakdown voltage	$BV_{EBO}$	-5	-	-	V	$I_E = -50\mu A$
Collector cutoff current	$I_{CBO}$	-	-	-1	$\mu A$	$V_{CB} = -40V$
Emitter cutoff current	$I_{EBO}$	-	-	-1	$\mu A$	$V_{EB} = -4V$
Collector-emitter saturation voltage	$V_{CE(sat)}$	-	-	-1	V	$I_C/I_B = -2A/-0.2A$ *
Base-emitter saturation voltage	$V_{BE(sat)}$	-	-	-1.2	V	$I_C/I_B = -1.5A/-0.15A$ *
DC current transfer ratio	$h_{FE}$	82	-	390	-	$V_{CE} = -3V, I_C = -0.5A$ *
Transition frequency	$f_T$	-	70	-	MHz	$V_{CE} = -5V, I_E = -0.5A, f = 30MHz$
Output capacitance	$C_{ob}$	-	50	-	pF	$V_{CB} = -10V, I_E = 0A, f = 1MHz$

\* Measured using pulse current.

●Packaging specifications and  $h_{FE}$ 

Type	$h_{FE}$	Package	Taping	
		Code	TL	TV2
		Basic ordering unit (pieces)	2500	2500
2SB1184	PQR		○	-
2SB1243	PQR		-	○

 $h_{FE}$  values are classified as follows :

Item	P	Q	R
$h_{FE}$	82~180	120~270	180~390

●Electrical characteristic curves

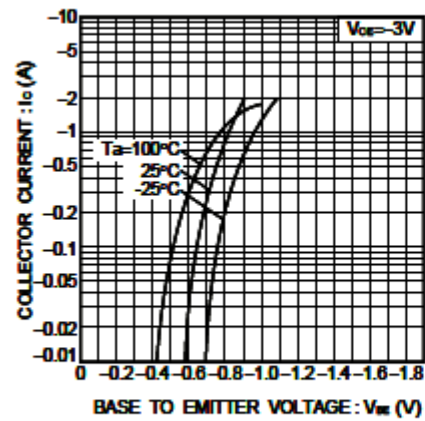


Fig.1 Grounded emitter propagation characteristics

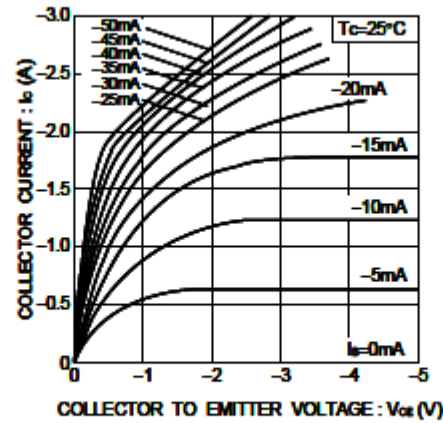


Fig.2 Grounded emitter output characteristics ( I )

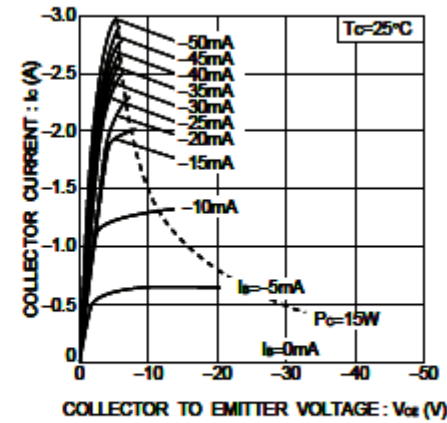


Fig.3 Grounded emitter output characteristics ( II )

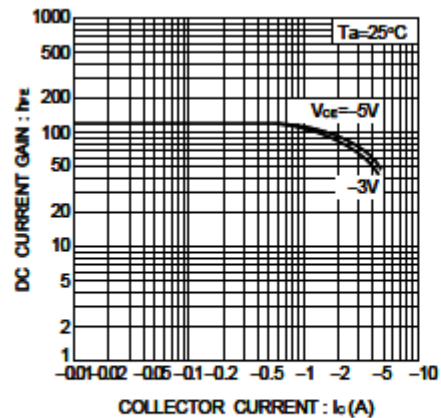


Fig.4 DC current gain vs. collector current ( I )

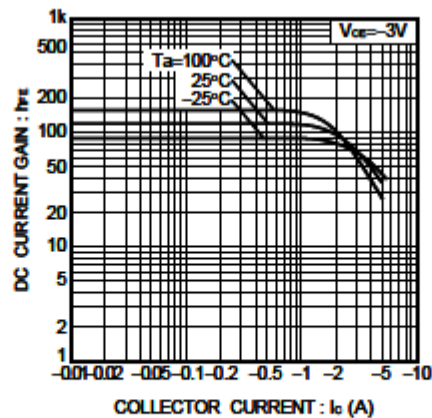


Fig.5 DC current gain vs. collector current ( II )

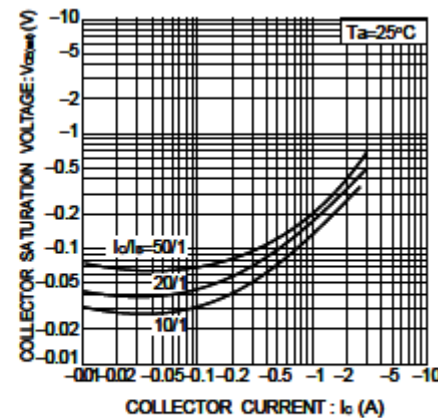


Fig.6 Collector-emitter saturation voltage vs. collector current



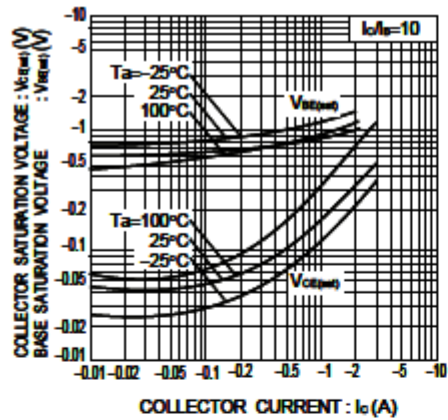


Fig.7 Collector-emitter saturation voltage vs. collector current  
Base-emitter saturation voltage vs. collector current

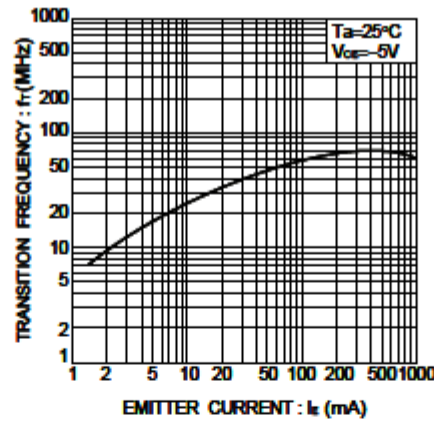


Fig.8 Gain bandwidth product vs. emitter current

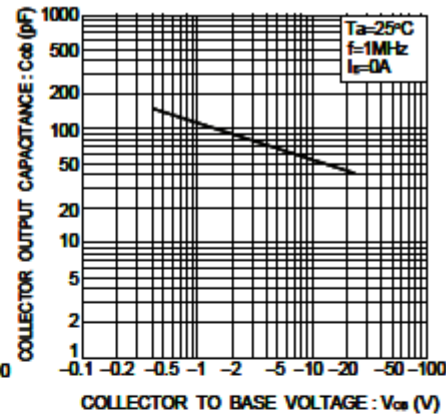


Fig.9 Collector output capacitance vs. collector base voltage

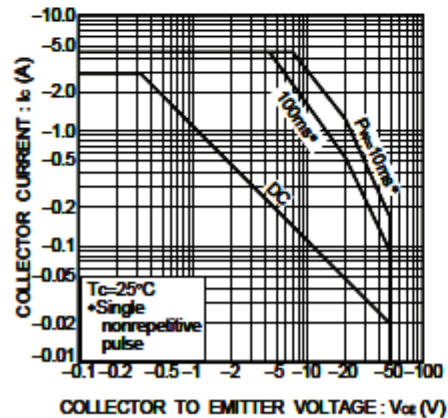


Fig.10 Safe operation area (2SB1184)

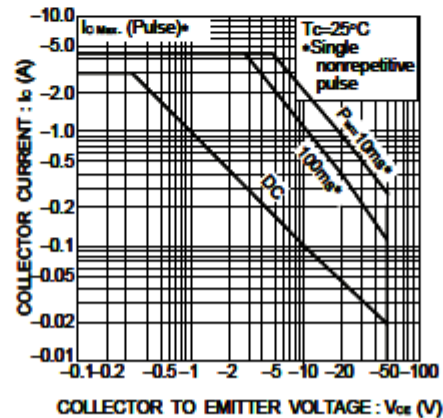


Fig.11 Safe operation area (2SB1243)

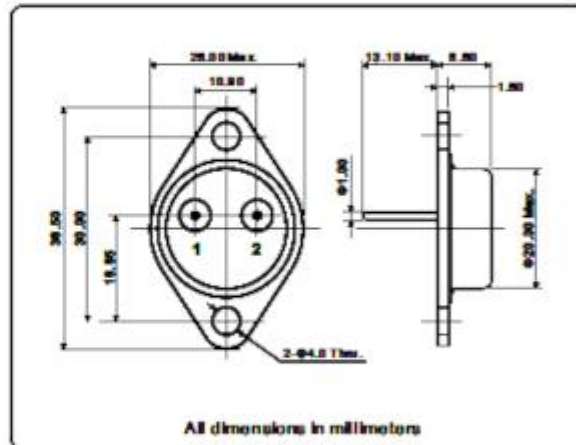
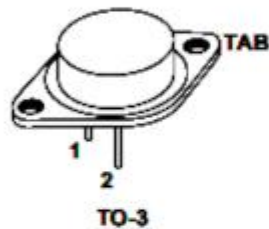
MJ15003 NPN MJ15004 PNP POTÊNICA DE AUDIO 20A 140V 250W METÁLICO TO-3



**MJ15003(NPN)**  
**MJ15004(PNP)**

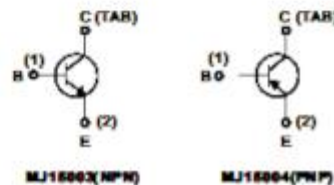
Nell High Power Products

**Complementary Silicon power transistors**  
**(20A / 140V / 250W)**



All dimensions in millimeters

INTERNAL SCHEMATIC DIAGRAM



**FEATURES**

- Designed for general-purpose switching and amplifier applications.
- DC current gain- $h_{FE} = 25$  (Min) @  $I_C = 5$  Adc
- High safe operation area (100% tested) 250W @ 50V
- For low distortion complementary designs

**DESCRIPTION**

The MJ15003 is a silicon epitaxial-base planar NPN transistor mounted in JEDEC TO-3 metal case.

It is designed for high power audio, disk head positioners and other linear applications.

The complementary PNP type is MJ15004.

ABSOLUTE MAXIMUM RATINGS ( $T_C = 25^\circ\text{C}$ unless otherwise specified)			
SYMBOL	PARAMETER		UNIT
$V_{CBO}$	Collector to base voltage ( $I_E = 0$ )		$V_{dc}$
$V_{CEO}$	Collector to emitter voltage ( $I_B = 0$ )		
$V_{EBO}$	Emitter to base voltage		
$I_C$	Collector current - continuous		$A_{dc}$
$I_B$	Base current - continuous		
$I_E$	Emitter current - continuous		
$P_D$	Total power dissipation	$T_C = 25^\circ\text{C}$	W
	Derate above $25^\circ\text{C}$		$W/^\circ\text{C}$
$T_J$	Junction temperature		$^\circ\text{C}$
$T_{stg}$	Storage temperature		
$T_L$	Maximum lead temperature for soldering purposes : 1/16" from case for $\leq 10$ seconds		$^\circ\text{C}$

\*For PNP types voltage and current values are negative.

THERMAL CHARACTERISTICS ( $T_C = 25^\circ\text{C}$ unless otherwise specified)			
SYMBOL	PARAMETER		UNIT
$R_{th(j-c)}$	Thermal resistance, junction to case		$^\circ\text{C}/\text{W}$

ELECTRICAL CHARACTERISTICS ( $T_C = 25^\circ\text{C}$ unless otherwise specified)					
SYMBOL	PARAMETER	CONDITIONS	MIN	MAX	UNIT
$I_{CEX}$	Collector cutoff current	$V_{CE} = 140\text{V}, V_{BE(off)} = 1.5\text{V}$		100	$\mu\text{A}$
		$V_{CE} = 140\text{V}, V_{BE(off)} = 1.5\text{V}, T_C = 150^\circ\text{C}$		2.0	mA
$I_{CEO}$	Collector cutoff current	$V_{CE} = 140\text{V}, I_B = 0$		250	$\mu\text{A}$
$I_{EBO}$	Emitter cutoff current	$V_{EBO} = 5\text{V}, I_C = 0$		100	
$V_{CEQ(sus)}^*$	Collector to emitter sustaining voltage	$I_C = 200\text{mA}, I_B = 0$	140		V
$V_{CBO}$	Collector to base voltage	$I_E = 0$	140		
$V_{EBO}$	Emitter to base voltage	$I_C = 0$	5		
$h_{FE}$	Forward current transfer ratio (DC current gain)	$I_C = 5\text{A}, V_{CE} = 2\text{V}$	25	150	
$V_{CE(sat)}^*$	Collector to emitter saturation voltage	$I_C = 5\text{A}, I_B = 0.5\text{A}$		1.0	V
$V_{BE(on)}^*$	Base to emitter on voltage	$I_C = 5\text{A}, V_{CE} = 2\text{V}$		2.0	
$f_T$	Transition frequency (current gain - bandwidth product)	$I_C = 0.5\text{A}, V_{CE} = 10\text{V}, f_{test} = 0.5\text{MHz}$	2.0		MHz
$C_{ob}$	Output capacitance	$V_{CB} = 10\text{V}, I_E = 0, f_{test} = 1\text{MHz}$		1000	pF
$I_{sfb}^*$	Second breakdown collector current with base forward biased	$V_{CE} = 50\text{V}, t = 1\text{s}, \text{non-repetitive}$	5		A
		$V_{CE} = 100\text{V}, t = 1\text{s}, \text{non-repetitive}$	1		

\*Pulsed : Pulse duration = 300  $\mu\text{s}$ , duty cycle 2%.

\*For PNP types voltage and current values are negative.

Fig.1 Power derating

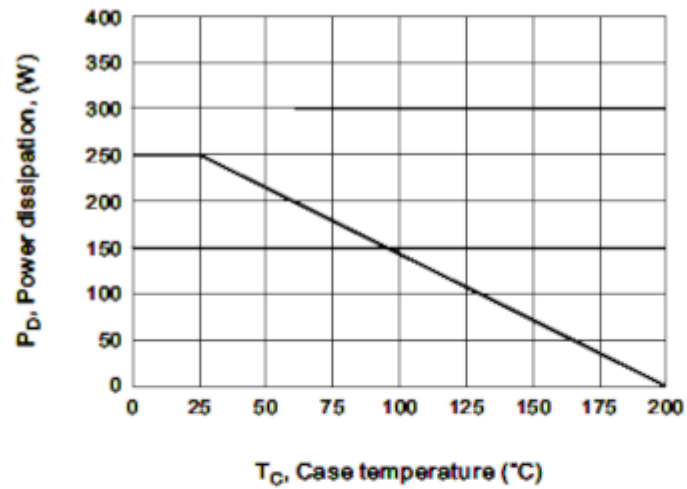
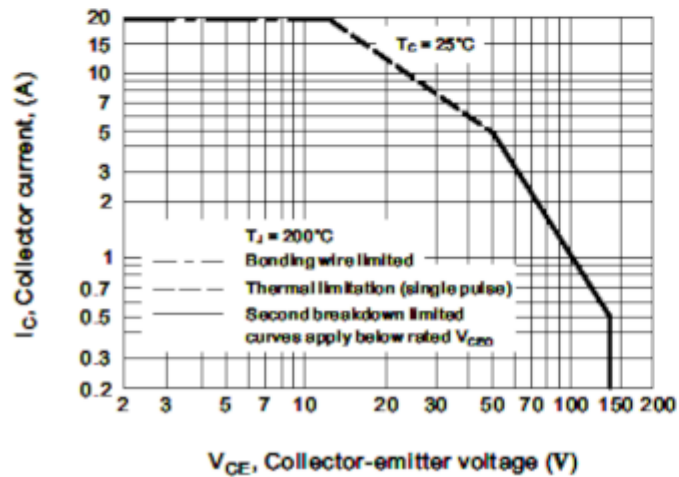


Fig.2 Active region safe operating area



There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate  $I_C$ - $V_{CE}$  limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of fig.2 is based on  $T_{J(pk)} = 200^\circ\text{C}$ ;  $T_C$  is variable depending on conditions. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

Fig.3 DC Current gain

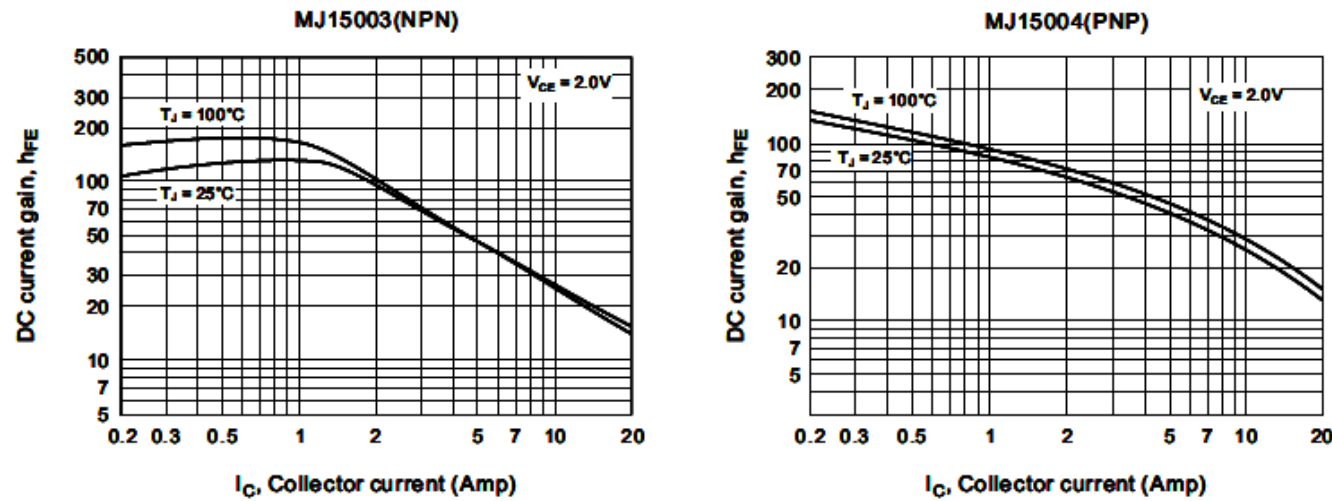
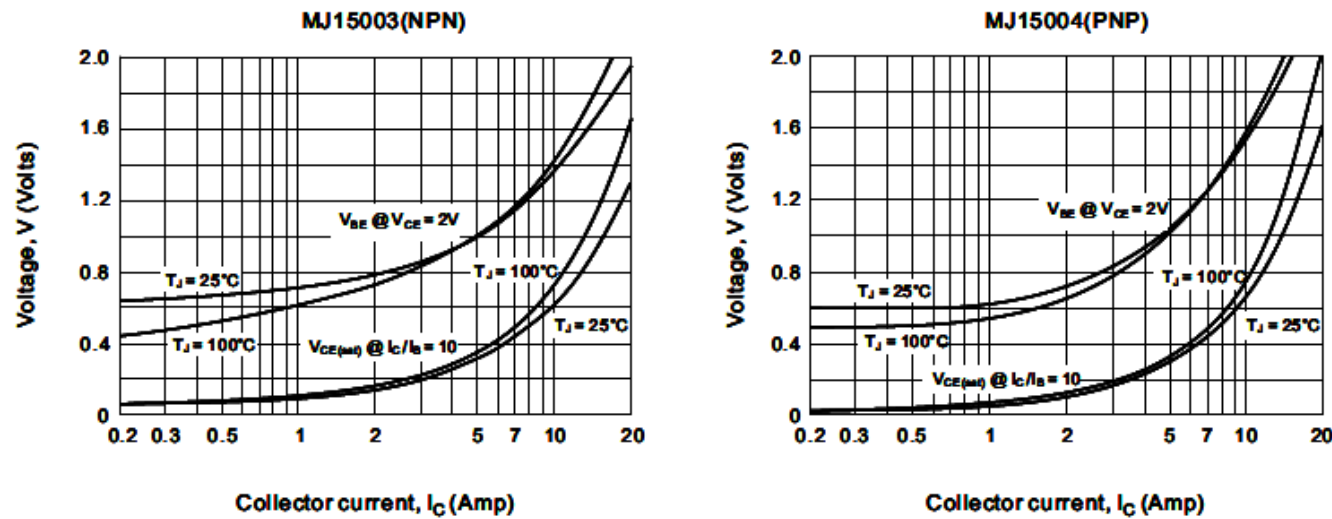


Fig.4 "ON" Characteristics



MJE15028 MJE15030 NPN MJE15029 MJE15031 PNP 8A 50 W 120\_150V

**MOTOROLA**  
**SEMICONDUCTOR TECHNICAL DATA**

Order this document  
by MJE15028/D

## Complementary Silicon Plastic Power Transistors

... designed for use as high-frequency drivers in audio amplifiers.

- DC Current Gain Specified to 4.0 Amperes  
 $h_{FE} = 40$  (Min) @  $I_C = 3.0$  Adc  
 $= 20$  (Min) @  $I_C = 4.0$  Adc
- Collector-Emitter Sustaining Voltage —  
 $V_{CEO(sus)} = 120$  Vdc (Min) — MJE15028, MJE15029  
 $= 150$  Vdc (Min) — MJE15030, MJE15031
- High Current Gain — Bandwidth Product  
 $f_T = 30$  MHz (Min) @  $I_C = 500$  mAdc
- TO-220AB Compact Package

**MAXIMUM RATINGS**

Rating	Symbol	MJE15028 MJE15029	MJE15030 MJE15031	Unit
Collector-Emitter Voltage	$V_{CEO}$	120	150	Vdc
Collector-Base Voltage	$V_{CB}$	120	150	Vdc
Emitter-Base Voltage	$V_{EB}$	5.0		Vdc
Collector Current — Continuous — Peak	$I_C$	8.0 16		Adc
Base Current	$I_B$	2.0		Adc
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	50 0.40		Watts W/°C
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	2.0 0.016		Watts W/°C
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-65 to +150		°C

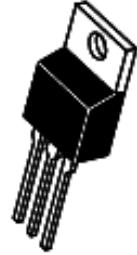
**NPN**  
**MJE15028\***

**MJE15030\***  
**PNP**  
**MJE15029\***

**MJE15031\***

\*Motorola Preferred Device

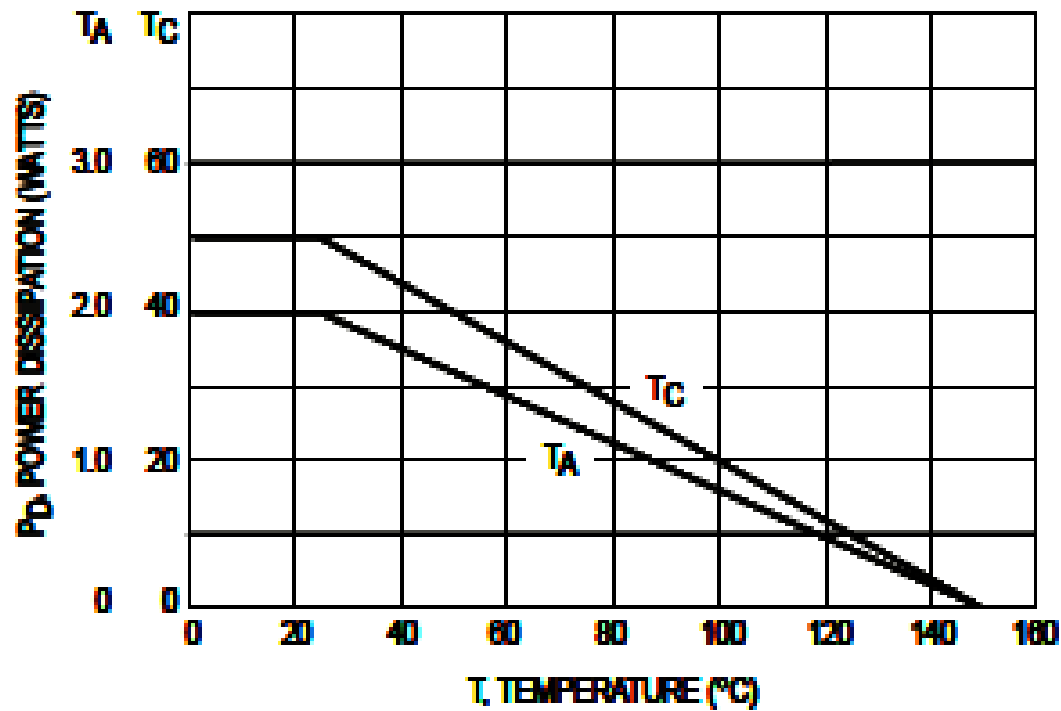
**8 AMPERE**  
**POWER TRANSISTORS**  
**COMPLEMENTARY**  
**SILICON**  
**120-150 VOLTS**  
**50 WATTS**



**CASE 221A-06**  
**TO-220AB**

**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	2.5	$^{\circ}C/W$
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	62.5	$^{\circ}C/W$



**Figure 1. Power Derating**



**MJE15028 MJE15030 MJE15029 MJE15031**

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
----------------	--------	-----	-----	------

**OFF CHARACTERISTICS**

Collector-Emitter Sustaining Voltage (1) ( $I_C = 10\text{ mAdc}$ , $I_B = 0$ )	MJE15028, MJE15029 MJE15030, MJE15031	$V_{CEO(sus)}$	120 150	— —	Vdc
Collector Cutoff Current ( $V_{CE} = 120\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 150\text{ Vdc}$ , $I_B = 0$ )	MJE15028, MJE15029 MJE15030, MJE15031	$I_{CEO}$	— —	0.1 0.1	mAdc
Collector Cutoff Current ( $V_{CB} = 120\text{ Vdc}$ , $I_E = 0$ ) ( $V_{CB} = 150\text{ Vdc}$ , $I_E = 0$ )	MJE15028, MJE15029 MJE15030, MJE15031	$I_{CBO}$	— —	10 10	$\mu\text{Adc}$
Emitter Cutoff Current ( $V_{BE} = 5.0\text{ Vdc}$ , $I_C = 0$ )		$I_{EBO}$	—	10	$\mu\text{Adc}$

**ON CHARACTERISTICS (1)**

DC Current Gain ( $I_C = 0.1\text{ Adc}$ , $V_{CE} = 2.0\text{ Vdc}$ ) ( $I_C = 2.0\text{ Adc}$ , $V_{CE} = 2.0\text{ Vdc}$ ) ( $I_C = 3.0\text{ Adc}$ , $V_{CE} = 2.0\text{ Vdc}$ ) ( $I_C = 4.0\text{ Adc}$ , $V_{CE} = 2.0\text{ Vdc}$ )		$h_{FE}$	40 40 40 20	— — — —	—
DC Current Gain Linearity ( $V_{CE}$ From 2.0 V to 20 V, $I_C$ From 0.1 A to 3 A) (NPN TO PNP)		$h_{FE}$	Typ 2 3		
Collector-Emitter Saturation Voltage ( $I_C = 1.0\text{ Adc}$ , $I_B = 0.1\text{ Adc}$ )		$V_{CE(sat)}$	—	0.5	Vdc
Base-Emitter On Voltage ( $I_C = 1.0\text{ Adc}$ , $V_{CE} = 2.0\text{ Vdc}$ )		$V_{BE(on)}$	—	1.0	Vdc

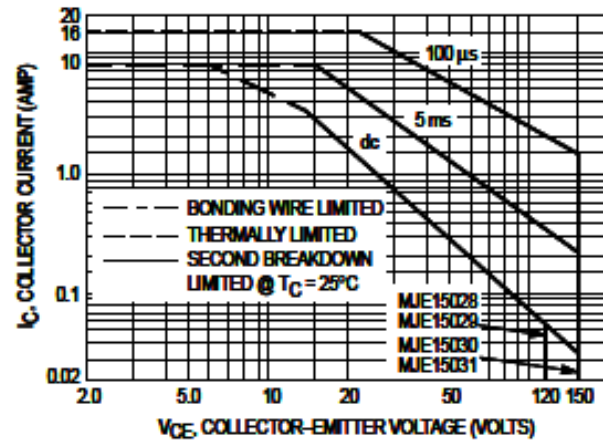
**DYNAMIC CHARACTERISTICS**

Current Gain — Bandwidth Product (2) ( $I_C = 500\text{ mAdc}$ , $V_{CE} = 10\text{ Vdc}$ , $f_{test} = 10\text{ MHz}$ )		$f_T$	30	—	MHz
---	--	-------	----	---	-----

(1) Pulse Test: Pulse Width  $\leq 300\ \mu\text{s}$ , Duty Cycle  $\leq 2.0\%$ .

(2)  $f_T = |h_{fe}| \cdot f_{test}$

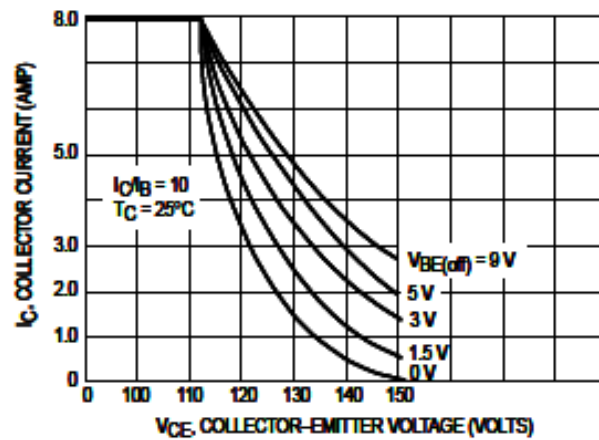
**MJE15028 MJE15030 MJE15029 MJE15031**



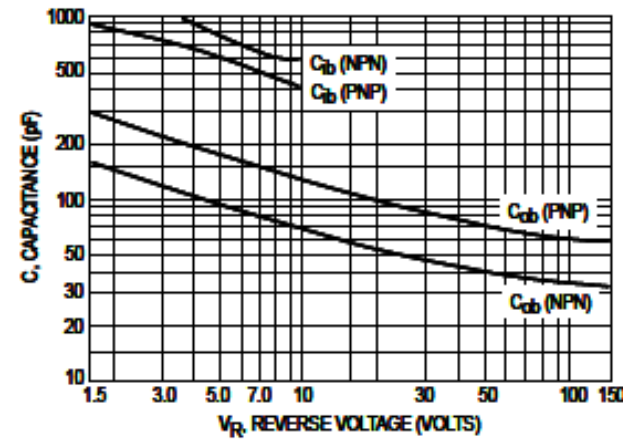
**Figure 3. Forward Bias Safe Operating Area**

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate  $I_C - V_{CE}$  limits of the transistor that must be observed for reliable operation, i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figures 3 and 4 is based on  $T_{J(pk)} = 150^\circ C$ ;  $T_C$  is variable depending on conditions. Second breakdown pulse limits are valid for duty cycles to 10% provided  $T_{J(pk)} < 150^\circ C$ .  $T_{J(pk)}$  may be calculated from the data in Figure 2. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.



**Figure 4. Reverse-Bias Switching Safe Operating Area**



**Figure 5. Capacitances**

**MJE15028 MJE15030 MJE15029 MJE15031**

**NPN — MJE15028 MJE15030**

**PNP — MJE15029 MJE15031**

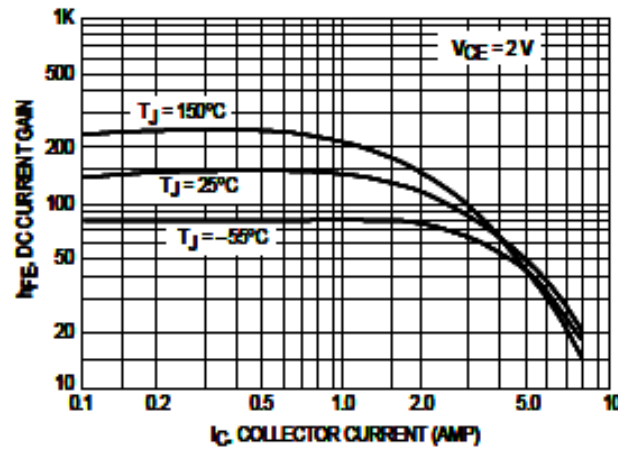
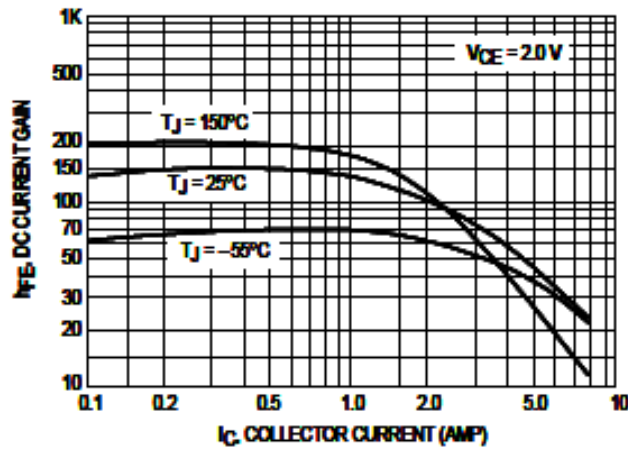


Figure 8. DC Current Gain

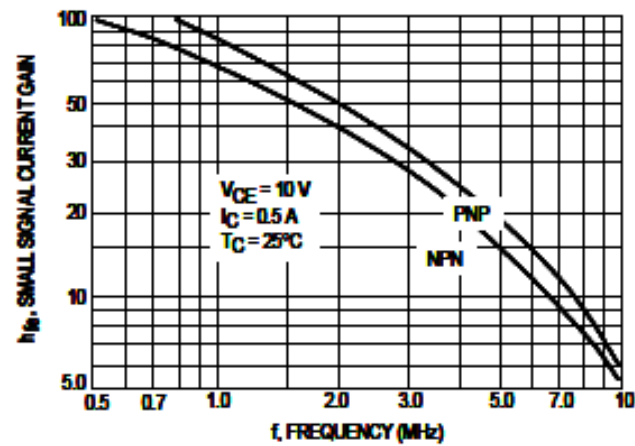


Figure 6. Small-Signal Current Gain

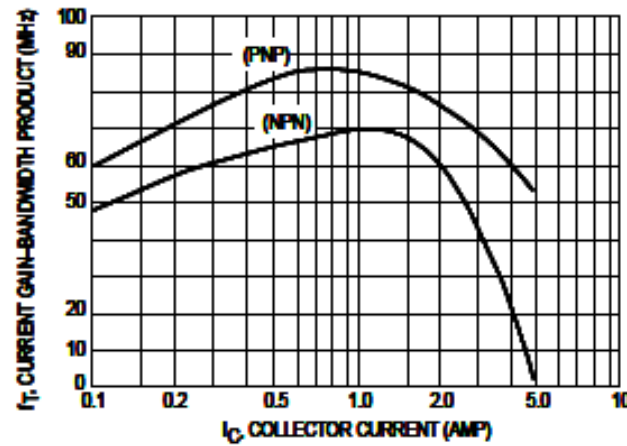


Figure 7. Current Gain-Bandwidth Product

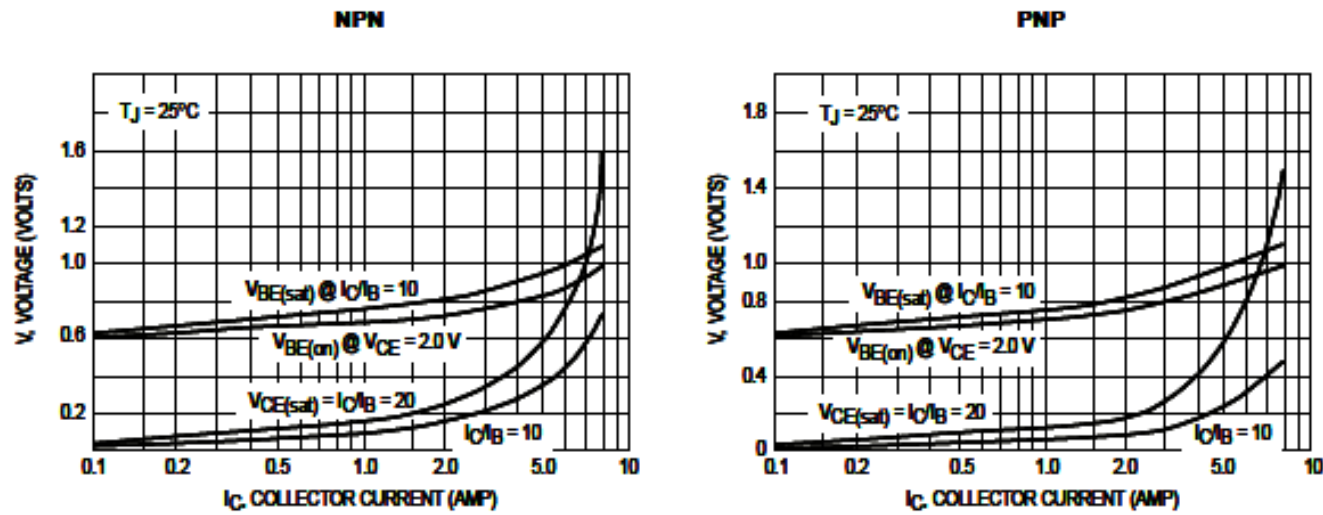


Figure 9. "On" Voltage

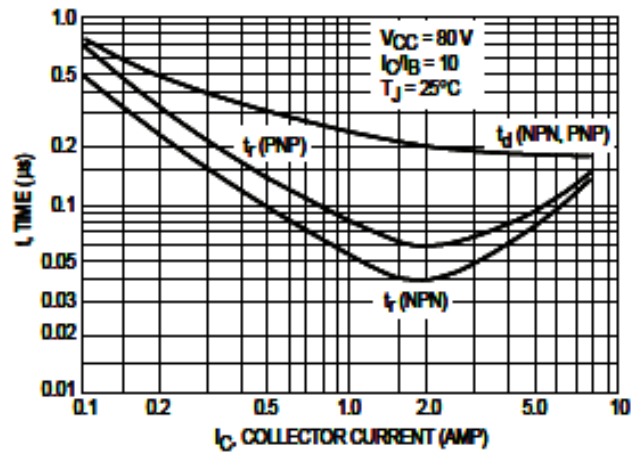


Figure 10. Turn-On Times

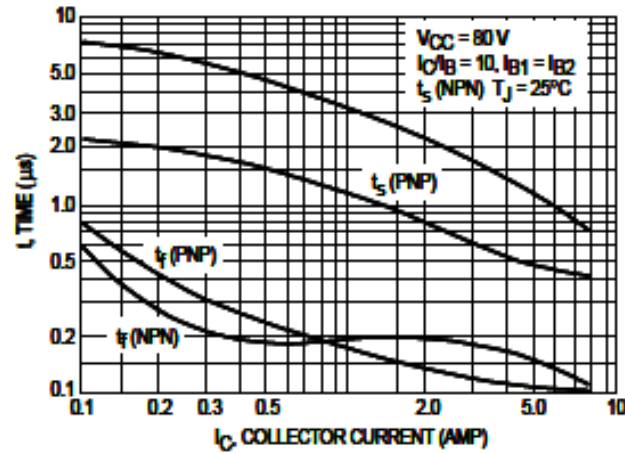
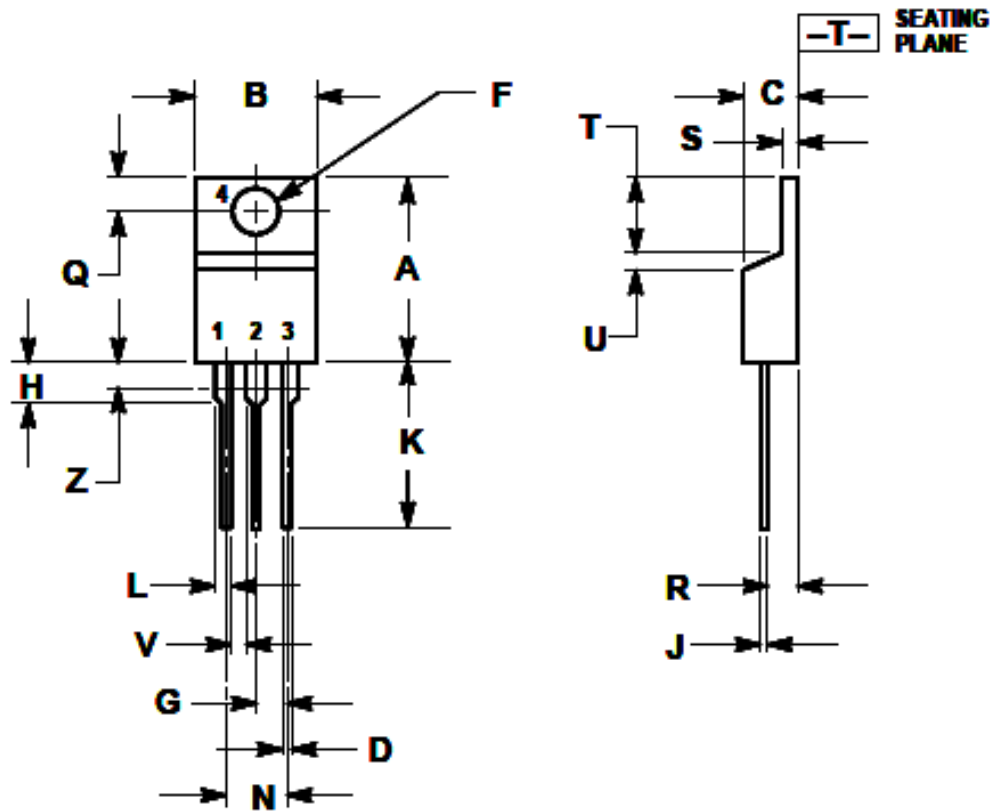


Figure 11. Turn-Off Times



**NOTES:**

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. DIMENSION Z DEFINES A ZONE WHERE ALL BODY AND LEAD IRREGULARITIES ARE ALLOWED.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.570	0.620	14.48	15.75
B	0.380	0.405	9.66	10.28
C	0.160	0.190	4.07	4.82
D	0.025	0.035	0.64	0.88
F	0.142	0.147	3.61	3.73
G	0.095	0.105	2.42	2.66
H	0.110	0.155	2.80	3.93
J	0.018	0.025	0.46	0.64
K	0.500	0.562	12.70	14.27
L	0.045	0.060	1.15	1.52
N	0.190	0.210	4.83	5.33
Q	0.100	0.120	2.54	3.04
R	0.080	0.110	2.04	2.79
S	0.045	0.055	1.15	1.39
T	0.235	0.255	5.97	6.47
U	0.000	0.050	0.00	1.27
V	0.045	—	1.15	—
Z	—	0.080	—	2.04

**STYLE 1:**

- PIN 1. BASE
- 2. COLLECTOR
- 3. EMITTER
- 4. COLLECTOR

MJE15032 (NPN) MJE15033 (PNP) 8A 250V 50W DRIVER DE POTÊNCIA DE AUDIO

## Complementary Silicon Plastic Power Transistors

... designed for use as high-frequency drivers in audio amplifiers.

- DC Current Gain Specified to 5.0 Amperes  
 $h_{FE} = 50$  (Min) @  $I_C = 0.5$  Adc  
 $= 10$  (Min) @  $I_C = 2.0$  Adc
- Collector-Emitter Sustaining Voltage —  
 $V_{CEO(sus)} = 250$  Vdc (Min) — MJE15032, MJE15033
- High Current Gain — Bandwidth Product  
 $f_T = 30$  MHz (Min) @  $I_C = 500$  mAdc
- TO-220AB Compact Package

### MAXIMUM RATINGS

Rating	Symbol	MJE15032 MJE15033	Unit
Collector-Emitter Voltage	$V_{CEO}$	250	Vdc
Collector-Base Voltage	$V_{CB}$	250	Vdc
Emitter-Base Voltage	$V_{EB}$	5.0	Vdc
Collector Current — Continuous — Peak	$I_C$	8.0 16	Adc
Base Current	$I_B$	2.0	Adc
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	50 0.40	Watts W/°C
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	2.0 0.016	Watts W/°C
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-65 to +150	°C

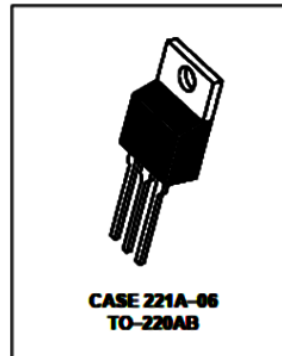
### THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	2.5	°C/W
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	62.5	°C/W

**NPN**  
**MJE15032\***  
**PNP**  
**MJE15033\***

\*Motorola Preferred Device

**8.0 AMPERES**  
**POWER TRANSISTORS**  
**COMPLEMENTARY**  
**SILICON**  
**250 VOLTS**  
**50 WATTS**



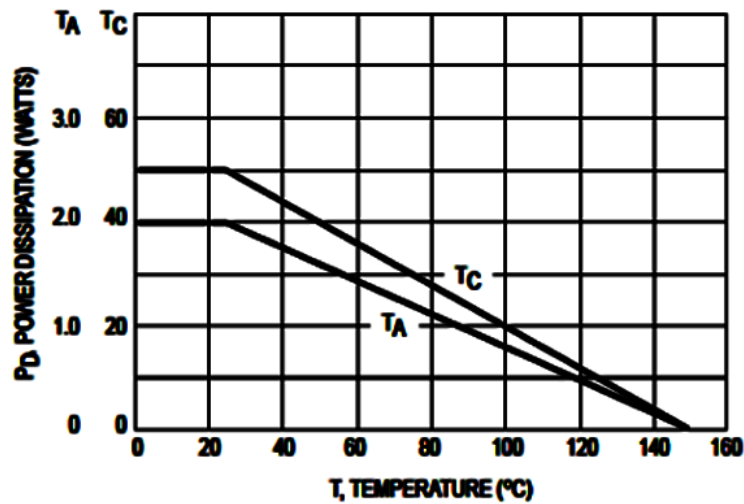


Figure 1. Power Derating

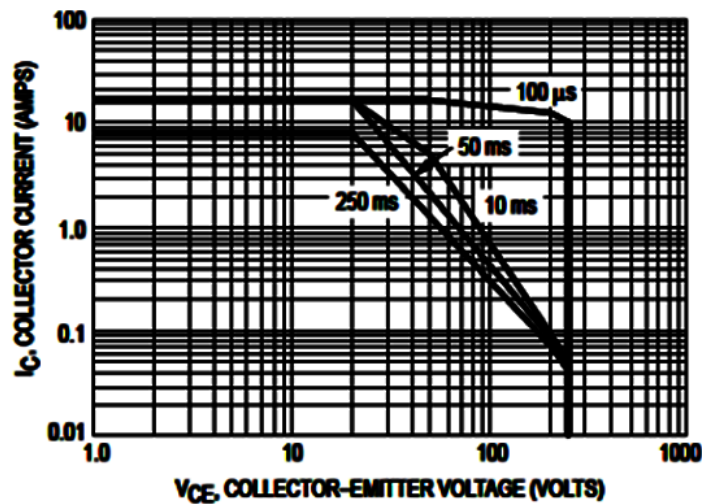
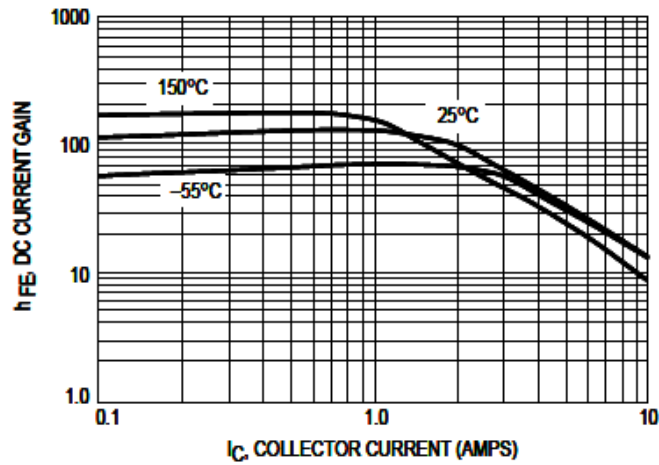


Figure 3. MJE15032 & MJE15033  
Safe Operating Area

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate  $I_C - V_{CE}$  limits of the transistor that must be observed for reliable operation, i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

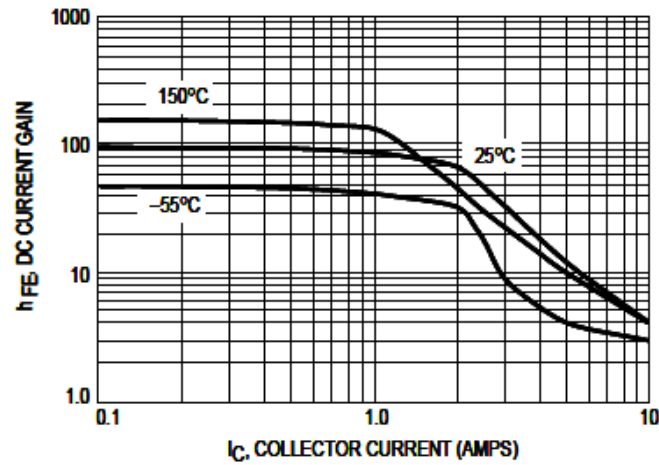
The data of Figures 3 and 4 is based on  $T_{J(pk)} = 150^\circ\text{C}$ ;  $T_C$  is variable depending on conditions. Second breakdown pulse limits are valid for duty cycles to 10% provided  $T_{J(pk)} < 150^\circ\text{C}$ .  $T_{J(pk)}$  may be calculated from the data in Figure 2. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

**NPN — MJE15032**

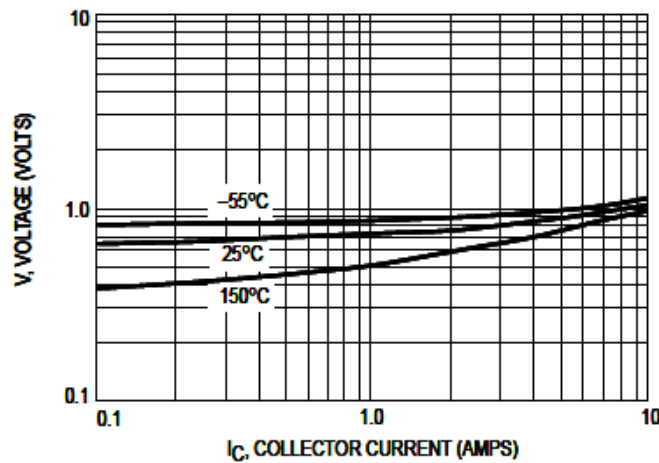


**Figure 4. NPN — MJE15032  
VCE = 5 V DC Current Gain**

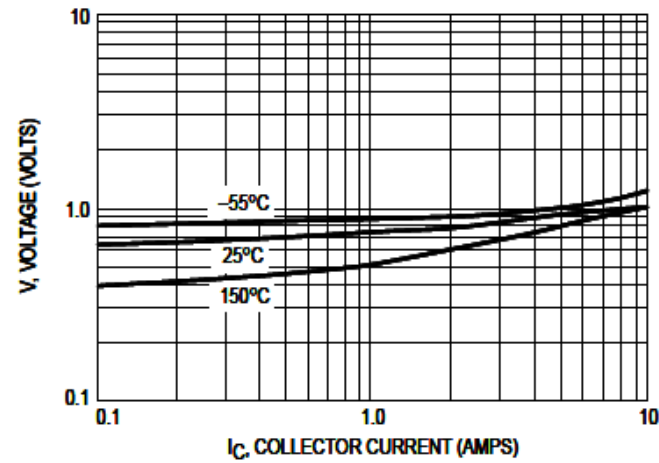
**PNP — MJE15033**



**Figure 5. PNP — MJE15033  
VCE = 5 V DC Current Gain**



**Figure 6. NPN — MJE15032  
VCE = 5 V VBE(on) Curve**

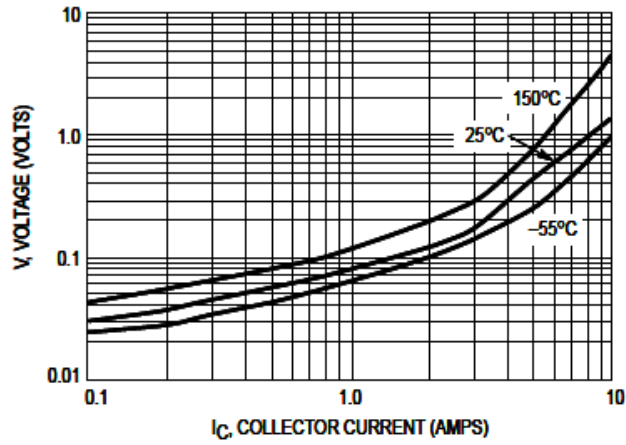


**Figure 7. PNP — MJE15033  
VCE = 5 V VBE(on) Curve**



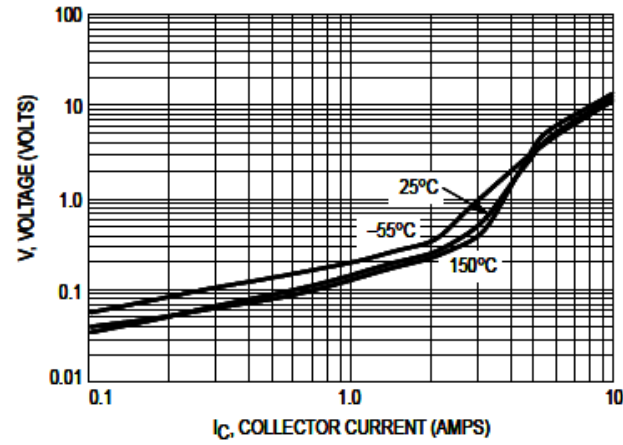
**MJE15032 MJE15033**

**NPN — MJE15032**

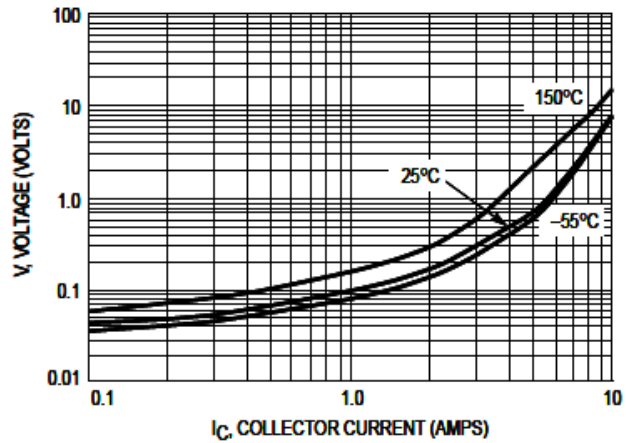


**Figure 8. NPN — MJE15032**  
 **$V_{CE(sat)}$   $I_C/I_B = 10$**

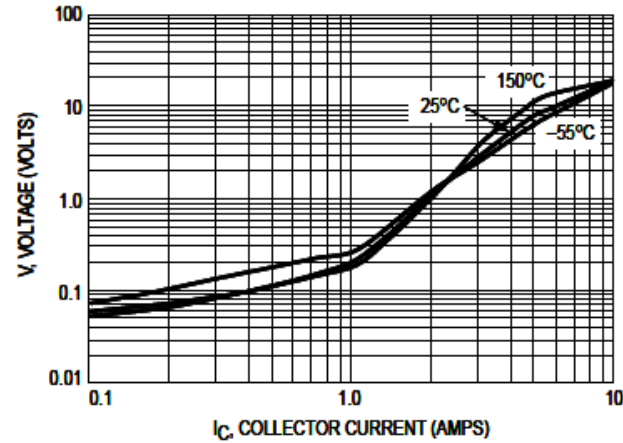
**PNP — MJE15033**



**Figure 9. PNP — MJE15033**  
 **$V_{CE(sat)}$   $I_C/I_B = 10$**

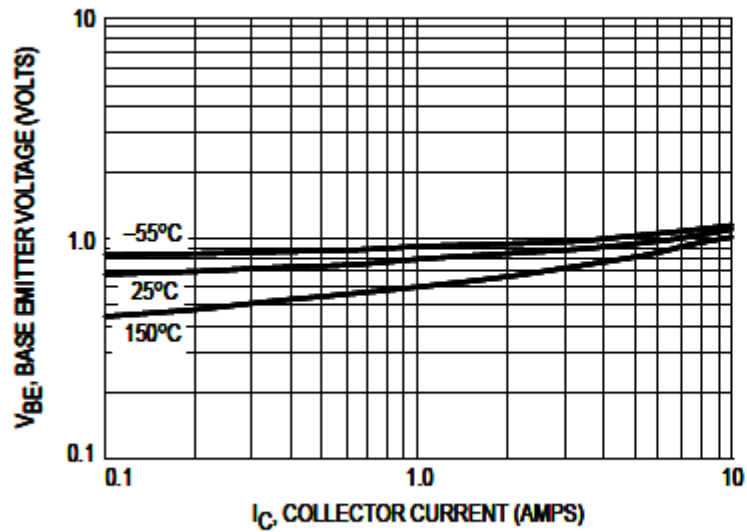


**Figure 10. NPN — MJE15032**  
 **$V_{CE(sat)}$   $I_C/I_B = 20$**

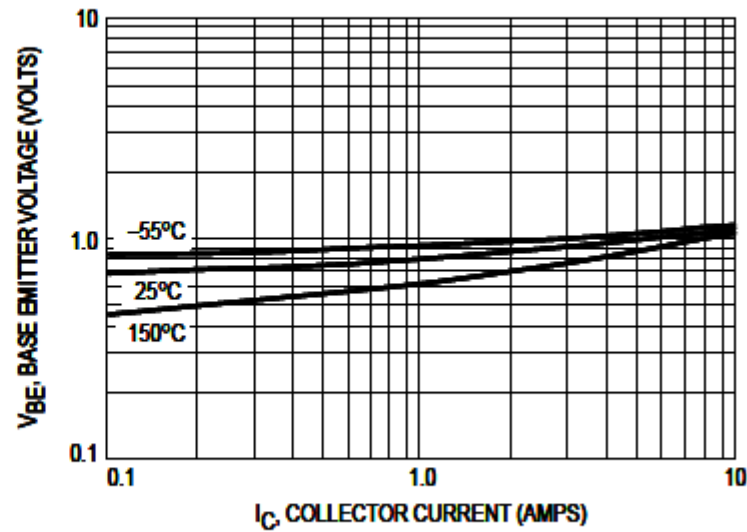


**Figure 11. PNP — MJE15033**  
 **$V_{CE(sat)}$   $I_C/I_B = 20$**

**Figure 10. NPN — MJE15032**  
 $V_{CE(sat)} I_C/I_B = 20$



**Figure 11. PNP — MJE15033**  
 $V_{CE(sat)} I_C/I_B = 20$



**Figure 12. NPN — MJE15032**  
 $V_{BE(sat)} I_C/I_B = 10$

**Figure 13. PNP — MJE15033**  
 $V_{BE(sat)} I_C/I_B = 10$

2SC5200 FJL4315 (NPN) 15A 230V POTÊNICA SAÍDA DE AUDIO (COMPLEMENTAR DO 2SA1943)



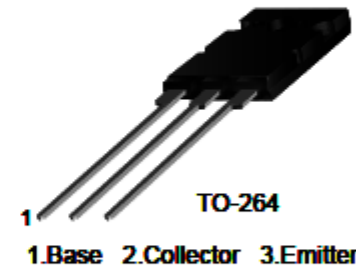
## 2SC5200/FJL4315 NPN Epitaxial Silicon Transistor

### Applications

- High-Fidelity Audio Output Amplifier
- General Purpose Power Amplifier

### Features

- High Current Capability:  $I_C = 15A$ .
- High Power Dissipation : 150watts.
- High Frequency : 30MHz.
- High Voltage :  $V_{CEO}=230V$
- Wide S.O.A for reliable operation.
- Excellent Gain Linearity for low THD.
- Complement to 2SA1943/FJL4215.
- Thermal and electrical Spice models are available.
- Same transistor is also available in:
  - TO3P package, 2SC5242/FJA4313 : 130 watts
  - TO220 package, FJP5200 : 80 watts
  - TO220F package, FJPF5200 : 50 watts



### Absolute Maximum Ratings\* $T_a = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Ratings	Units
$BV_{CBO}$	Collector-Base Voltage	230	V
$BV_{CEO}$	Collector-Emitter Voltage	230	V
$BV_{EBO}$	Emitter-Base Voltage	5	V
$I_C$	Collector Current(DC)	15	A
$I_B$	Base Current	1.5	A
$P_D$	Total Device Dissipation( $T_C=25^\circ\text{C}$ ) Derate above $25^\circ\text{C}$	150 1.04	W W/ $^\circ\text{C}$
$T_J, T_{STG}$	Junction and Storage Temperature	- 50 ~ +150	$^\circ\text{C}$

\* These ratings are limiting values above which the serviceability of any semiconductor device may be impaired.

### Thermal Characteristics\* $T_a=25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Max.	Units
$R_{\theta JC}$	Thermal Resistance, Junction to Case	0.83	$^\circ\text{C/W}$

\* Device mounted on minimum pad size

### $h_{FE}$ Classification

Classification	R	O
$h_{FE1}$	55 ~ 110	80 ~ 160

### Electrical Characteristics\* T<sub>a</sub>=25°C unless otherwise noted

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Units
$BV_{CBO}$	Collector-Base Breakdown Voltage	$I_C=5mA, I_E=0$	230			V
$BV_{CEO}$	Collector-Emitter Breakdown Voltage	$I_C=10mA, R_{BE}=\infty$	230			V
$BV_{EBO}$	Emitter-Base Breakdown Voltage	$I_E=5mA, I_C=0$	5			V
$I_{CBO}$	Collector Cut-off Current	$V_{CB}=230V, I_E=0$			5.0	$\mu A$
$I_{EBO}$	Emitter Cut-off Current	$V_{EB}=5V, I_C=0$			5.0	$\mu A$
$h_{FE1}$	DC Current Gain	$V_{CE}=5V, I_C=1A$	55		160	
$h_{FE2}$	DC Current Gain	$V_{CE}=5V, I_C=7A$	35	60		
$V_{CE(sat)}$	Collector-Emitter Saturation Voltage	$I_C=8A, I_B=0.8A$		0.4	3.0	V
$V_{BE(on)}$	Base-Emitter On Voltage	$V_{CE}=5V, I_C=7A$		1.0	1.5	V
$f_T$	Current Gain Bandwidth Product	$V_{CE}=5V, I_C=1A$		30		MHz
$C_{ob}$	Output Capacitance	$V_{CB}=10V, f=1MHz$		200		pF

\* Pulse Test: Pulse Width=20 $\mu s$ , Duty Cycle $\leq 2\%$

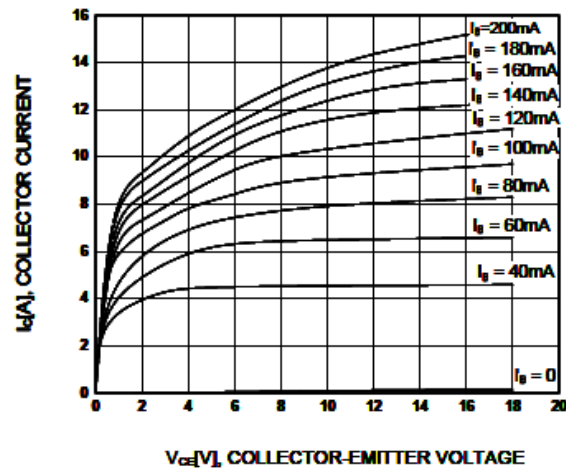


Figure 1. Static Characteristic

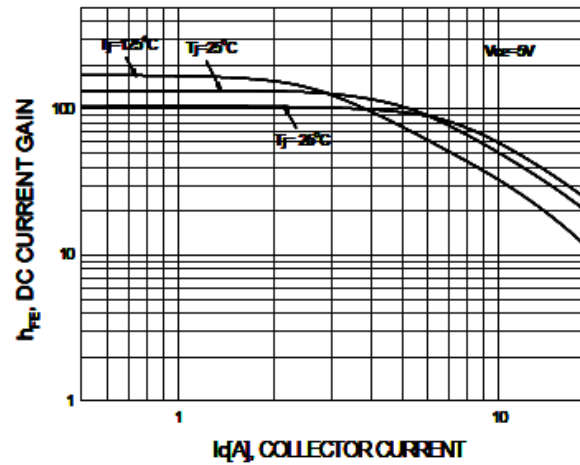


Figure 2. DC current Gain

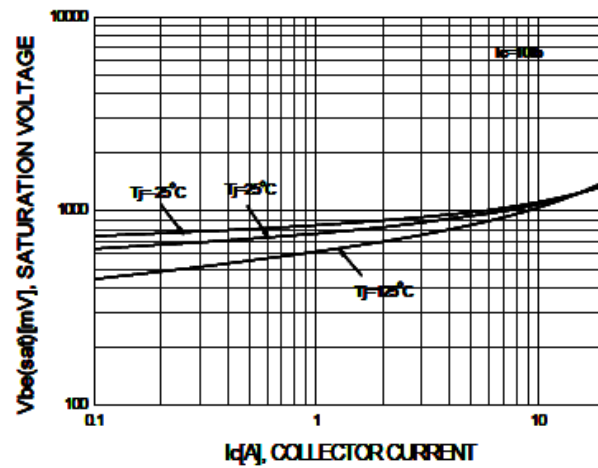


Figure 3. Base-Emitter Saturation Voltage

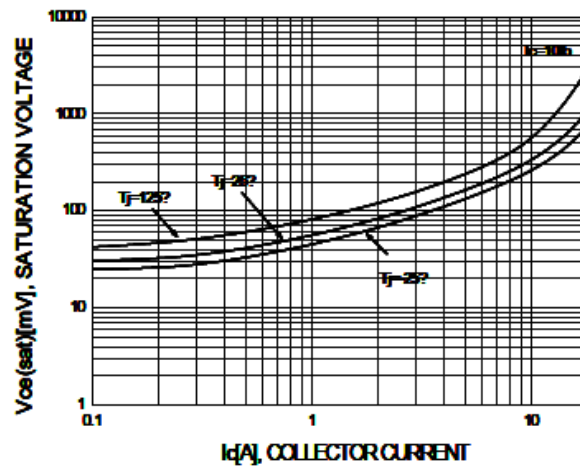


Figure 4. Collector-Emitter Saturation Voltage

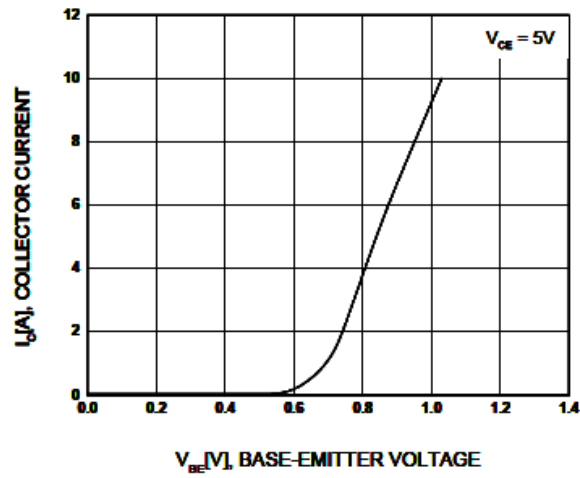


Figure 5. Base-Emitter On Voltage

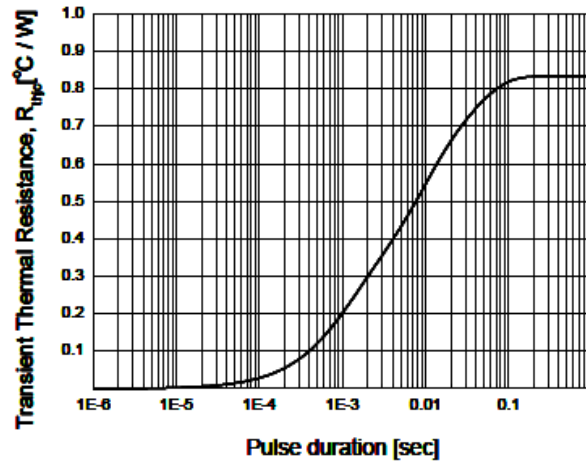


Figure 6. Thermal Resistance

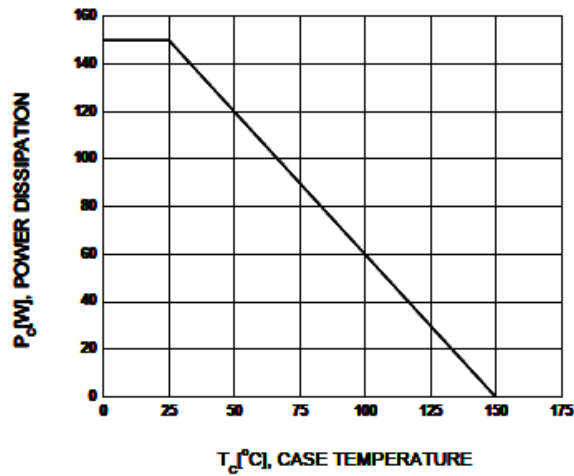


Figure 7. Power Derating

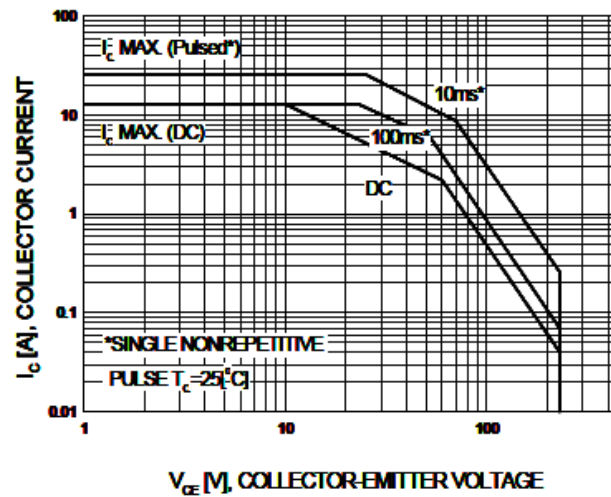


Figure 8. Safe Operating Area

2SA1943 FJL4215 (PNP) -15A -230V 150W ÁUDIO POTÊNCIA



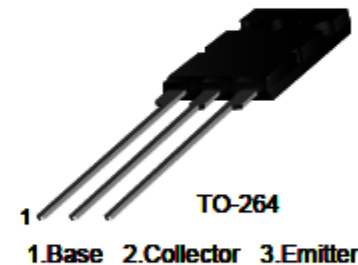
## 2SA1943/FJL4215 PNP Epitaxial Silicon Transistor

### Applications

- High-Fidelity Audio Output Amplifier
- General Purpose Power Amplifier

### Features

- High Current Capability:  $I_C = -15A$ .
- High Power Dissipation : 150watts.
- High Frequency : 30MHz.
- High Voltage :  $V_{CEO} = -230V$
- Wide S.O.A for reliable operation.
- Excellent Gain Linearity for low THD.
- Complement to 2SC5200/FJL4315.
- Full thermal and electrical Spice models are available.
- Same transistor is also available in:
  - TO3P package, 2SA1962/FJA4213 : 130 watts
  - TO220 package, FJP1943 : 80 watts
  - TO220F package, FJPF1943 : 50 watts





**Electrical Characteristics\***  $T_c=25^\circ\text{C}$  unless otherwise noted

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Units
$BV_{CBO}$	Collector-Base Breakdown Voltage	$I_C=-5\text{mA}, I_E=0$	-230			V
$BV_{CEO}$	Collector-Emitter Breakdown Voltage	$I_C=-10\text{mA}, R_{BE}=\infty$	-230			V
$BV_{EBO}$	Emitter-Base Breakdown Voltage	$I_E=-5\text{mA}, I_C=0$	-5			V
$I_{CBO}$	Collector Cut-off Current	$V_{CB}=-230\text{V}, I_E=0$			-5.0	$\mu\text{A}$
$I_{EBO}$	Emitter Cut-off Current	$V_{EB}=-5\text{V}, I_C=0$			-5.0	$\mu\text{A}$
$h_{FE1}$	DC Current Gain	$V_{CE}=-5\text{V}, I_C=-1\text{A}$	55		160	
$h_{FE2}$	DC Current Gain	$V_{CE}=-5\text{V}, I_C=-7\text{A}$	35	60		
$V_{CE(sat)}$	Collector-Emitter Saturation Voltage	$I_C=-8\text{A}, I_B=-0.8\text{A}$		-0.4	-3.0	V
$V_{BE(on)}$	Base-Emitter On Voltage	$V_{CE}=-5\text{V}, I_C=-7\text{A}$		-1.0	-1.5	V
$f_T$	Current Gain Bandwidth Product	$V_{CE}=-5\text{V}, I_C=-1\text{A}$		30		MHz
$C_{ob}$	Output Capacitance	$V_{CB}=-10\text{V}, f=1\text{MHz}$		360		pF

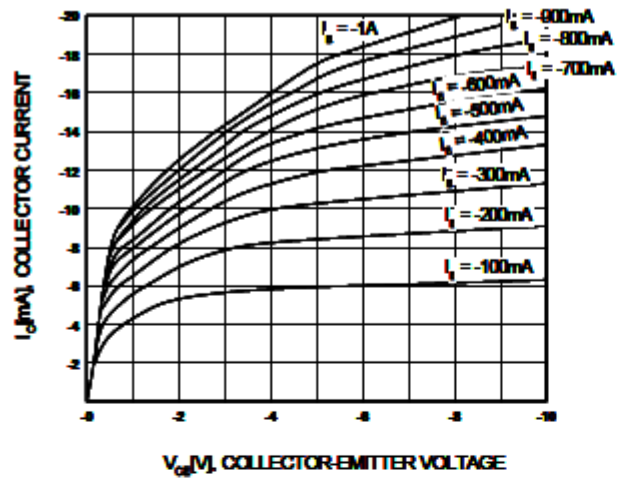


Figure 1. Static Characteristic

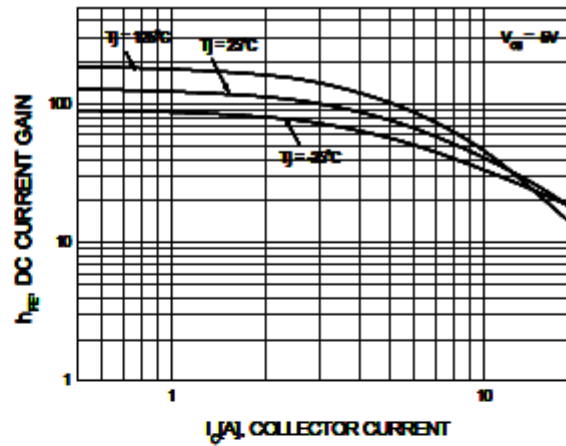
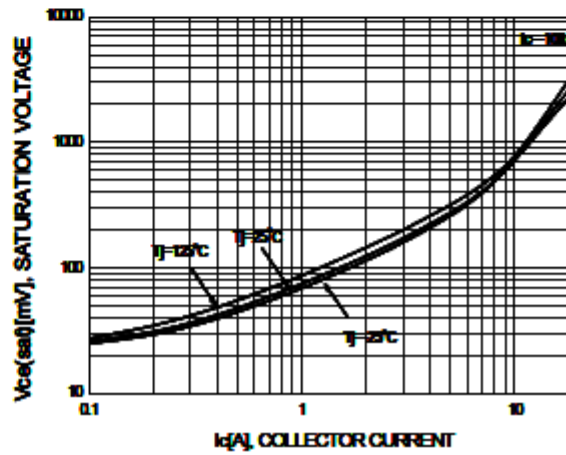
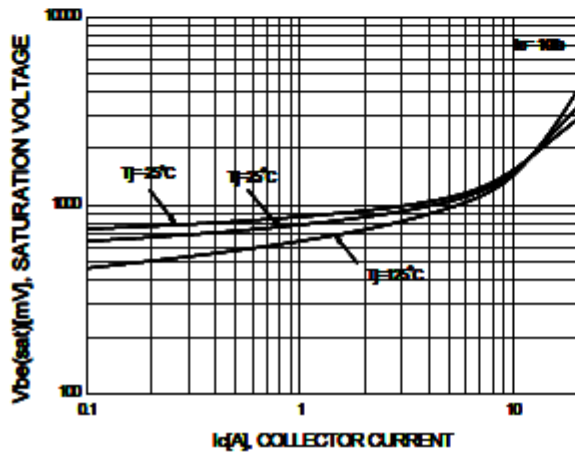


Figure 2. DC current Gain



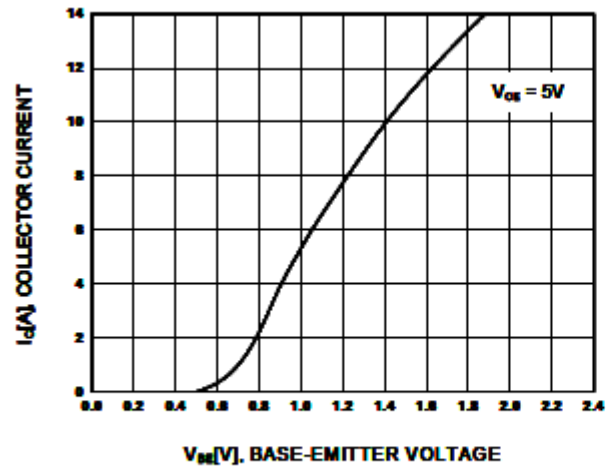


Figure 5. Base-Emitter On Voltage

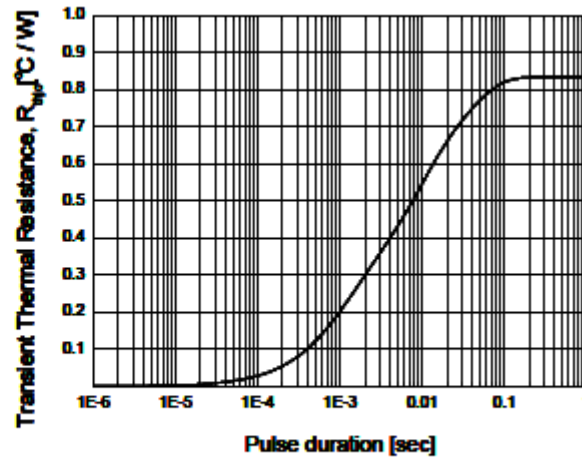


Figure 6. Thermal Resistance

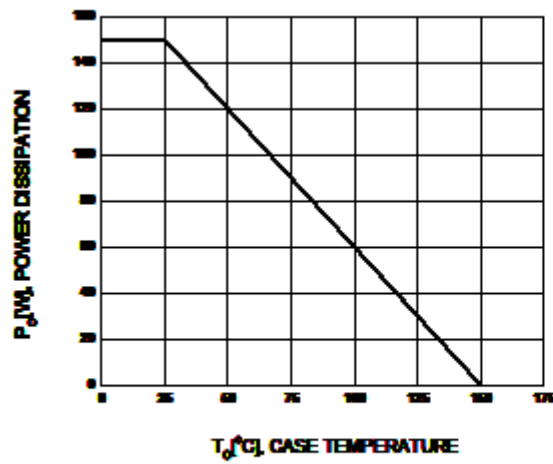


Figure 7. Power Derating

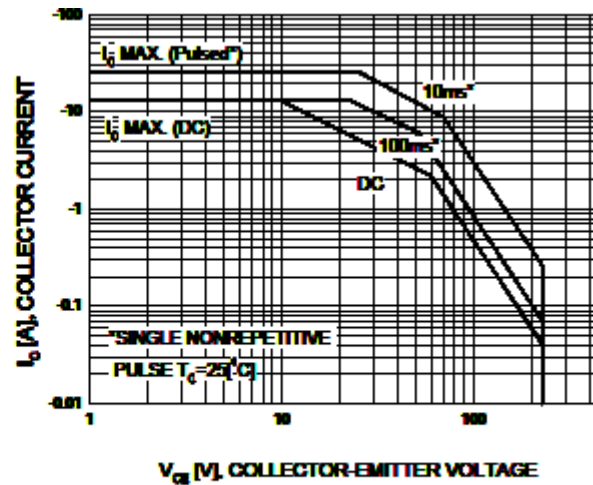


Figure 8. Safe Operating Area

KSP42 KSP43 HIGH VOLTAGE TRANSISTOR 300V NPN

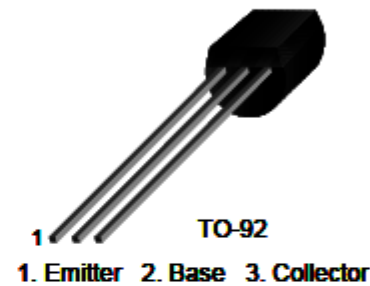
KSP42/43



### KSP42/43

#### High Voltage Transistor

- Collector-Emitter Voltage:  $V_{CE0}$  = KSP42: 300V  
KSP43: 200V
- Collector Power Dissipation:  $P_C(\text{max})$  = 625mW



#### NPN Epitaxial Silicon Transistor

#### Absolute Maximum Ratings $T_B = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Value	Units
$V_{CB0}$	Collector Base Voltage		
	: KSP42	300	V
	: KSP43	200	V
$V_{CE0}$	Collector-Emitter Voltage		
	: KSP42	300	V
	: KSP43	200	V
$V_{EB0}$	Emitter-Base Voltage	6	V
$I_C$	Collector Current	500	mA
$P_C$	Collector Power Dissipation	625	mW
$T_J$	Junction Temperature	150	$^\circ\text{C}$
$T_{STG}$	Storage Temperature	-55 ~ 150	$^\circ\text{C}$

**Electrical Characteristics**  $T_a=25^\circ\text{C}$  unless otherwise noted

Symbol	Parameter	Test Condition	Min.	Max.	Units
$BV_{CB0}$	Collector-Base Breakdown Voltage	$I_C=100\mu\text{A}, I_E=0$	300		V
	: KSP42				
	: KSP43		200		V
$BV_{CEO}$	* Collector-Emitter Breakdown Voltage	$I_C=1\text{mA}, I_B=0$	300		V
	: KSP42				
	: KSP43		200		V
$BV_{EBO}$	Emitter-Base Breakdown Voltage	$I_E=100\mu\text{A}, I_C=0$	6		V
$I_{CBO}$	Collector Cut-off Current	$V_{CB}=200\text{V}, I_E=0$ $V_{CB}=160\text{V}, I_E=0$			nA
	: KSP42				
	: KSP43		100		nA
$I_{EBO}$	Emitter Cut-off Current	$V_{BE}=6\text{V}, I_C=0$ $V_{BE}=4\text{V}, I_C=0$			nA
	: KSP42				
	: KSP43		100		nA
$h_{FE}$	* DC Current Gain	$V_{CE}=10\text{V}, I_C=1\text{mA}$ $V_{CE}=10\text{V}, I_C=10\text{mA}$ $V_{CE}=10\text{V}, I_C=30\text{mA}$	25		
			40		
			40		
$V_{CE}(\text{sat})$	* Collector-Emitter Saturation Voltage	$I_C=20\text{mA}, I_B=2\text{mA}$		0.5	V
$V_{BE}(\text{sat})$	* Base-Emitter Saturation Voltage	$I_C=20\text{mA}, I_B=2\text{mA}$		0.9	V
$C_{ob}$	Output Capacitance	$V_{CB}=20\text{V}, I_E=0$ $f=1\text{MHz}$			pF
	: KSP42				
	: KSP43		3		pF
$f_T$	Current Gain Bandwidth Product	$V_{CE}=20\text{V}, I_C=10\text{mA}$ $f=100\text{MHz}$	50		MHz

\* Pulse Test:  $PW \leq 300\mu\text{s}$ , Duty Cycle  $\leq 2\%$

## Typical Characteristics

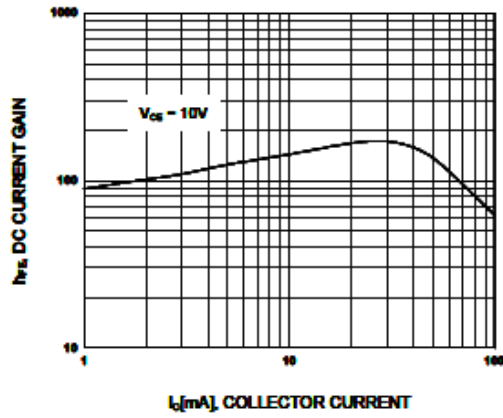


Figure 1. DC current Gain

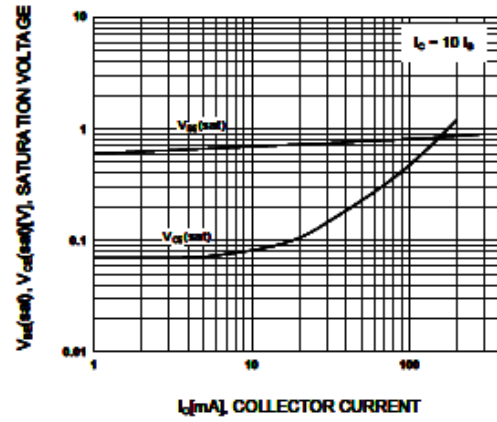


Figure 2. Collector-Emitter Saturation Voltage  
Base-Emitter Saturation Voltage

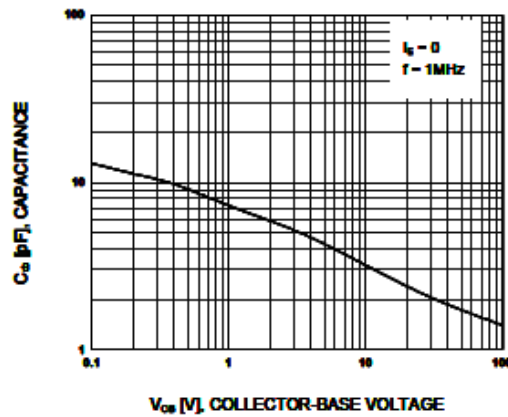


Figure 3. Collector-Base Capacitance

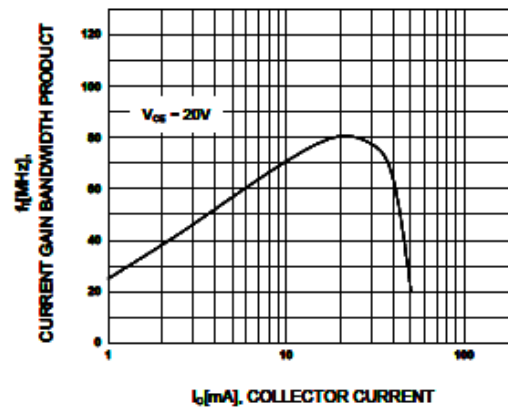
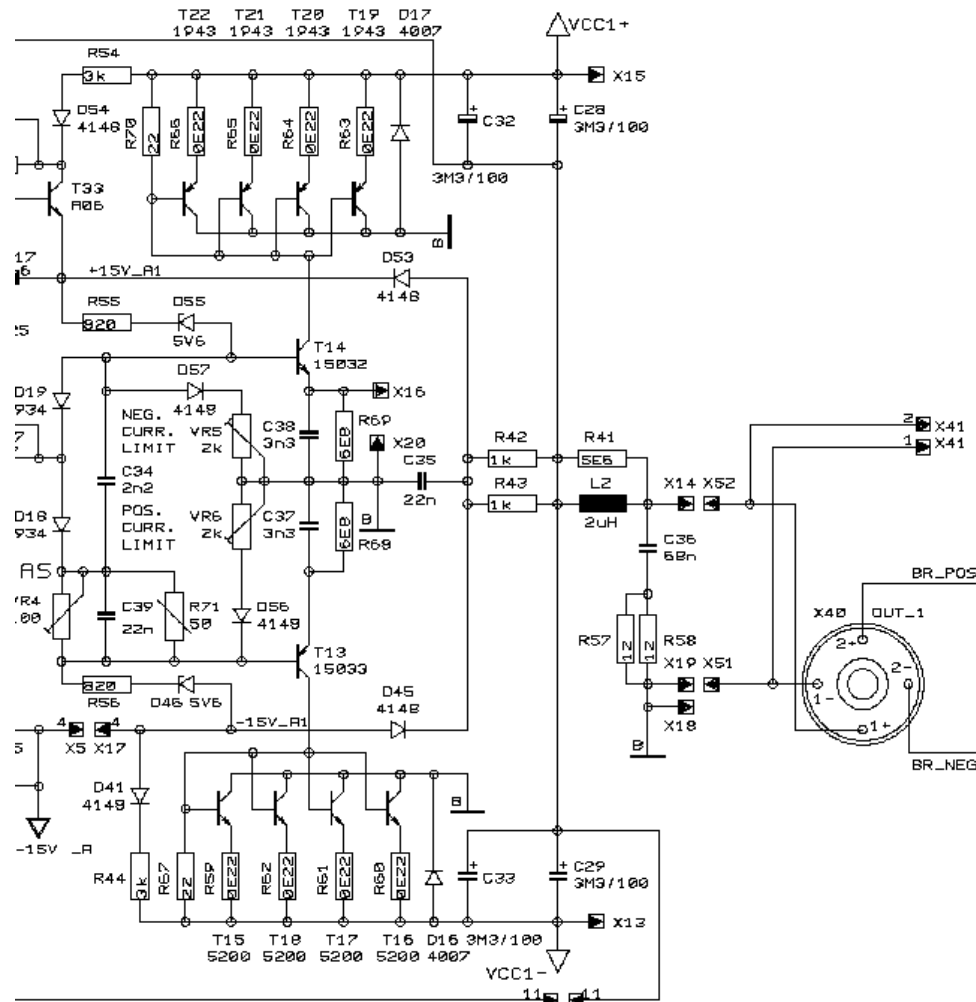


Figure 4. Current Gain Bandwidth Product

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- 2SD1760 2SD1864 2SD1762 NPN 3A 50V DRIVER BP
- 2SB1184 2SB1243 PNP 3A 50V DRIVER BP
- MJE1528 MJE1530 MJE1532 8A 50W NPN DRIVER POT
- MJE1529 MJE1531 MJE1533 8A 50W PNP DRIVER POT
- 2SA1943 15A 150W PNP POT
- 2SC5200 15A 150W NPN POT
- KSP42 KSP42 HIGH VOLTAGE DRIVER

TRANSISTOR DARLINGTON



TIP 100 101 102 NPN TIP 105 106 107 PNP DARLINGTON 8A 80W

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- High DC Current Gain —  
hFE = 2500 (Typ) @ IC = 4.0 Adc
- Collector-Emitter Sustaining Voltage — @ 30 mAdc  
VCEO(sus) = 60 Vdc (Min) — TIP100, TIP105  
= 80 Vdc (Min) — TIP101, TIP106  
= 100 Vdc (Min) — TIP102, TIP107
- Low Collector-Emitter Saturation Voltage —  
VCE(sat) = 2.0 Vdc (Max) @ IC = 3.0 Adc  
= 2.5 Vdc (Max) @ IC = 8.0 Adc
- Monolithic Construction with Built-in Base-Emitter Shunt Resistors
- TO-220AB Compact Package

**\*MAXIMUM RATINGS**

Rating	Symbol	TIP100, TIP105	TIP101, TIP106	TIP102, TIP107	Unit
Collector-Emitter Voltage	VCEO	60	80	100	Vdc
Collector-Base Voltage	VCB	60	80	100	Vdc
Emitter-Base Voltage	VEB	5.0			Vdc
Collector Current — Continuous	IC	8.0			Adc
Peak		15			Adc
Base Current	IB	1.0			Adc
Total Power Dissipation @ TC = 25°C	PD	80			Watts
Derate above 25°C		0.64			W/°C
Unclamped Inductive Load Energy (1)	E	30			mJ
Total Power Dissipation @ TA = 25°C	PD	2.0			Watts
Derate above 25°C		0.016			W/°C
Operating and Storage Junction Temperature Range	TJ, Tstg	-85 to +150			°C

**THERMAL CHARACTERISTICS**

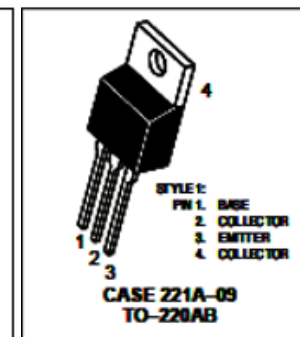
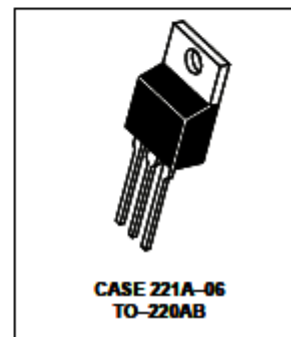
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	RθJC	1.56	°C/W
Thermal Resistance, Junction to Ambient	RθJA	62.5	°C/W

(1) IC = 1.1 A, L = 50 mH, P.R.F. = 10 Hz, VCC = 20 V, RPF = 100 Ω

**NPN**  
**TIP100**  
**TIP101\***  
**TIP102\***  
**PNP**  
**TIP105**  
**TIP106\***  
**TIP107\***

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**POWER TRANSISTORS**  
**60-80-100 VOLTS**  
**80 WATTS**



**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
<b>OFF CHARACTERISTICS</b>				
Collector–Emitter Sustaining Voltage (1) ( $I_C = 30\text{ mA}_{dc}$ , $I_B = 0$ )	TIP100, TIP105 TIP101, TIP106 TIP102, TIP107 $V_{CE(sus)}$	60 80 100	— — —	Vdc
Collector Cutoff Current ( $V_{CE} = 30\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 40\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 50\text{ Vdc}$ , $I_B = 0$ )	TIP100, TIP105 TIP101, TIP106 TIP102, TIP107 $I_{CEO}$	— — —	50 50 50	$\mu\text{A}_{dc}$
Collector Cutoff Current ( $V_{CB} = 60\text{ Vdc}$ , $I_E = 0$ ) ( $V_{CB} = 80\text{ Vdc}$ , $I_E = 0$ ) ( $V_{CB} = 100\text{ Vdc}$ , $I_E = 0$ )	TIP100, TIP105 TIP101, TIP106 TIP102, TIP107 $I_{CBO}$	— — —	50 50 50	$\mu\text{A}_{dc}$
Emitter Cutoff Current ( $V_{BE} = 5.0\text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	—	8.0	$\text{mA}_{dc}$
<b>ON CHARACTERISTICS (1)</b>				
DC Current Gain ( $I_C = 3.0\text{ A}_{dc}$ , $V_{CE} = 4.0\text{ Vdc}$ ) ( $I_C = 8.0\text{ A}_{dc}$ , $V_{CE} = 4.0\text{ Vdc}$ )	$h_{FE}$	1000 200	20,000 —	—
Collector–Emitter Saturation Voltage ( $I_C = 3.0\text{ A}_{dc}$ , $I_B = 6.0\text{ mA}_{dc}$ ) ( $I_C = 8.0\text{ A}_{dc}$ , $I_B = 80\text{ mA}_{dc}$ )	$V_{CE(sat)}$	— —	2.0 2.5	Vdc
Base–Emitter On Voltage ( $I_C = 8.0\text{ A}_{dc}$ , $V_{CE} = 4.0\text{ Vdc}$ )	$V_{BE(on)}$	—	2.8	Vdc
<b>DYNAMIC CHARACTERISTICS</b>				
Small–Signal Current Gain ( $I_C = 3.0\text{ A}_{dc}$ , $V_{CE} = 4.0\text{ Vdc}$ , $f = 1.0\text{ MHz}$ )	$h_{fe}$	4.0	—	—
Output Capacitance ( $V_{CB} = 10\text{ Vdc}$ , $I_E = 0$ , $f = 0.1\text{ MHz}$ )	TIP105, TIP106, TIP107 TIP100, TIP101, TIP102 $C_{ob}$	— —	300 200	pF

(1) Pulse Test: Pulse Width  $\leq 300\ \mu\text{s}$ , Duty Cycle  $\leq 2\%$ .

FIGURE -1 POWER DERATING

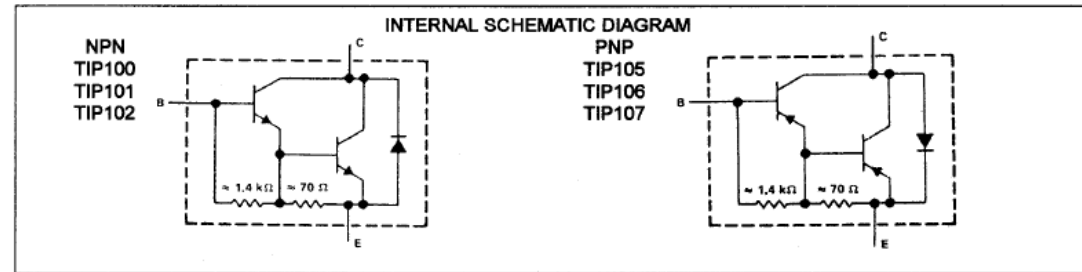
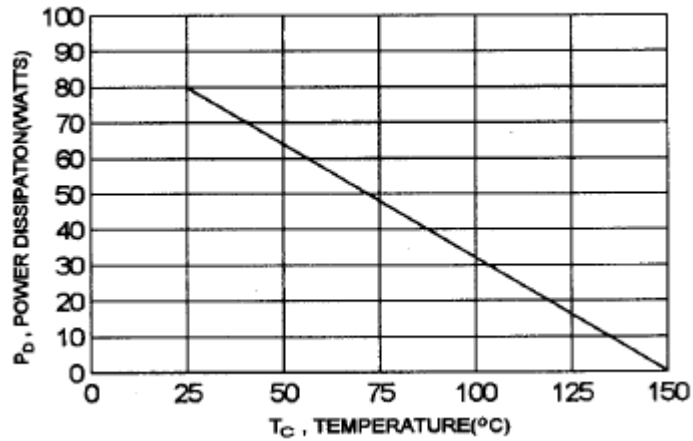
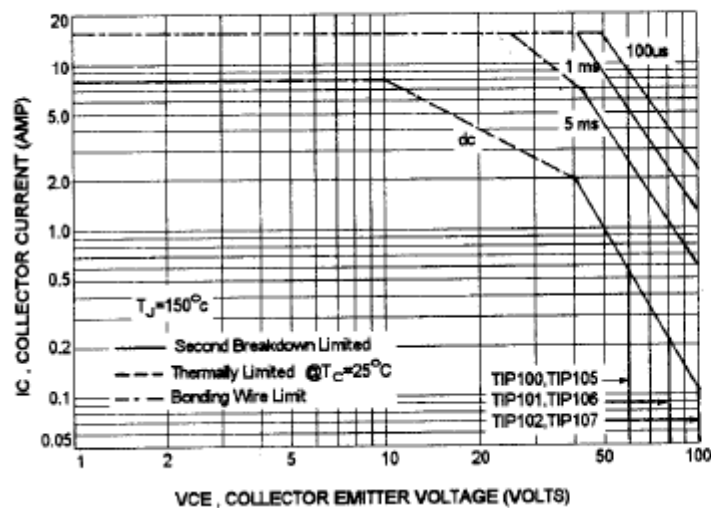


FIG-6 ACTIVE REGION SAFE OPERATING AREA



There are two limitation on the power handling ability of a transistor: average junction temperature and second breakdown safe operating area curves indicate  $I_C$ - $V_{CE}$  limits of the transistor that must be observed for reliable operation i.e., the transistor must not be subjected to greater dissipation than curves indicate.

The data of FIG-6 is base on  $T_{J(PK)}=150^\circ\text{C}$ ;  $T_C$  is variable depending on power level. second breakdown pulse limits are valid for duty cycles to 10% provided  $T_{J(PK)} \leq 150^\circ\text{C}$ . At high case temperatures, thermal limitation will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

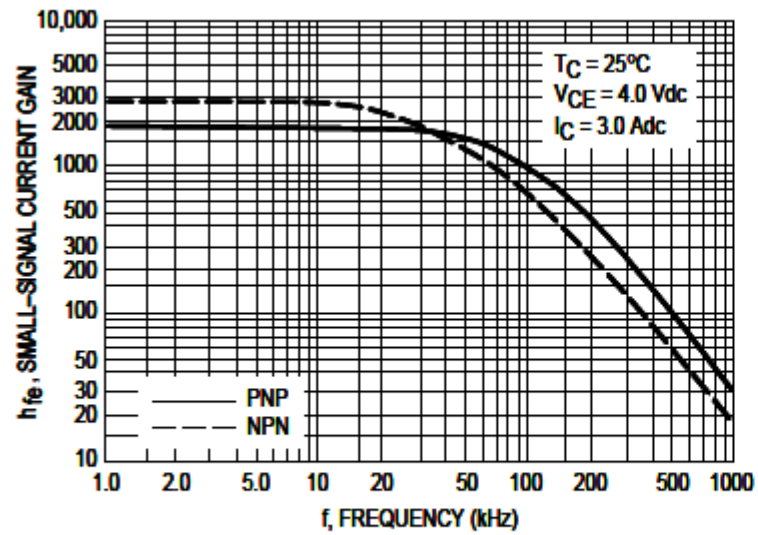


Figure 6. Small-Signal Current Gain

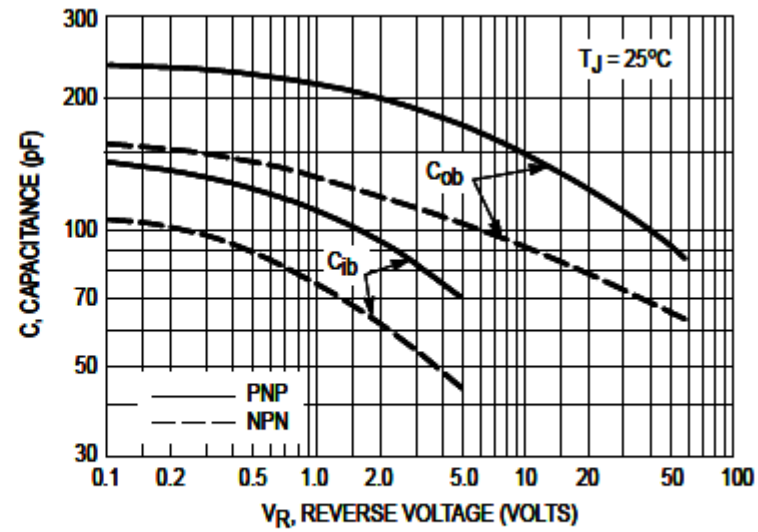


Figure 7. Capacitance

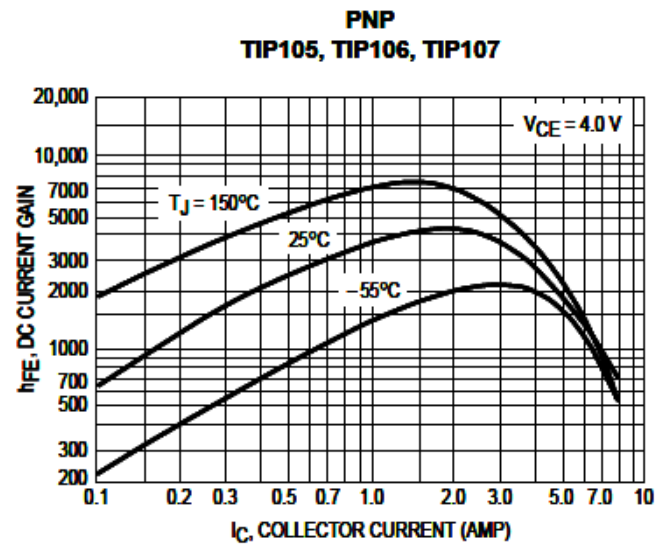
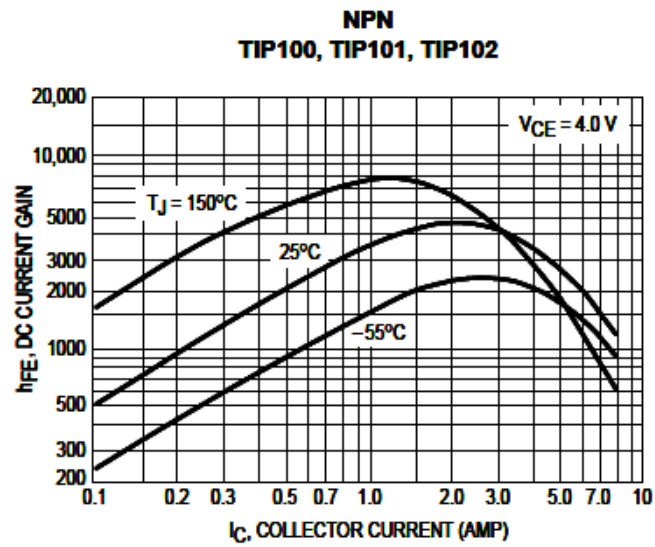


Figure 8. DC Current Gain

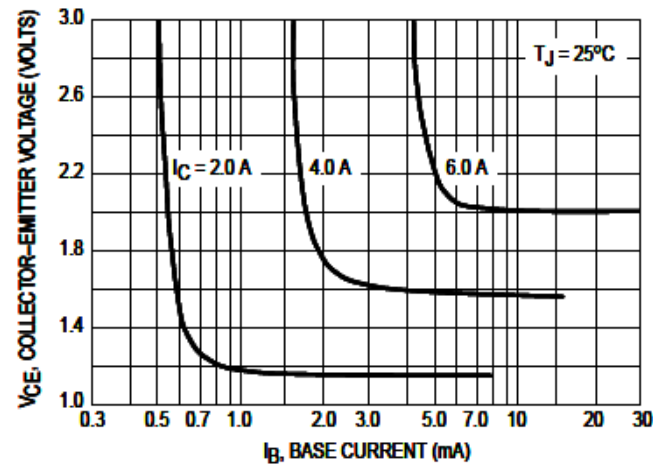
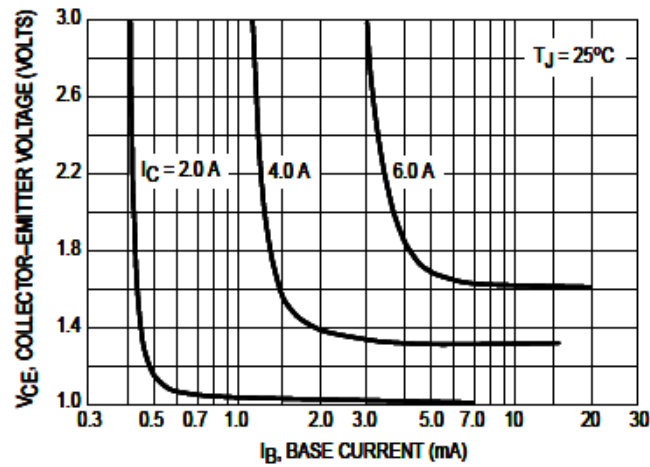


Figure 9. Collector Saturation Region

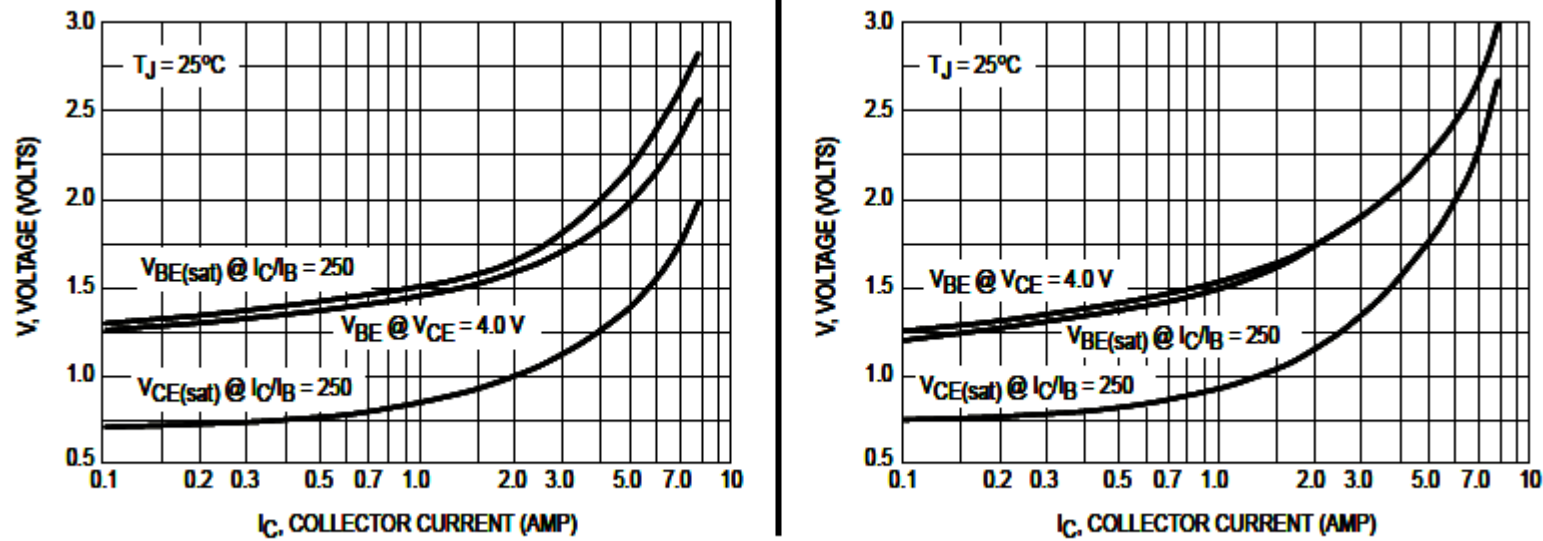


Figure 10. "On" Voltages

TIP120 TIP121 / TIP122 / NPN TIP125 TIP126 / TIP127 / PNP DARLINGTON 5A 65W

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 = 4.0 Adc
- Collector-Emitter Sustaining Voltage — @ 100 mAdc  
 $V_{CE(sus)} = 60$  Vdc (Min) — TIP120, TIP125  
 = 80 Vdc (Min) — TIP121, TIP126  
 = 100 Vdc (Min) — TIP122, TIP127
- Low Collector-Emitter Saturation Voltage —  
 $V_{CE(sat)} = 2.0$  Vdc (Max) @  $I_C = 3.0$  Adc  
 = 4.0 Vdc (Max) @  $I_C = 5.0$  Adc
- Monolithic Construction with Built-In Base-Emitter Shunt Resistors
- TO-220AB Compact Package

**\*MAXIMUM RATINGS**

Rating	Symbol	TIP120, TIP125	TIP121, TIP126	TIP122, TIP127	Unit
Collector-Emitter Voltage	$V_{CE0}$	60	80	100	Vdc
Collector-Base Voltage	$V_{CB}$	60	80	100	Vdc
Emitter-Base Voltage	$V_{EB}$		5.0		Vdc
Collector Current — Continuous	$I_C$		5.0		A dc
Collector Current — Peak			8.0		
Base Current	$I_B$		120		mAdc
Total Power Dissipation @ $T_C = 25^\circ C$	$P_D$		65		Watts
Derate above $25^\circ C$			0.52		W/ $^\circ C$
Total Power Dissipation @ $T_A = 25^\circ C$	$P_D$		2.0		Watts
Derate above $25^\circ C$			0.016		W/ $^\circ C$
Unclamped Inductive Load Energy (1)	$E$		50		mJ
Operating and Storage Junction, Temperature Range	$T_J, T_{stg}$		-65 to +150		$^\circ C$

**THERMAL CHARACTERISTICS**

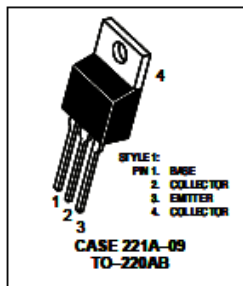
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	1.92	$^\circ C/W$
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	62.5	$^\circ C/W$

(1)  $I_C = 1$  A,  $L = 100$  mH, P.R.F. = 10 Hz,  $V_{CC} = 20$  V,  $R_{\theta E} = 100 \Omega$

- NPN**
- TIP120\***
- TIP121\***
- TIP122\***
- PNP**
- TIP125\***
- TIP126\***
- TIP127\***

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**60-80-100 VOLTS**  
**65 WATTS**



TIP120 TIP121 TIP122 TIP125 TIP126 TIP127

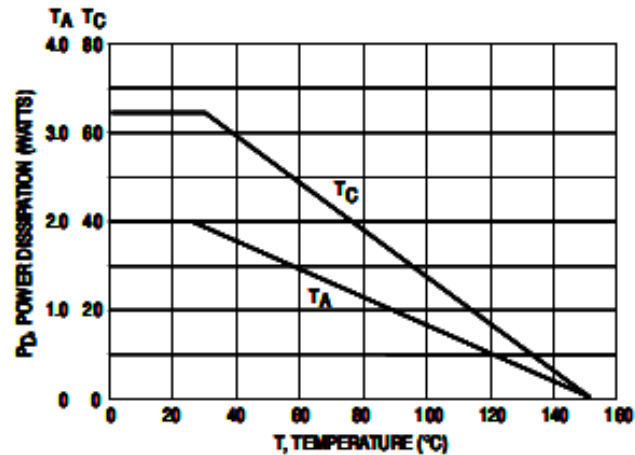


Figure 1. Power Derating

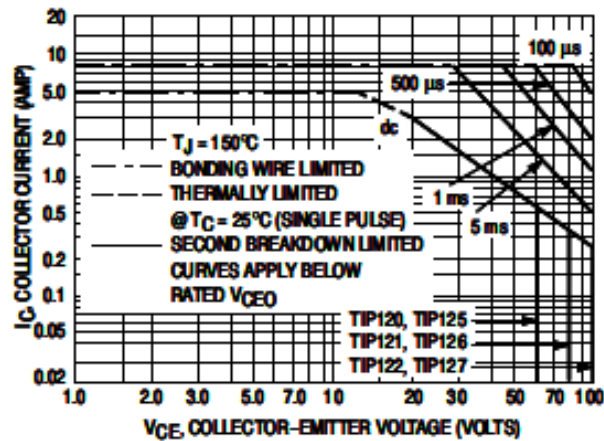


Figure 5. Active-Region Safe Operating Area

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate I<sub>C</sub> - V<sub>CE</sub> limits of the transistor that must be observed for reliable operation, i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figure 5 is based on T<sub>J(pk)</sub> = 150°C; T<sub>C</sub> is variable depending on conditions. Second breakdown pulse limits are valid for duty cycles to 10% provided T<sub>J(pk)</sub> < 150°C. T<sub>J(pk)</sub> may be calculated from the data in Figure 4. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.



**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
<b>OFF CHARACTERISTICS</b>				
Collector–Emitter Sustaining Voltage (1) ( $I_C = 100\text{ mAdc}$ , $I_B = 0$ )	TIP120, TIP125 TIP121, TIP126 TIP122, TIP127 $V_{CEO(sus)}$	60 80 100	— — —	Vdc
Collector Cutoff Current ( $V_{CE} = 30\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 40\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 50\text{ Vdc}$ , $I_B = 0$ )	TIP120, TIP125 TIP121, TIP126 TIP122, TIP127 $I_{CEO}$	— — —	0.5 0.5 0.5	mAdc
Collector Cutoff Current ( $V_{CB} = 60\text{ Vdc}$ , $I_E = 0$ ) ( $V_{CB} = 80\text{ Vdc}$ , $I_E = 0$ ) ( $V_{CB} = 100\text{ Vdc}$ , $I_E = 0$ )	TIP120, TIP125 TIP121, TIP126 TIP122, TIP127 $I_{CBO}$	— — —	0.2 0.2 0.2	mAdc
Emitter Cutoff Current ( $V_{BE} = 5.0\text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	—	2.0	mAdc
<b>ON CHARACTERISTICS (1)</b>				
DC Current Gain ( $I_C = 0.5\text{ Adc}$ , $V_{CE} = 3.0\text{ Vdc}$ ) ( $I_C = 3.0\text{ Adc}$ , $V_{CE} = 3.0\text{ Vdc}$ )	$h_{FE}$	1000 1000	— —	—
Collector–Emitter Saturation Voltage ( $I_C = 3.0\text{ Adc}$ , $I_B = 12\text{ mAdc}$ ) ( $I_C = 5.0\text{ Adc}$ , $I_B = 20\text{ mAdc}$ )	$V_{CE(sat)}$	— —	2.0 4.0	Vdc
Base–Emitter On Voltage ( $I_C = 3.0\text{ Adc}$ , $V_{CE} = 3.0\text{ Vdc}$ )	$V_{BE(on)}$	—	2.5	Vdc
<b>DYNAMIC CHARACTERISTICS</b>				
Small–Signal Current Gain ( $I_C = 3.0\text{ Adc}$ , $V_{CE} = 4.0\text{ Vdc}$ , $f = 1.0\text{ MHz}$ )	$h_{fe}$	4.0	—	—
Output Capacitance ( $V_{CB} = 10\text{ Vdc}$ , $I_E = 0$ , $f = 0.1\text{ MHz}$ )	TIP125, TIP126, TIP127 TIP120, TIP121, TIP122 $C_{ob}$	— —	300 200	pF

(1) Pulse Test: Pulse Width  $\leq 300\ \mu\text{s}$ , Duty Cycle  $\leq 2\%$ .

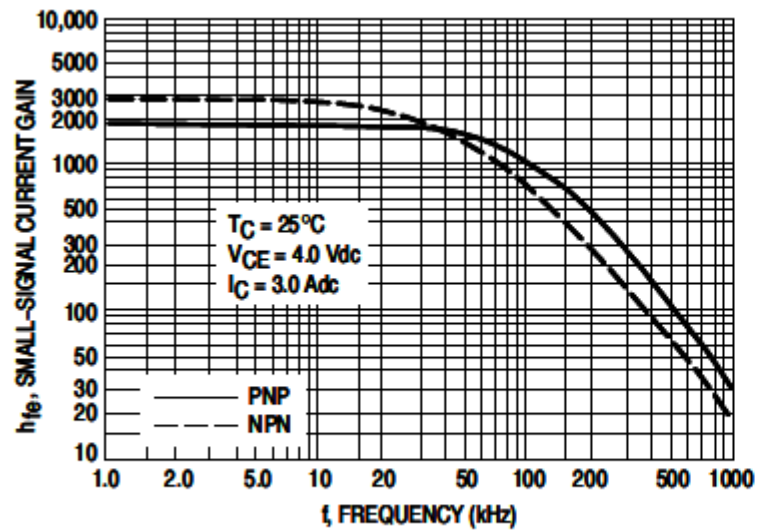


Figure 6. Small-Signal Current Gain

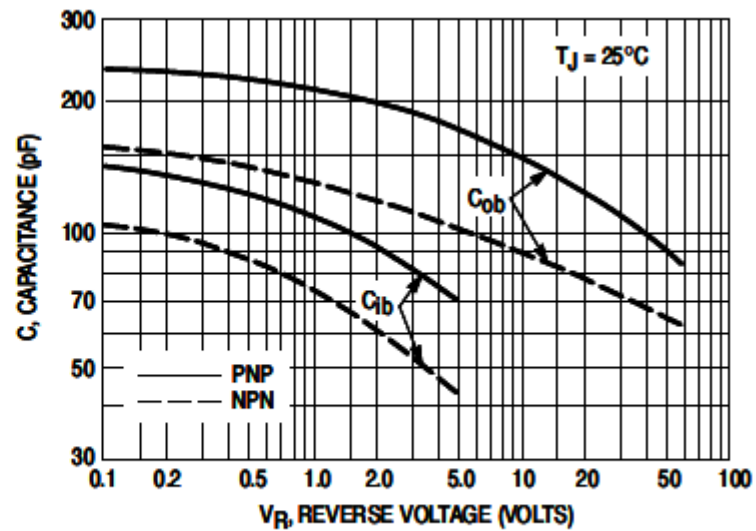


Figure 7. Capacitance

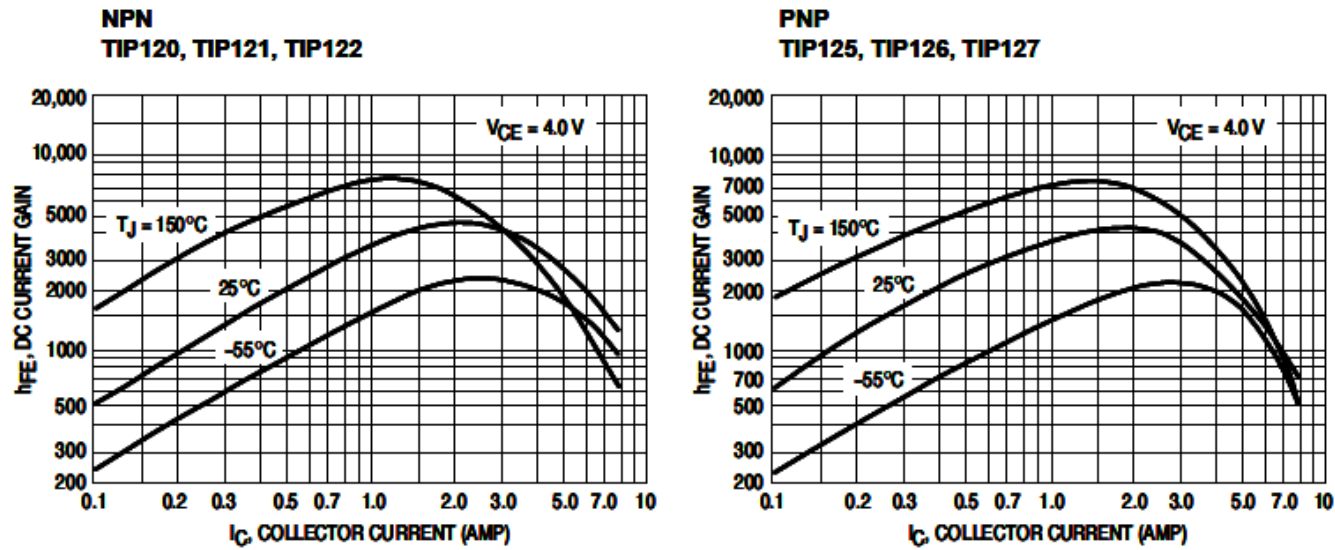


Figure 8. DC Current Gain

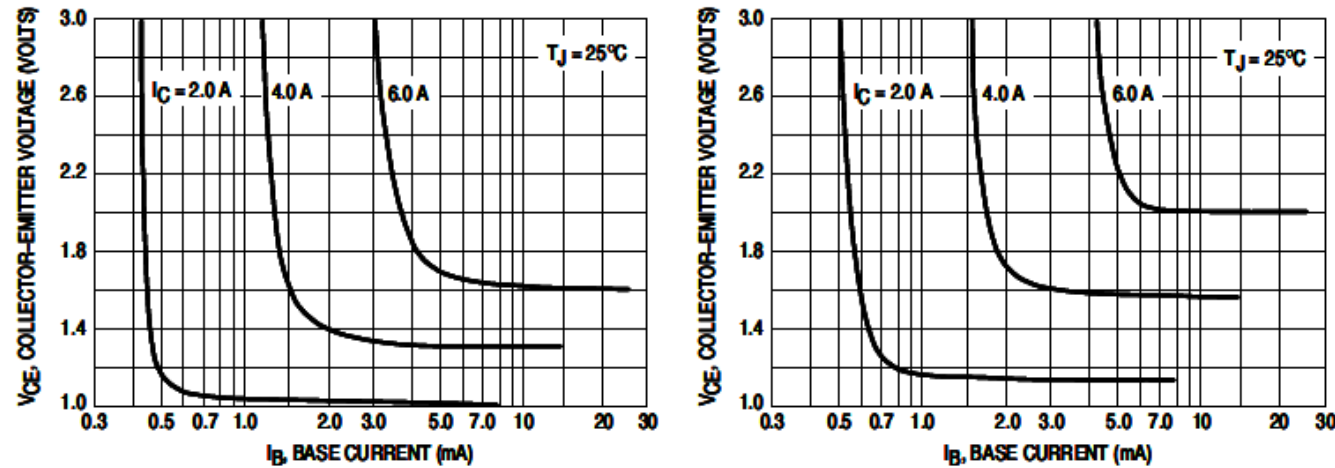


Figure 9. Collector Saturation Region

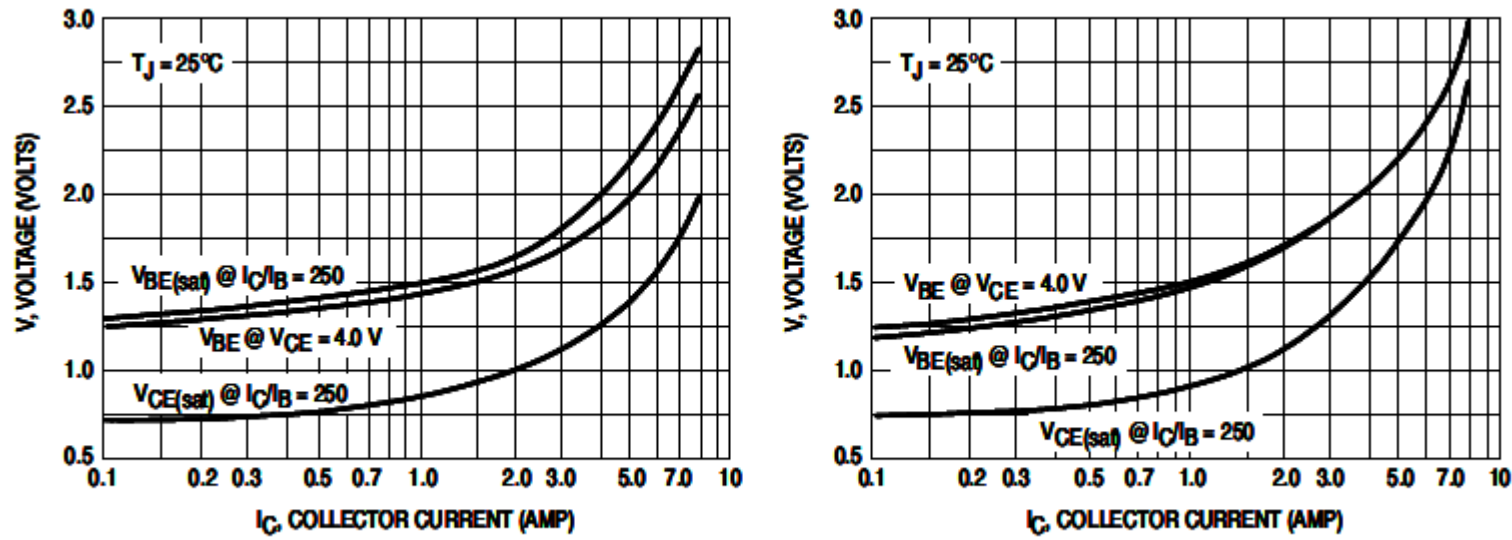


Figure 10. "On" Voltages

TIP140/141/142 NPN TIP 145/46/47 PNP DARLINGTON 10A

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- High DC Current Gain — Min  $h_{FE} = 1000$  @  $I_C = 5$  A,  $V_{CE} = 4$  V
- Collector-Emitter Sustaining Voltage — @ 30 mA  
 $V_{CEO(sus)} = 60$  Vdc (Min) — TIP140, TIP145  
 80 Vdc (Min) — TIP141, TIP146  
 100 Vdc (Min) — TIP142, TIP147
- Monolithic Construction with Built-In Base-Emitter Shunt Resistor

**MAXIMUM RATINGS**

Rating	Symbol	TIP140 TIP145	TIP141 TIP146	TIP142 TIP147	Unit
Collector-Emitter Voltage	$V_{CEO}$	60	80	100	Vdc
Collector-Base Voltage	$V_{CB}$	60	80	100	Vdc
Emitter-Base Voltage	$V_{EB}$	5.0			Vdc
Collector Current — Continuous Peak (1)	$I_C$	10 15			Adc
Base Current — Continuous	$I_B$	0.5			Adc
Total Device Dissipation @ $T_C = 25^\circ\text{C}$	$P_D$	125			Watts
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-65 to +150			$^\circ\text{C}$

**THERMAL CHARACTERISTICS**

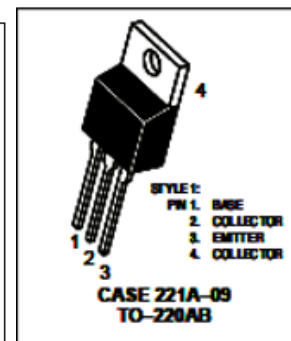
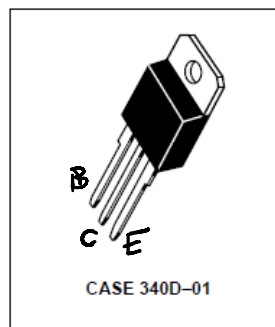
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	1.0	$^\circ\text{C/W}$
Thermal Resistance, Case to Ambient	$R_{\theta JA}$	35.7	$^\circ\text{C/W}$

(1) 5 ms,  $\leq$  10% Duty Cycle.

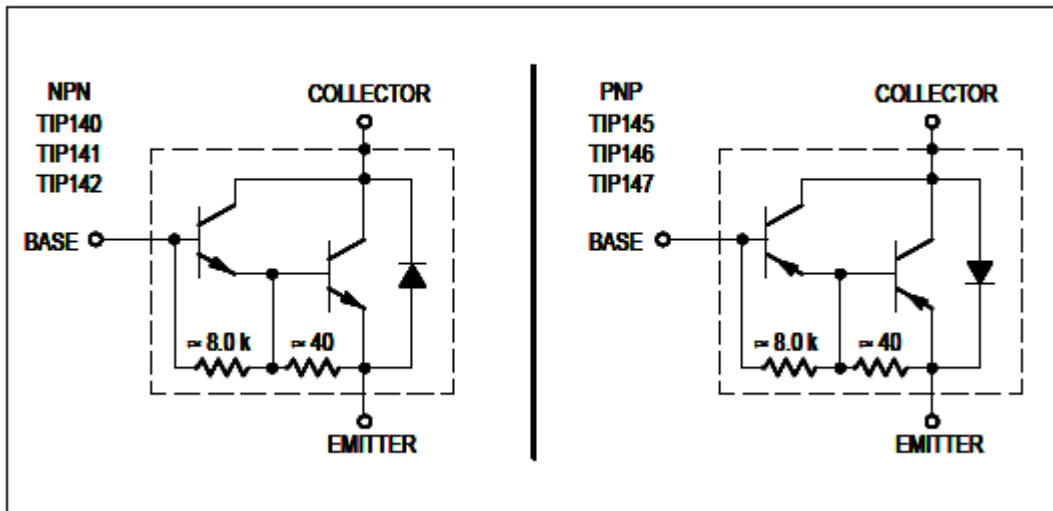
**NPN**  
**TIP140**  
**TIP141\***  
**TIP142\***  
**PNP**  
**TIP145**  
**TIP146\***  
**TIP147\***

\*Motorola Preferred Device

**10 AMPERE**  
**DARLINGTON**  
**COMPLEMENTARY SILICON**  
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**60-100 VOLTS**  
**125 WATTS**



**DARLINGTON SCHEMATICS**



**TIP140 TIP141 TIP142 TIP145 TIP146 TIP147**

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
----------------	--------	-----	-----	-----	------

**OFF CHARACTERISTICS**

Collector–Emitter Sustaining Voltage (1) ( $I_C = 30\text{ mA}$ , $I_B = 0$ )	TIP140, TIP145 TIP141, TIP146 TIP142, TIP147	$V_{CE(sus)}$	60 80 100	— — —	— — —	Vdc
Collector Cutoff Current ( $V_{CE} = 30\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 40\text{ Vdc}$ , $I_B = 0$ ) ( $V_{CE} = 50\text{ Vdc}$ , $I_B = 0$ )	TIP140, TIP145 TIP141, TIP146 TIP142, TIP147	$I_{CEO}$	— — —	— — —	2.0 2.0 2.0	mA
Collector Cutoff Current ( $V_{CB} = 60\text{ V}$ , $I_E = 0$ ) ( $V_{CB} = 80\text{ V}$ , $I_E = 0$ ) ( $V_{CB} = 100\text{ V}$ , $I_E = 0$ )	TIP140, TIP145 TIP141, TIP146 TIP142, TIP147	$I_{CBO}$	— — —	— — —	1.0 1.0 1.0	mA
Emitter Cutoff Current ( $V_{BE} = 5.0\text{ V}$ )		$I_{EBO}$	—	—	2.0	mA

**ON CHARACTERISTICS (1)**

DC Current Gain ( $I_C = 5.0\text{ A}$ , $V_{CE} = 4.0\text{ V}$ ) ( $I_C = 10\text{ A}$ , $V_{CE} = 4.0\text{ V}$ )		$h_{FE}$	1000 500	— —	— —	—
Collector–Emitter Saturation Voltage ( $I_C = 5.0\text{ A}$ , $I_B = 10\text{ mA}$ ) ( $I_C = 10\text{ A}$ , $I_B = 40\text{ mA}$ )		$V_{CE(sat)}$	— —	— —	2.0 3.0	Vdc
Base–Emitter Saturation Voltage ( $I_C = 10\text{ A}$ , $I_B = 40\text{ mA}$ )		$V_{BE(sat)}$	—	—	3.5	Vdc
Base–Emitter On Voltage ( $I_C = 10\text{ A}$ , $V_{CE} = 4.0\text{ Vdc}$ )		$V_{BE(on)}$	—	—	3.0	Vdc

**SWITCHING CHARACTERISTICS**

Resistive Load (See Figure 1)						
Delay Time	$(V_{CC} = 30\text{ V}$ , $I_C = 5.0\text{ A}$ , $I_B = 20\text{ mA}$ , Duty Cycle $\leq 2.0\%$ , $I_{B1} = I_{B2}$ , $R_C$ & $R_B$ Varied, $T_J = 25^\circ\text{C}$ )	$t_d$	—	0.15	—	$\mu\text{s}$
Rise Time		$t_r$	—	0.55	—	$\mu\text{s}$
Storage Time		$t_s$	—	2.5	—	$\mu\text{s}$
Fall Time		$t_f$	—	2.5	—	$\mu\text{s}$

(1) Pulse Test: Pulse Width = 300  $\mu\text{s}$ , Duty Cycle  $\leq 2.0\%$ .

**TIP140 TIP141 TIP142 TIP145 TIP146 TIP147**

**TYPICAL CHARACTERISTICS**

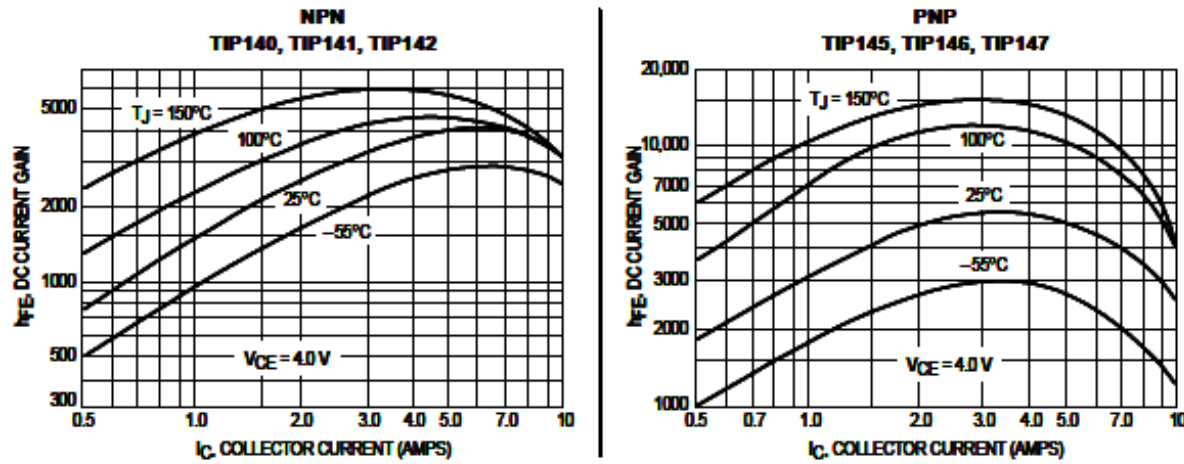


Figure 3. DC Current Gain versus Collector Current

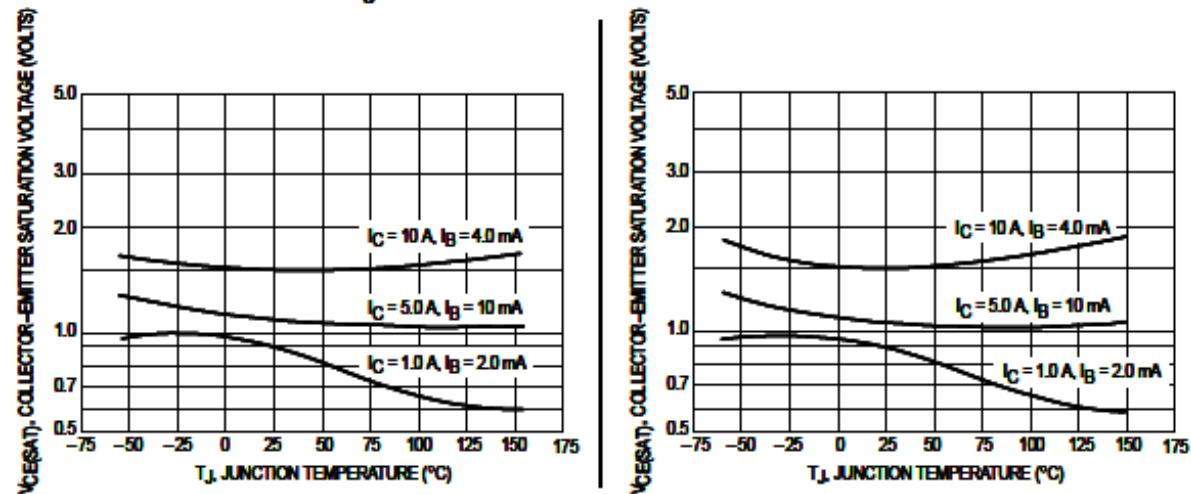


Figure 4. Collector–Emitter Saturation Voltage



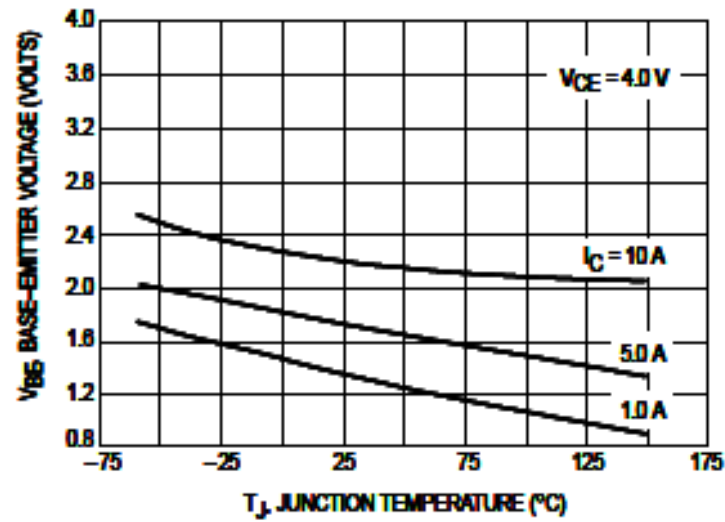
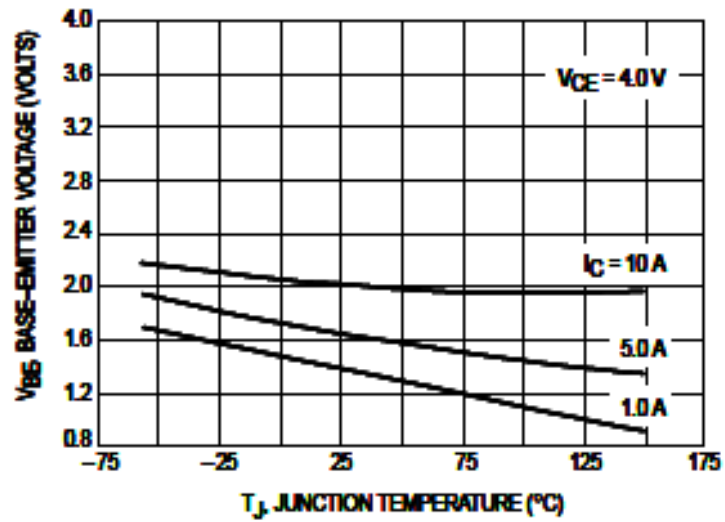


Figure 5. Base-Emitter Voltage

**TIP140 TIP141 TIP142 TIP145 TIP146 TIP147**

**ACTIVE-REGION SAFE OPERATING AREA**

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate  $I_C - V_{CE}$  limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figure 6 is based on  $T_{J(pk)} = 150^\circ\text{C}$ ;  $T_C$  is variable depending on conditions. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

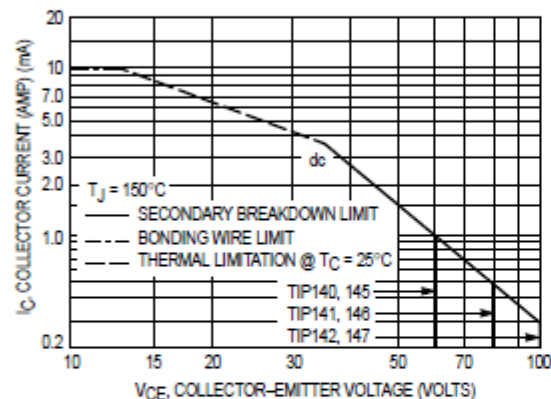


Figure 6. Active-Region Safe Operating Area

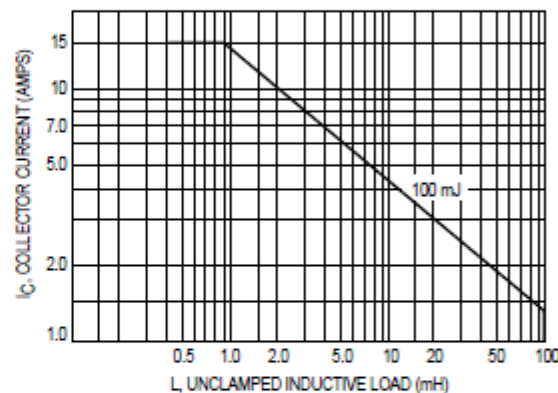
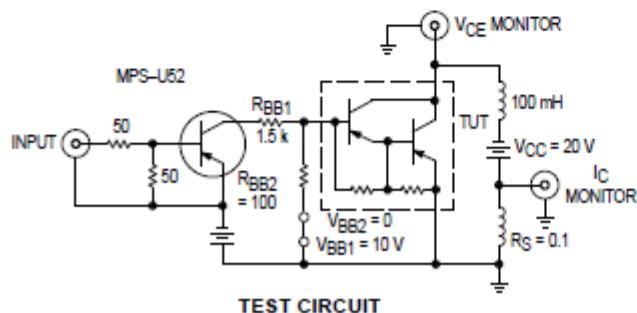
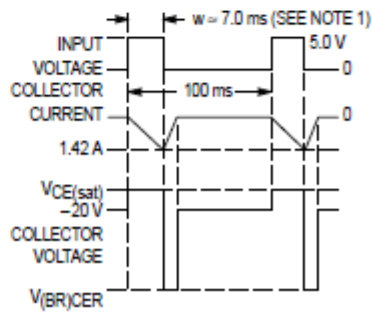


Figure 7. Unclamped Inductive Load



TEST CIRCUIT

NOTE 1: Input pulse width is increased until  $I_{CM} = 1.42\text{ A}$ .  
NOTE 2: For NPN test circuit reverse polarities.



VOLTAGE AND CURRENT WAVEFORMS

Figure 8. Inductive Load

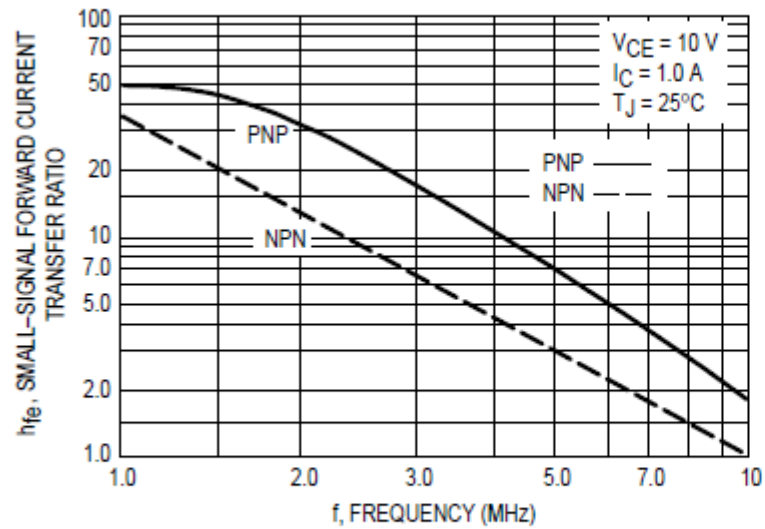


Figure 9. Magnitude of Common Emitter Small-Signal Short-Circuit Forward Current Transfer Ratio

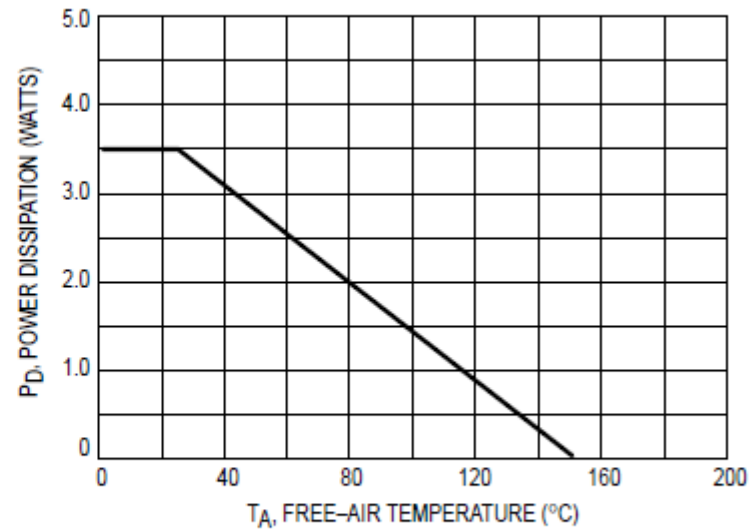
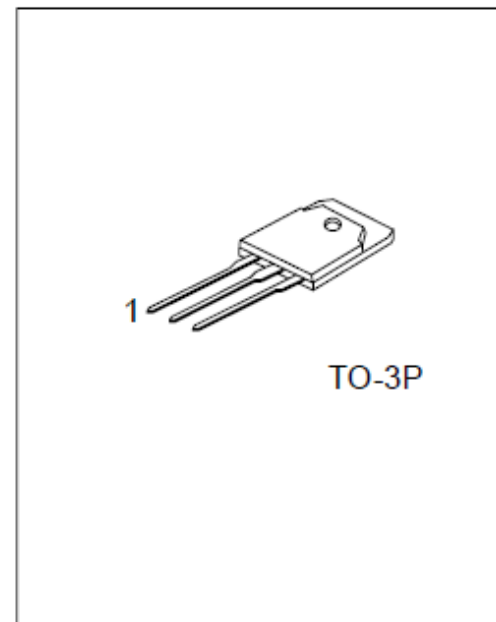
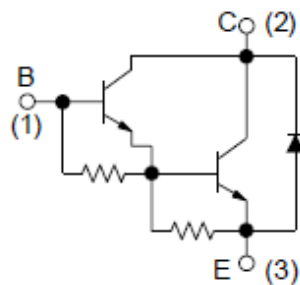


Figure 10. Free-Air Temperature Power Derating

BU931 DARLINGTON NPN 12A 175W


**UNISONIC TECHNOLOGIES CO., LTD**
**BU931**
*NPN SILICON TRANSISTOR*
**NPN POWER DARLINGTON**
**■ FEATURES**

- \* High operating junction temperature
- \* High voltage ignition coil driver
- \* Very rugged bipolar technology

**■ INTERNAL SCHEMATIC DIAGRAM**


TO-3P

\*Pb-free plating product number: BU931L

**BU931*****NPN SILICON TRANSISTOR*****■ ABSOLUTE MAXIMUM RATINGS** ( $T_a=25^\circ\text{C}$ )

PARAMETER	SYMBOL	RATINGS	UNIT
Collector-Emitter Voltage ( $V_{BE} = 0$ )	$V_{CES}$	500	V
Collector-Emitter Voltage ( $I_B = 0$ )	$V_{CEO}$	400	V
Emitter-Base Voltage ( $I_C = 0$ )	$V_{EBO}$	5	V
Collector Current	$I_C$	15	A
Collector Peak Current	$I_{CM}$	30	A
Base Current	$I_B$	1	A
Base Peak Current	$I_{BM}$	5	A
Total Dissipation ( $T_c = 25^\circ\text{C}$ )	$P_D$	175	W
Junction Temperature	$T_J$	+200	$^\circ\text{C}$
Storage Temperature	$T_{STG}$	-65 ~ +200	$^\circ\text{C}$

Note: Absolute maximum ratings are those values beyond which the device could be permanently damaged.  
Absolute maximum ratings are stress ratings only and functional device operation is not implied.

**■ THERMAL DATA**

PARAMETER	SYMBOL	RATING	UNIT
Thermal Resistance Junction-Case Max	$\theta_{JC}$	1.1	$^\circ\text{C/W}$

### ■ ELECTRICAL CHARACTERISTICS

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Collector Cut-off Current ( $I_B = 0$ )	$I_{CEO}$	$V_{CE} = 450 \text{ V}$			100	$\mu\text{A}$
		$V_{CE} = 450 \text{ V}, T_J = 125 \text{ }^\circ\text{C}$			0.5	$\text{mA}$
Emitter Cut-off Current ( $I_C = 0$ )	$I_{EBO}$	$V_{EB} = 5 \text{ V}$			20	$\text{mA}$
Collector-Emitter Saturation Voltage(Note)	$V_{CE(SAT)}$	$I_C = 7 \text{ A}, I_B = 70 \text{ mA}$			1.6	$\text{V}$
		$I_C = 8 \text{ A}, I_B = 100 \text{ mA}$			1.8	$\text{V}$
		$I_C = 10 \text{ A}, I_B = 250 \text{ mA}$			1.8	$\text{V}$
Base-Emitter Saturation Voltage(Note)	$V_{BE(SAT)}$	$I_C = 7 \text{ A}, I_B = 70 \text{ mA}$			2.2	$\text{V}$
		$I_C = 8 \text{ A}, I_B = 100 \text{ mA}$			2.4	$\text{V}$
		$I_C = 10 \text{ A}, I_B = 250 \text{ mA}$			2.5	$\text{V}$
DC Current Gain	$h_{FE}$	$I_C = 5 \text{ A}, V_{CE} = 10 \text{ V}$	300			
Diode Forward Voltage	$V_F$	$I_F = 10 \text{ A}$			2.5	$\text{V}$
Functional Test		$V_{CC} = 24 \text{ V}, V_{clamp} = 400 \text{ V}$ $L = 7 \text{ mH}$	8			$\text{A}$
Inductive Load Storage Time / Fall Time	$t_S$	$V_{CC} = 12 \text{ V}, V_{clamp} = 300 \text{ V}$ $L = 7 \text{ mH}$		15		$\mu\text{s}$
	$t_F$	$I_C = 7 \text{ A}, I_B = 70 \text{ mA}$ $V_{BE} = 0, R_{BE} = 47\Omega$		0.5		$\mu\text{s}$

Note: Pulsed: Pulse duration = 300  $\mu\text{s}$ , duty cycle 1.5 %

TRANSITORES DE GERMÂNIO.

TABELA DE TRANSISTORES DE GERMÂNIO PARA PEQUENOS SINAIS

**GERMANIUM SMALL SIGNAL TRANSISTORS**  
PRO ELECTRON TYPES

Type	Polar- ity	$V_{CB0}$ V Max	$V_{EB0}$ V Max	$V_{CE}$ V Max		$I_{CO}$ $\mu A$ Max	$R_{EX}$ Min Max	$\beta_{IC}$ m.A	$C_{ob}$ pF Max	$f_{ob}$ MHz Min	Pack Outline	Power Dissipation @ 25°C mW	
AC107	P	15											
AC116	P	30	12	18		6	8	35-50-140	300 <sup>2</sup> 20	21 <sup>2</sup>	TO-1	80	
AC117	P	32	10	18		6	18	40-	150		NS257 NS257	145 260	
AC121-IV	P	20	10	20				30-60	100	40	1.5 <sup>2</sup>	TO-1	900
AC121-V	P	20	10	20				50-100	100	40	1.5 <sup>2</sup>	TO-1	900
AC121-VI	P	20	10	20				75-150	100	40	1.5 <sup>2</sup>	TO-1	900
AC121-VII	P	20	10	20				125-250	100	40	1.5 <sup>2</sup>	TO-1	900
AC122	P	30	18	18		6	8	40-200	2 <sup>1</sup>	21 <sup>2</sup>			
AC122/30	P	45	12	32		6	8	40-200	2 <sup>1</sup>	21 <sup>2</sup>	TO-1	130	
AC123	P	45	12	32		6	8	50-140	20	21 <sup>2</sup>			
AC124	P	45	10	32		6	18	40-170	150		NS257	145	
AC125	P	32	10	12		10	10	50-	2	50	NS257 TO-1	260 500	
AC126	P	32	10	12		10	10	65-	2	50	TO-1	500	
AC127	N	32	10	12				50-	500	70 <sup>2</sup>	TO-1	340	
AC128	P	32	10	16		10	10	55-175	50	100	TO-1	1,000	
AC128K	P	32	10	16		10	10	55-175	50	100	NS257	1,000	
AC130	N	20									2.0	TO-1	
AC131	P	30	10	18		6	18	40-	150		TO-1	212	
AC132	P	32	10	12		0.5	10	135 <sup>2</sup>	20	40 <sup>2</sup>	TO-1	500	
AC138	P	32	10	20		10	15	30-	5 <sup>1</sup>		TO-1	720	
AC139	P	32	10	20		10	15	40-160	400		TO-1	720	
AC141	N	32	10	18		10	35	40-160	400		TO-1	720	
AC142	P	32	10	20		10	15	40-160	400		TO-1	720	
AC142K	P	32	10	20		10	15	40-160	400		NS257	860	
AC151	P	32	10	2Y		10	10	30 <sup>1</sup> -	2	27 <sup>2</sup>	TO-1	900	
AC152	P	32	10	2Y		0.5	10	30-150	100	40	1.5 <sup>2</sup>	TO-1	900
AC153	P	32	10	18		10	10	50-250	300	100	TO-1	1,000	
AC153K	P	32	10	18		10	10	50-250	300	100	NS257	1,000	
AC162	P	32	10	2Y		10	10	100 <sup>2</sup>	50	40	TO-1	900	
AC163	P	32	10	2Y		10	10	65-	2	40	TO-1	900	
AC173	P	32	10	2Y				50-	2 <sup>1</sup>		TO-1	300	
AC176	N	32	10	18		10	35	50-250	300	100 <sup>2</sup>	TO-1	700	
AC178	P	20	10	15		6	35	60-	150		NS257	180	
AC179	N	20	10	15		6	10 <sup>2</sup>	60-	150		NS257	180	
AC180	P	32	20	16		6		50-250	600		TO-1	600	



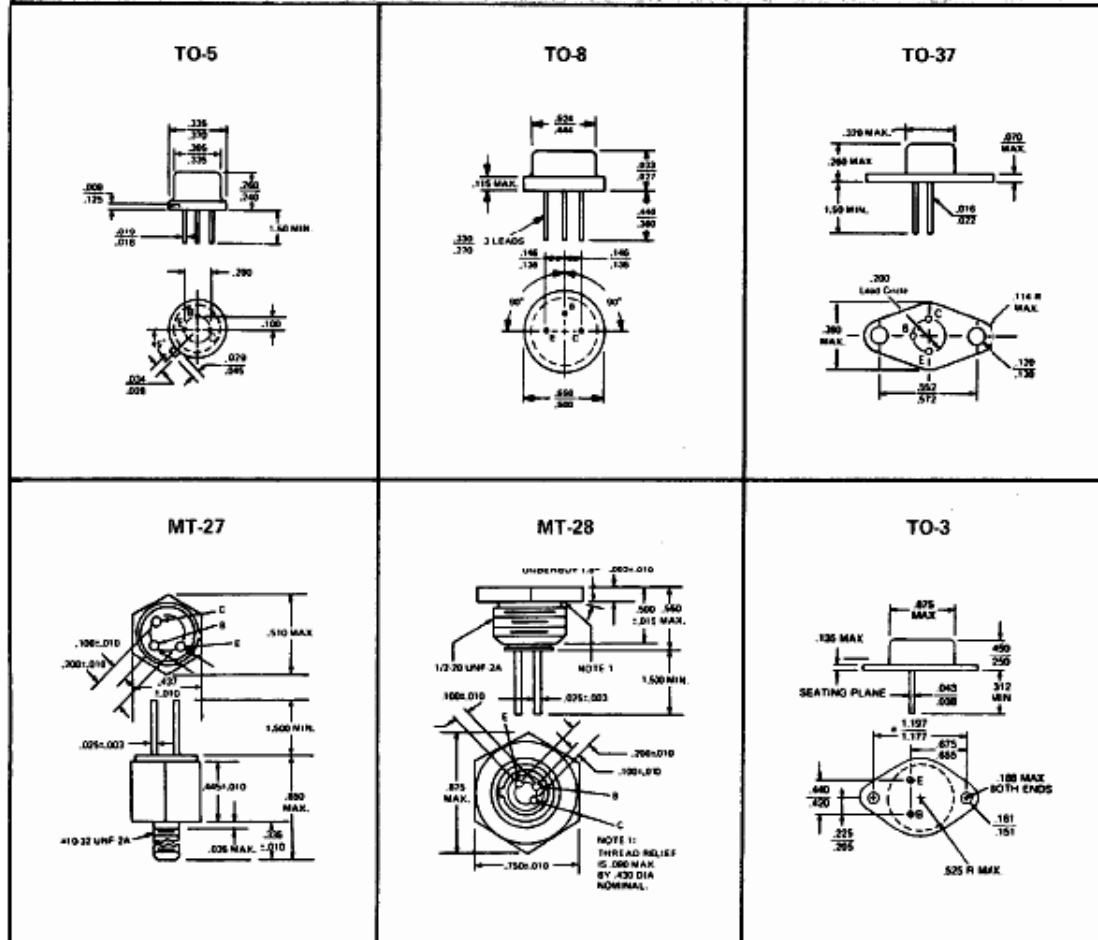
DICAS DE COMPONENTES DO PROFESSOR BAIROS

434

AC180K	P	32	20	16				50-250	600		2.5 <sup>2</sup>	NS257	2,500
AC181	N	32	20	16				50-250	600		4.5 <sup>2</sup>	TO-1	600
AC181K	N	32	20	16				50-250	600		4.5 <sup>2</sup>	NS257	2,500
AC182	P	32	20	18				50-	1'		4.0 <sup>2</sup>	TO-1	200
AC183	N	32	20	16				50-	2'		4.0 <sup>2</sup>	TO-1	250
AC184	P	32	20	16				50-250	300		2.5 <sup>2</sup>	TO-1	600
AC185	N	32	20	16				50-250	300		2.5 <sup>2</sup>	TO-1	600
AC187	N	25	10	15	10	35	100-500	300	180	1.0	TO-1	1,000	
AC187K	N	25	10	15	10	35	100-500	300	180	1.0	NS257	1,000	
AC188	P	25	10	15	10	15	100-500	300	110	1.0	TO-1	1,000	
AC188K	P	25	10	15	10	15	100-500	300	100	1.0	NS257	1,000	
AC193	P	32	10	15	10	15	90-400	400	40 <sup>2</sup>	3.0 <sup>2</sup>	TO-1	220	
AC193K	P	32	10	15	10	15	90-400	400	40 <sup>2</sup>	3.0 <sup>2</sup>	TO-1	260	
AC194	N	32	10	15	10	35	90-400	400	80 <sup>2</sup>	5.0 <sup>2</sup>	TO-1	220	
AC194K	N	32	10	15	10	35	90-400	400	80 <sup>2</sup>	5.0 <sup>2</sup>	TO-1	260	
ACY11	P	32	16	30		5	38-	10	35	0.4	TO-1	150	
ACY14	P	32	16	30		5	54-	10	35	0.4	TO-1	150	
ACY17		70	12	32		6	50-150	300	40	1.0	TO-5	200	
ACY18		50	12	30		6	40-120	300	40	1.0	TO-5	200	
ACY19		50	12	30		6	80-250	300	40	1.3	TO-5	200	
ACY20		40	12	20		6	50-145	50	40	1.0	TO-5	200	
ACY21		40	12	20		6	90-250	50	40	1.3	TO-5	200	
ACY22		20	12	15		6	30-300	300	40	1.0	TO-5	200	
ACY23	P	32	16	30			50-	1'		0.5	TO-1	900	
ACY27	P	40	30	20	30	12	20-55	1'	40	1.0 <sup>2</sup>	TO-1	200	
ACY28	P	40	30	15	30	12	45-150	1'	40	1.0 <sup>2</sup>	TO-1	200	
ACY29	P	40	30	15	30	12	45-150	1'	40	1.0 <sup>2</sup>	TO-1	200	
ACY30	P	40	40	20	30	12	60-200	1'	40	1.0 <sup>2</sup>	TO-1	200	
ACY31	P	40	20		12	5	35-70	1'	40	1.0 <sup>2</sup>	TO-1	200	
ACY32	P	32	16	30	10	10	50-150	1'	27 <sup>2</sup>	0.5	TO-1	900	

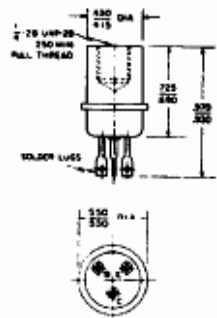
GERMANIUM POWER TRANSISTORS

CASE OUTLINE DRAWINGS & DIMENSIONS

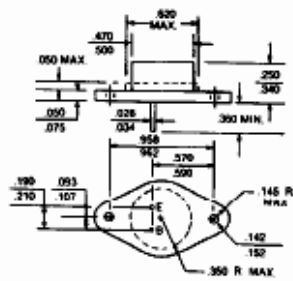


### CASE OUTLINE DRAWINGS & DIMENSIONS

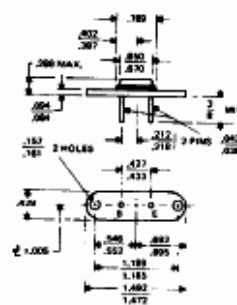
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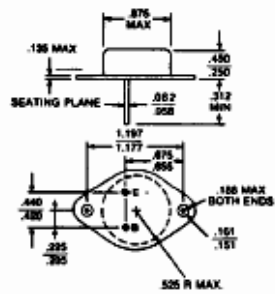
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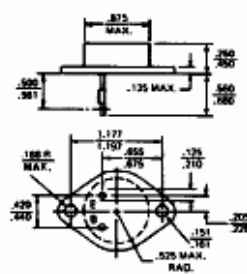
MS-7



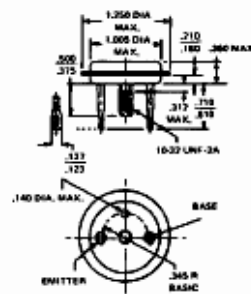
MODIFIED TO-3  
(60 mil pins)



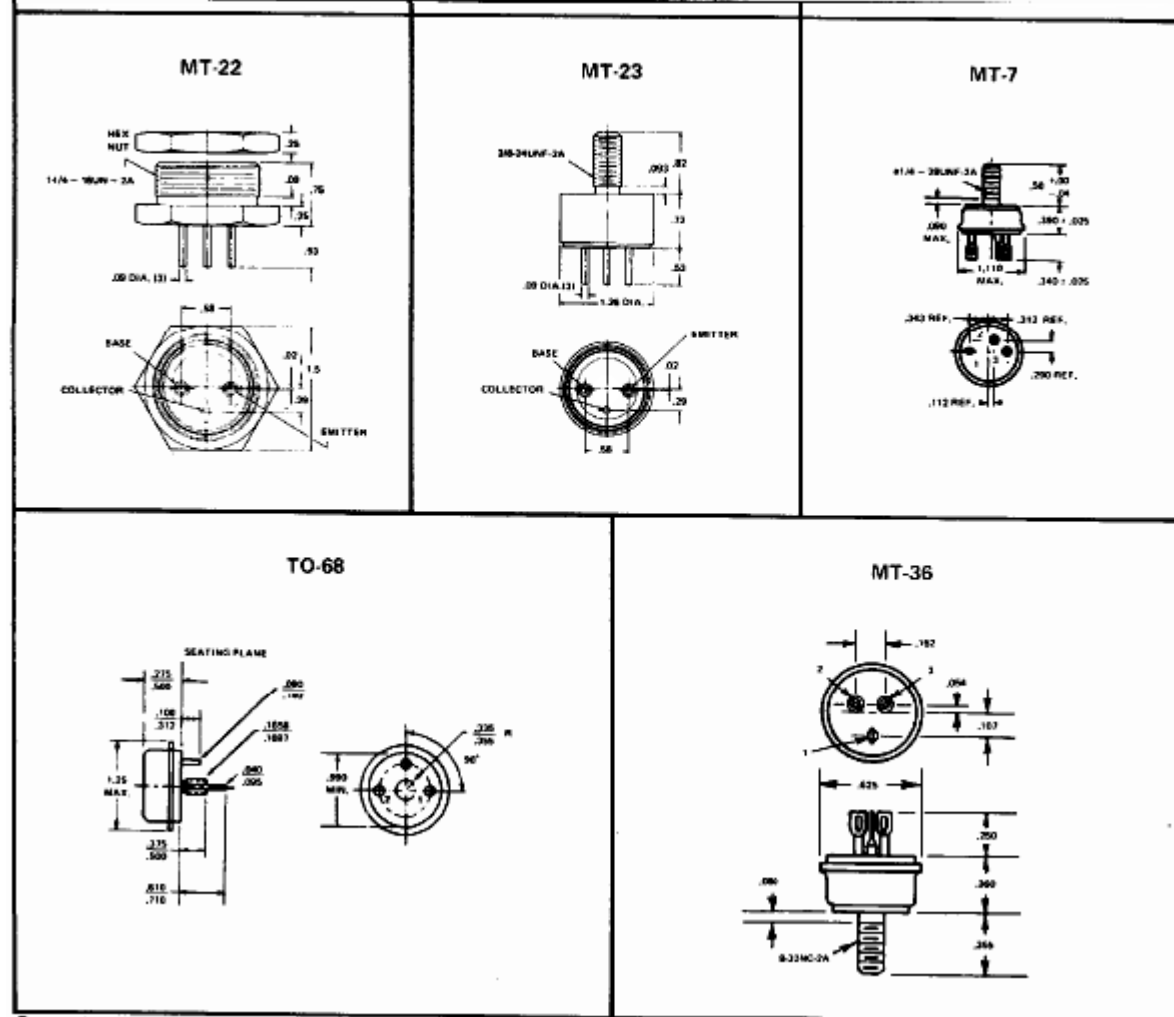
TO-41



TO-36



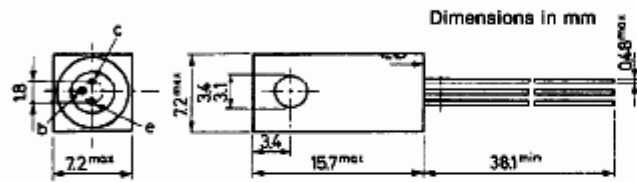
### CASE OUTLINE DRAWINGS & DIMENSIONS



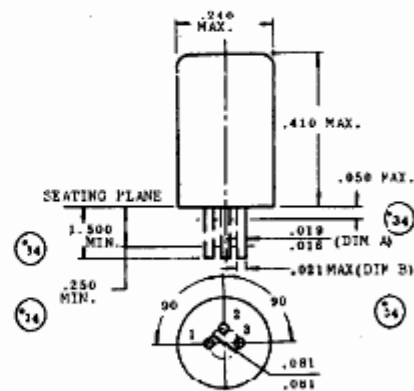
Germanium Power Devices Corporation

### CASE OUTLINE DRAWINGS & DIMENSIONS

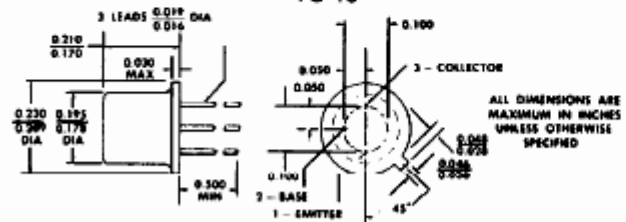
NS257



TO-1



TO-18



THE COLLECTOR IS ELECTRICAL CONTACT WITH THE CASE.

ALL JEDEC TO-18 DIMENSIONS AND NOTES ARE APPLICABLE.

## LENDO CÓDIGOS E MARCAS EM TRANSISTORES

A maioria das indicações dos transistores segue um dos códigos a seguir: JEDEC, JIS ou Pro-Electron.

Para CI's, aparece com números (Por ex. 741, 4001, 7400) entre o prefixo e o sufixo.

Joint Electron Device Engineering Council (JEDEC):

Estes números estão na forma de dígitos, letras e números sequenciais [sufixo]. A letra é sempre 'N', e o primeiro dígito é 1 para diodos, 2 para transistores, 3 para dispositivos de 4 camadas e assim por diante. Mas 4N e 5N são reservados para acopladores óticos. Os números sequenciais começam em 100 até 9999.

Se presente o sufixo pode indicar várias coisas. Por exemplo 2N2222A é uma versão melhorada do 2N2222. Ele tem maior ganho, frequência, limites de tensão. Na dúvida sempre checar uma Data Sheet.

Exemplos: 1N914 (diode), 2N2222, 2N2222A, 2N904 (transistors).

NOTA: Quando uma versão metálica de um transistor JEDEC é refeita em encapsulamento plástico, muitas vezes é adicionado um número ou letra. Por exemplo o transistor PN2222A é uma versão em plástico do 2N2222A. (metálico)

Obs: Ao invés de 2N muitas vezes o fabricante usa a sua própria designação.

Japanese Industrial Standard (JIS):

Os números da parte tomam a forma: Dígito, duas letras, número sequencial (sufixo opcional)T

Dígito 1 é para diodos 2 para transistores. A letra indica o tipo e a aplicação do dispositivo de acordo com o seguinte código:

Exemplo o transistor PN2222A é uma versão em plástico do 2N2222A. (metálico)

SA:	PNP transistor para HF	SB:	PNP AF transistor
SC:	NPN transistor para HF	SD:	NPN AF transistor
SE:	Diodos	SF:	Thyristors
SG:	Gunn devices	SH:	UJT
SJ:	P-channel FET	SK:	N-channel FET
SM:	Triac	SQ:	LED
SR:	Rectifier	SS:	Signal diodes
ST:	Avalanche diodes	SV:	Varicaps
SZ:	Zener diodes		

O número sequencial começa em 10 até 9999. O sufixo, opcional, indica a aprovação por várias organizações Japonesas. Como o código sempre começa com 2S, muitas vezes é omitido, por exemplo, um 2SC733 pode ser encontrado como C733.

Exemplos: 2SA1187, 2SB646, 2SC733.

Pro-Electron (European):

A parte do código é formado por duas letras e número sequencial.

A primeira letra indica o tipo de material: A = Ge    B = Si    C = GaAs    R = Material composto

A segunda letra indica o tipo de dispositivo e aplicação:

A: diodo, RF

B: diodo, varactor

C: transistor, AF, pequeno sinal

D: transistor, AF, potência

E: diodo Tunnel

F: transistor, HF, pequeno sinal

K: Dispositivo de efeito Hall

L: Transistor, HF, potência

N: Opto-Acoplador

P: Dispositivo sensível à radiação

Q: Dispositivo que produz radiação

R: Tiristor, baixa potência

T: Tiristor, potência

U: Transistor, potencia, chaveamento

Y: Retificador

Z: diodo Zener ou regulador de tensão



A terceira letra indica se o dispositivo é usado em aplicação industrial ou comercial, é usualmente W, X, Y, ou Z.

Exemplos: BC108A, BAW68, BF239, BFY51.

Os prefixos mais comuns são:

MJ: Motorola potência, encapsulamento metálico

MJE: Motorola potência, encapsulamento plástico

MPS: Motorola baixa potência, encapsulamento plástico

MRF: Motorola HF, VHF transistor para microondas

RCA: dispositivo RCA

TIP: Texas Instruments (TI) transistor de potência , encapsulamento plástico

TIPL: TI transistor planar de potência

TIS: TI transistor de pequeno sinal (encapsulamento plástico )

ZT: Ferranti

ZTX: Ferranti

DIODO

DIODO RETIFICADOR

1N4001 4002 4003 4004 4004 4005 4006 4007 DIODO RETIFICADOR 1A

## **1N4001, 1N4002, 1N4003, 1N4004, 1N4005, 1N4006, 1N4007**

### **Axial Lead Standard Recovery Rectifiers**

This data sheet provides information on subminiature size, axial lead mounted rectifiers for general-purpose low-power applications.

#### **Features**

- Shipped in Plastic Bags, 1000 per bag
- Available Tape and Reeled, 5000 per reel, by adding a "RL" suffix to the part number
- Available in Fan-Fold Packaging, 3000 per box, by adding a "FF" suffix to the part number
- Pb-Free Packages are Available

#### **Mechanical Characteristics**

- Case: Epoxy, Molded
- Weight: 0.4 gram (approximately)
- Finish: All External Surfaces Corrosion Resistant and Terminal Leads are Readily Solderable
- Lead and Mounting Surface Temperature for Soldering Purposes: 260°C Max. for 10 Seconds, 1/16 in. from case
- Polarity: Cathode Indicated by Polarity Band



**ON Semiconductor®**

<http://onsemi.com>

**LEAD MOUNTED RECTIFIERS  
50-1000 VOLTS  
DIFFUSED JUNCTION**



**CASE 59-10  
AXIAL LEAD  
PLASTIC**

**MAXIMUM RATINGS**

Rating	Symbol	1N4001	1N4002	1N4003	1N4004	1N4005	1N4006	1N4007	Unit
†Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	$V_{RRM}$ $V_{RWM}$ $V_R$	50	100	200	400	600	800	1000	V
†Non-Repetitive Peak Reverse Voltage (halfwave, single phase, 60 Hz)	$V_{RSM}$	60	120	240	480	720	1000	1200	V
†RMS Reverse Voltage	$V_{R(RMS)}$	35	70	140	280	420	560	700	V
†Average Rectified Forward Current (single phase, resistive load, 60 Hz, $T_A = 75^\circ\text{C}$ )	$I_O$	1.0							A
†Non-Repetitive Peak Surge Current (surge applied at rated load conditions)	$I_{FSM}$	30 (for 1 cycle)							A
Operating and Storage Junction Temperature Range	$T_J$ $T_{stg}$	-65 to +175							$^\circ\text{C}$

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

† Indicates JEDEC Registered Data

**THERMAL CHARACTERISTICS**

Rating	Symbol	Max	Unit
Maximum Thermal Resistance, Junction-to-Ambient	$R_{\theta JA}$	Note 1	$^\circ\text{C}/\text{W}$

**ELECTRICAL CHARACTERISTICS†**

Rating	Symbol	Typ	Max	Unit
Maximum Instantaneous Forward Voltage Drop, ( $i_F = 1.0$ Amp, $T_J = 25^\circ\text{C}$ )	$V_F$	0.93	1.1	V
Maximum Full-Cycle Average Forward Voltage Drop, ( $I_O = 1.0$ Amp, $T_L = 75^\circ\text{C}$ , 1 inch leads)	$V_{F(AV)}$	-	0.8	V
Maximum Reverse Current (rated DC voltage) ( $T_J = 25^\circ\text{C}$ ) ( $T_J = 100^\circ\text{C}$ )	$I_R$	0.05 1.0	10 50	$\mu\text{A}$
Maximum Full-Cycle Average Reverse Current, ( $I_O = 1.0$ Amp, $T_L = 75^\circ\text{C}$ , 1 inch leads)	$I_{R(AV)}$	-	30	$\mu\text{A}$

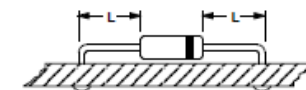
NOTE 1. - AMBIENT MOUNTING DATA

Data shown for thermal resistance, junction-to-ambient ( $R_{\theta JA}$ ) for the mountings shown is to be used as typical guideline values for preliminary engineering or in case the tie point temperature cannot be measured.

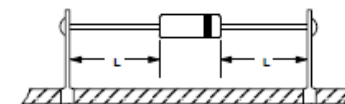
TYPICAL VALUES FOR  $R_{\theta JA}$  IN STILL AIR

Mounting Method	Lead Length, L			Units
	1/8	1/4	1/2	
1	52	65	72	$^\circ\text{C}/\text{W}$
2	67	80	87	$^\circ\text{C}/\text{W}$
3	50			$^\circ\text{C}/\text{W}$

MOUNTING METHOD 1

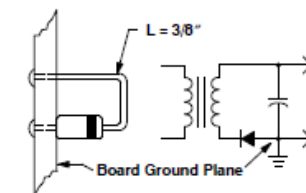


MOUNTING METHOD 2



Vector Pin Mounting

MOUNTING METHOD 3



P.C. Board with  
1-1/2" X 1-1/2" Copper Surface

1N 914 AB 916 A B 4148 4448 DIODO RÁPIDO DE USO GERAL 200MA

## Small Signal Diode

### Absolute Maximum Ratings\* T<sub>A</sub> = 25°C unless otherwise noted

Symbol	Parameter	Value	Units
V <sub>RRM</sub>	Maximum Repetitive Reverse Voltage	100	V
I <sub>F(AV)</sub>	Average Rectified Forward Current	200	mA
I <sub>FSM</sub>	Non-repetitive Peak Forward Surge Current		
	Pulse Width = 1.0 second	1.0	A
	Pulse Width = 1.0 microsecond	4.0	A
T <sub>stg</sub>	Storage Temperature Range	-65 to +200	°C
T <sub>J</sub>	Operating Junction Temperature	175	°C

### Thermal Characteristics

Symbol	Characteristic	Max	Units
		1N/FDLL 914/A/B / 4148 / 4448	
P <sub>D</sub>	Power Dissipation	500	mW
R <sub>θJA</sub>	Thermal Resistance, Junction to Ambient	300	°C/W

**Electrical Characteristics** $T_A = 25^\circ\text{C}$  unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Max	Units
$V_R$	Breakdown Voltage	$I_R = 100 \mu\text{A}$	100		V
		$I_R = 5.0 \mu\text{A}$	75		V
$V_F$	Forward Voltage	<b>1N914B/4448</b> <b>1N916B</b> $I_F = 5.0 \text{ mA}$	620	720	mV
		<b>1N914/916/4148</b> $I_F = 5.0 \text{ mA}$	630	730	mV
		<b>1N914A/916A</b> $I_F = 10 \text{ mA}$		1.0	V
		<b>1N916B</b> $I_F = 20 \text{ mA}$		1.0	V
		<b>1N914B/4448</b> $I_F = 100 \text{ mA}$		1.0	V
$I_R$	Reverse Current	$V_R = 20 \text{ V}$		25	nA
		$V_R = 20 \text{ V}, T_A = 150^\circ\text{C}$		50	$\mu\text{A}$
		$V_R = 75 \text{ V}$		5.0	$\mu\text{A}$
$C_T$	Total Capacitance	<b>1N916A/B/4448</b> $V_R = 0, f = 1.0 \text{ MHz}$		2.0	pF
		<b>1N914A/B/4148</b> $V_R = 0, f = 1.0 \text{ MHz}$		4.0	pF
$t_{rr}$	Reverse Recovery Time	$I_F = 10 \text{ mA}, V_R = 6.0 \text{ V (60mA)},$ $I_{rr} = 1.0 \text{ mA}, R_L = 100\Omega$		4.0	ns

---

DIODO DE SINAL, DIODO DE GERMÂNIO 1N270/1N34/1N60

1N270: diodo germânio otimizado demodulador detector  $V_f=1,00V$ ,  $V_r=100V$ ,  $I_F=40mA$ ,  $I_r=100\mu A$

1N34: diodo detector de RF Germânio bigode de gato

1N60: diodo detector de RF Germânio bigode de gato



DIODO 1N4151 52 53 54 ULTRA HIGH SPEED SILICON PLANAR EPITAXIAL DIODES

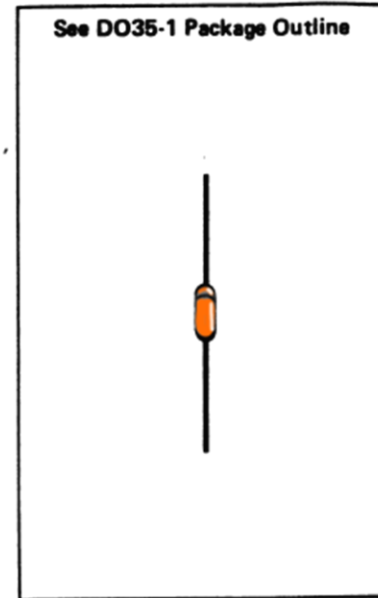
# 1N4151 • 1N4152 • 1N4153 • 1N4154

ULTRA HIGH SPEED  
SILICON PLANAR\* EPITAXIAL DIODES

- C . . . 4 pF @  $V_R = 0$ ,  $f = 1.0$  MHz
- $t_{rr}$  . . . 2.0 ns @  $I_F = 10$  mA,  $V_R = -6.0$  V,  $R_L = 100 \Omega$

**ABSOLUTE MAXIMUM RATINGS (Note 1)**

<b>Maximum Temperature</b>				
Storage Temperature				-65°C to +200°C
Lead Temperature (20 seconds)				300°C
<b>Maximum Power Dissipation (Note 2)</b>				
Total Dissipation at 25°C Ambient Temperature	500 mW	500 mW	500 mW	500 mW
Linear Derating Factor	2.85 mW/°C	2.85 mW/°C	2.85 mW/°C	2.85 mW/°C
<b>Maximum Voltage</b>				
$V_R$ Reverse Voltage	75 V	40 V	75 V	35 V



Mais rápido que o 1N4148, mas menor corrente!

DIODO RF

---

**DIODOS DE RF E BIGODE DE GATO 1N34/1N270/OA70/OA90/OA91/OA95**

1N34: diodo de contato de germânio detector de RF  $V_r=30V$ ,  $I_F=50mA$ ,  $I_r=2mA$

1N60: diodo de contato de germânio detector de RF  $V_r=50V$ ,  $I_F=500mA$ ,  $I_r=40\mu A$

OA70: diodo germânio demodulador detector  $V_r=15V$ ,  $I_F=50mA$

OA90: diodo germânio demodulador detector  $V_r=30V$ ,  $I_F=500mA$ ,  $V_f=1,2V$ ,  $I_r=80\mu A$

OA91: diodo germânio demodulador detector  $V_r=115V$ ,  $I_F=500mA$ ,  $V_f=1,02V$ ,  $I_r=80\mu A$

OA95: diodo germânio demodulador detector  $V_r=30V$ ,  $I_F=500mA$ ,  $V_f=1,05V$ ,  $I_r=30\mu A$

O diodo bigode de gato é um diodo sem junção por contato.

O termo velocidade se refere a velocidade de comutação, a velocidade de trocar do estado ligado para desligado quando a polaridade inverte, observe no data sheet o tempo de comutação é o parâmetro  $T_{rr}$  (reverse recovery time), conhecendo o tempo de recuperação você pode avaliar a frequência de trabalho simplesmente invertendo esse valor.

O parâmetro  $V_F$  (Forward Voltage) indica a tensão da junção, ao redor de  $0,7V$  indica um diodo de silício, ao redor de  $0,3$  indica um diodo de germânio.

Os diodos de junção possuem correntes reversas melhores, mas tensões diretas maiores.

Um diodo detector é usado em aplicações de rádio e TV, principalmente na etapa detectora de RF, esses são diodos de contato, são os melhores para trabalhar como detector.


Esses diodos rápidos são selados e encapsulados em vidro e vem normalmente no encapsulamento DO-35

## BA281 DIODO RF DETECTOR DE RF PHILIPS

**BA281**

**Diodo detector de FM - Philips.**

$V_R = 50 \text{ V}$   
 $I_F = 200 \text{ mA}$   
 $C_d \text{ a } (V_R = 0 \text{ V}) < 1,2 \text{ pF}$   
 $V_F = 360 - 420 \text{ mV (a } 0,01 \text{ mA)}$



## SILICON RATIO DETECTOR DIODE

Silicon planar epitaxial diode in DO-35 envelope, intended for use in ratio detector circuits. Due to small spreads of forward voltage at low currents and of junction capacitance, the diodes can be used as matched pairs.

## QUICK REFERENCE DATA

Continuous reverse voltage	$V_R$	max.	50 V
Forward current (d.c.)	$I_F$	max.	200 mA
Repetitive peak forward current	$I_{FRM}$	max.	450 mA
Forward voltage	$V_F$		360 to 420 mV
Diode capacitance	$C_d$	<	1,2 pF
Junction temperature	$T_j$	max.	200 °C

## MECHANICAL DATA

Fig. 1 DO-35 (SOD-27).

Dimensions in mm

## BA482 DIODO RF COMUTAÇÃO DE BANDA TV VHF PHILIPS E

**BA482**

Diodo de Comutação de banda - TV - VHF - Philips.

$$V_R = 35 \text{ V}$$

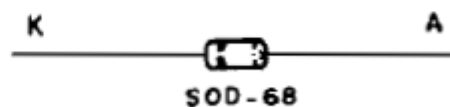
$$I_F = 100 \text{ mA}$$

$$C_d < 1,2 \text{ pF a } V_R = 3 \text{ V}$$

$$r_d < 0,7 \Omega$$

$$I_F = 3 \text{ mA}$$

$$V_F < 1,2 \text{ V}$$



**DIODO SCHOTTKY DE VIDRO BAIXA POTÊNCIA**

1N60/1N60P/BA317/BA318/BAT46

Um diodo schottky pequenos com encapsulamento de vidro são diodos que não tem junção, são diodos rápidos e baixas correntes reversas, são usados para chaveamento, são muito parecidos com os diodos bigode de gato, mas a são usados em aplicações de potência fontes chaveadas pequenas.

A tensão direta é muito baixa, da ordem de 0,3V.

---

**1N60/1N60P:**

Schottky de 80mW um dos mais populares. Diodo Schottky de baixa tensão reversa ao redor de 0,3V, tensão de trabalho 20V, corrente de trabalho de 150 mA podendo ser usado como detector. O diodo 1N60P é para corrente de trabalho de 500mA.

---

BA317/18:

Schoottky de 200 mA.



---

BAT46:

schottky de pequeno sinal de uso geral 100 mA.

DIODO RÁPIDO PARA FONTE CHAVEADA

TABELA SELEÇÃO DIODO RÁPIDO PARA FONTE CHAVEADA

Table 1. Diode Selection Guide

V <sub>R</sub>	Schottky				Fast Recovery			
	3.0 A		4.0 – 6.0 A		3.0 A		4.0 – 6.0 A	
	Through Hole	Surface Mount	Through Hole	Surface Mount	Through Hole	Surface Mount	Through Hole	Surface Mount
20 V	<b>1N5820</b> MBR320P SR302	SK32	<b>1N5823</b> SR502 SB520		MUR320 31DF1 HER302  (all diodes rated to at least 100 V)	<b>MURS320T3</b> MURD320 30WF10  (all diodes rated to at least 100 V)	MUR420 HER602  (all diodes rated to at least 100 V)	<b>MURD620CT</b> 50WF10  (all diodes rated to at least 100 V)
30 V	<b>1N5821</b> MBR330 SR303 31DQ03	SK33 30WQ03	<b>1N5824</b> SR503 SB530	50WQ03				
40 V	<b>1N5822</b> <b>MBR340</b> SR304 31DQ04	SK34 30WQ04 <b>MBRS340T3</b> <b>MBRD340</b>	<b>1N5825</b> SR504 SB540	<b>MBRD640CT</b> 50WQ04				
50 V	MBR350 31DQ05 SR305	SK35 30WQ05	SB550	50WQ05				
60 V	<b>MBR360</b> DQ06 SR306	<b>MBRS360T3</b> <b>MBRD360</b>	50SQ080	<b>MBRD660CT</b>				

NOTE: Diodes listed in bold are available from ON Semiconductor.

Fonte: manual do LM2576

UF4001 - UF4007 (1A)



## UF4001 – UF4007

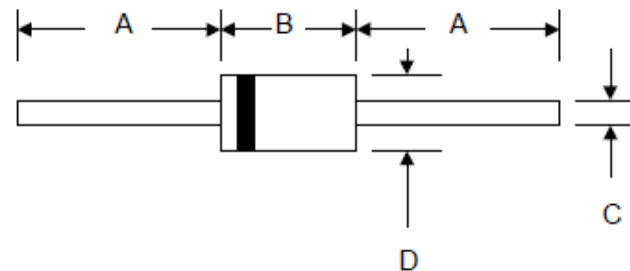
### 1.0A ULTRAFAST RECOVERY RECTIFIER

#### Features

- Diffused Junction
- Low Forward Voltage Drop
- High Current Capability
- High Reliability
- High Surge Current Capability

#### Mechanical Data

- Case: Molded Plastic
- Terminals: Plated Leads Solderable per MIL-STD-202, Method 208
- Polarity: Cathode Band
- Weight: 0.34 grams (approx.)
- Mounting Position: Any
- Marking: Type Number



DO-41		
Dim	Min	Max
A	25.4	—
B	4.06	5.21
C	0.71	0.864
D	2.00	2.72
All Dimensions in mm		

**Maximum Ratings and Electrical Characteristics** @ $T_A=25^{\circ}\text{C}$  unless otherwise specified

Single Phase, half wave, 60Hz, resistive or inductive load.

For capacitive load, derate current by 20%.

Characteristic	Symbol	UF 4001	UF 4002	UF 4003	UF 4004	UF 4005	UF 4006	UF 4007	Unit
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	VRRM VRWM VR	50	100	200	400	600	800	1000	V
RMS Reverse Voltage	VR(RMS)	35	70	140	280	420	560	700	V
Average Rectified Output Current (Note 1) @ $T_A = 55^{\circ}\text{C}$	I <sub>o</sub>	1.0							A
Non-Repetitive Peak Forward Surge Current 8.3ms Single half sine-wave superimposed on rated load (JEDEC Method)	I <sub>FSM</sub>	30							A
Forward Voltage @ $I_F = 1.0\text{A}$	V <sub>FM</sub>	1.0		1.3		1.7		V	
Peak Reverse Current @ $T_A = 25^{\circ}\text{C}$ At Rated DC Blocking Voltage @ $T_A = 100^{\circ}\text{C}$	I <sub>RM</sub>	5.0				100			$\mu\text{A}$
Reverse Recovery Time (Note 2)	t <sub>rr</sub>	50				75			nS
Typical Junction Capacitance (Note 3)	C <sub>j</sub>	20				10			pF
Operating Temperature Range	T <sub>j</sub>	-65 to +125							$^{\circ}\text{C}$
Storage Temperature Range	T <sub>STG</sub>	-65 to +150							$^{\circ}\text{C}$

**MOTOROLA**  
**SEMICONDUCTOR TECHNICAL DATA**

Order this document  
 by MUR420/D

**SWITCHMODE™ Power Rectifiers**

... designed for use in switching power supplies, inverters and as free wheeling diodes, these state-of-the-art devices have the following features:

- Ultrafast 25, 50 and 75 Nanosecond Recovery Times
- 175°C Operating Junction Temperature
- Low Forward Voltage
- Low Leakage Current
- High Temperature Glass Passivated Junction
- Reverse Voltage to 600 Volts

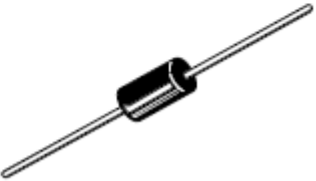
**Mechanical Characteristics:**

- Case: Epoxy, Molded
- Weight: 1.1 gram (approximately)
- Finish: All External Surfaces Corrosion Resistant and Terminal Leads are Readily Solderable
- Lead and Mounting Surface Temperature for Soldering Purposes: 220°C Max. for 10 Seconds, 1/16" from case
- Shipped in plastic bags, 5,000 per bag
- Available Tape and Reeled, 1500 per reel, by adding a "RL" suffix to the part number
- Polarity: Cathode indicated by Polarity Band
- Marking: U420, U460

**MUR420**  
**MUR460**

MUR420 and MUR460 are  
 Motorola Preferred Devices

**ULTRAFAST**  
**RECTIFIERS**  
**4.0 AMPERES**  
**200-600 VOLTS**



CASE 267-03  
 PLASTIC



MUR1510 - MUR1515 - MUR1520 - MUR1540 - MUR1560 (15A)

**MUR1510, MUR1515,  
MUR1520, MUR1540,  
MUR1560**

Preferred Devices

**SWITCHMODE™  
Power Rectifiers**

These state-of-the-art devices are a series designed for use in switching power supplies, inverters and as free wheeling diodes.

**Features**

- Ultrafast 35 and 60 Nanosecond Recovery Time
- 175°C Operating Junction Temperature
- Popular TO-220 Package
- High Voltage Capability to 600 V
- Low Forward Drop
- Low Leakage Specified @ 150°C Case Temperature
- Current Derating Specified @ Both Case and Ambient Temperatures
- Pb-Free Packages are Available\*

**Mechanical Characteristics:**

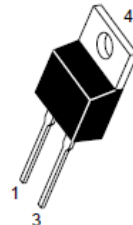
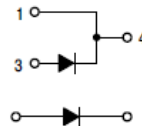
- Case: Epoxy, Molded
- Weight: 1.9 Grams (Approximately)
- Finish: All External Surfaces Corrosion Resistant and Terminal Leads are Readily Solderable
- Lead Temperature for Soldering Purposes: 260°C Max. for 10 Seconds



**ON Semiconductor®**

<http://onsemi.com>

**ULTRAFAST RECTIFIERS  
15 AMPERES, 100-600 VOLTS**



TO-220AC  
CASE 221B  
PLASTIC

**MUR1510, MUR1515, MUR1520, MUR1540, MUR1560****MAXIMUM RATINGS**

Rating	Symbol	MUR					Unit
		1510	1515	1520	1540	1560	
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	$V_{RRM}$ $V_{RWM}$ $V_R$	100	150	200	400	600	V
Average Rectified Forward Current (Rated $V_R$ )	$I_{F(AV)}$	15 @ $T_C = 150^\circ\text{C}$			15 @ $T_C = 145^\circ\text{C}$		A
Peak Rectified Forward Current (Rated $V_R$ , Square Wave, 20 kHz)	$I_{FRM}$	30 @ $T_C = 150^\circ\text{C}$			30 @ $T_C = 145^\circ\text{C}$		A
Nonrepetitive Peak Surge Current (Surge applied at rated load conditions halfwave, single phase, 60 Hz)	$I_{FSM}$	200			150		A
Operating Junction Temperature and Storage Temperature Range	$T_J, T_{stg}$	-65 to +175					$^\circ\text{C}$

Maximum ratings are those values beyond which device damage can occur. Maximum ratings applied to the device are individual stress limit values (not normal operating conditions) and are not valid simultaneously. If these limits are exceeded, device functional operation is not implied, damage may occur and reliability may be affected.

**THERMAL CHARACTERISTICS**

Parameter	Symbol	Value	Unit
Maximum Thermal Resistance, Junction-to-Case	$R_{\theta JC}$	1.5	$^\circ\text{C/W}$

**ELECTRICAL CHARACTERISTICS**

Characteristic	Symbol	1520	1540	1560	Unit
Maximum Instantaneous Forward Voltage (Note 1) ( $I_F = 15\text{ A}$ , $T_C = 150^\circ\text{C}$ ) ( $I_F = 15\text{ A}$ , $T_C = 25^\circ\text{C}$ )	$V_F$	0.85 1.05	1.12 1.25	1.20 1.50	V
Maximum Instantaneous Reverse Current (Note 1) (Rated DC Voltage, $T_C = 150^\circ\text{C}$ ) (Rated DC Voltage, $T_C = 25^\circ\text{C}$ )	$i_R$	500 10	500 10	1000 10	$\mu\text{A}$
Maximum Reverse Recovery Time ( $I_F = 1.0\text{ A}$ , $di/dt = 50\text{ A}/\mu\text{s}$ )	$t_{rr}$	35	60		ns

1. Pulse Test: Pulse Width = 300  $\mu\text{s}$ , Duty Cycle  $\leq 2.0\%$ .



MUR840 - MUR860 - RURP840 - RURP (8A)



**MUR840, MUR860, RURP840, RURP860**

*Data Sheet*

*January 2002*

**8A, 400V - 600V Ultrafast Diodes**

The MUR840, MUR860, RURP840 and RURP860 are low forward voltage drop ultrafast recovery rectifiers ( $t_{rr} < 60ns$ ). They use a glass-passivated ion-implanted, epitaxial construction.

These devices are intended for use as output rectifiers and flywheel diodes in a variety of high-frequency pulse-width modulated switching regulators. Their low stored charge and attendant fast reverse-recovery behavior minimize electrical noise generation and in many circuits markedly reduce the turn-on dissipation of the associated power switching transistors.

Formerly developmental type TA09616.

**Ordering Information**

PART NUMBER	PACKAGE	BRAND
MUR840	TO-220AC	MUR840
RURP840	TO-220AC	RURP840
MUR860	TO-220AC	MUR860
RURP860	TO-220AC	RURP860

NOTE: When ordering, use the entire part number.

**Symbol**



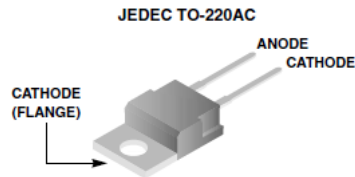
**Features**

- Ultrafast with Soft Recovery . . . . . <60ns
- Operating Temperature . . . . . 175°C
- Reverse Voltage . . . . . 600V
- Avalanche Energy Rated
- Planar Construction

**Applications**

- Switching Power Supplies
- Power Switching Circuits
- General Purpose

**Packaging**



**Absolute Maximum Ratings**  $T_C = 25^{\circ}\text{C}$ , Unless Otherwise Specified

	<b>MUR840</b> <b>RURP840</b>	<b>MUR860</b> <b>RURP860</b>	<b>UNITS</b>
Peak Repetitive Reverse Voltage..... $V_{RRM}$	400	600	V
Working Peak Reverse Voltage..... $V_{RWM}$	400	600	V
DC Blocking Voltage..... $V_R$	400	600	V
Average Rectified Forward Current..... $I_{F(AV)}$ ( $T_C = 155^{\circ}\text{C}$ )	8	8	A
Repetitive Peak Surge Current..... $I_{FRM}$ (Square Wave, 20kHz)	16	16	A
Nonrepetitive Peak Surge Current..... $I_{FSM}$ (Halfwave, 1 Phase, 60Hz)	100	100	A
Maximum Power Dissipation..... $P_D$	75	75	W
Avalanche Energy (See Figures 10 and 11)..... $E_{AVL}$	20	20	mJ
Operating and Storage Temperature..... $T_{STG}, T_J$	-65 to 175	-65 to 175	$^{\circ}\text{C}$
Maximum Lead Temperature for Soldering			
Leads at 0.063 in. (1.6mm) from case for 10s..... $T_L$	300	300	$^{\circ}\text{C}$
Package Body for 10s, see Tech Brief 334..... $T_{PKG}$	260	260	$^{\circ}\text{C}$

1N5817 1N5818 1N5819 SCHOTTKY 1A 20/30/40V

## MOTOROLA SEMICONDUCTOR TECHNICAL DATA

Order this document  
by 1N5817/D

### Axial Lead Rectifiers

... employing the Schottky Barrier principle in a large area metal-to-silicon power diode. State-of-the-art geometry features chrome barrier metal, epitaxial construction with oxide passivation and metal overlap contact. Ideally suited for use as rectifiers in low-voltage, high-frequency inverters, free wheeling diodes, and polarity protection diodes.

- Extremely Low  $v_f$
- Low Stored Charge, Majority Carrier Conduction
- Low Power Loss/High Efficiency

#### Mechanical Characteristics

- Case: Epoxy, Molded
- Weight: 0.4 gram (approximately)
- Finish: All External Surfaces Corrosion Resistant and Terminal Leads are Readily Solderable
- Lead and Mounting Surface Temperature for Soldering Purposes: 220°C Max. for 10 Seconds, 1/16" from case
- Shipped in plastic bags, 1000 per bag.
- Available Tape and Reeled, 5000 per reel, by adding a "RL" suffix to the part number
- Polarity: Cathode Indicated by Polarity Band
- Marking: 1N5817, 1N5818, 1N5819

**1N5817**  
**1N5818**  
**1N5819**

1N5817 and 1N5819 are  
Motorola Preferred Devices

**SCHOTTKY BARRIER  
RECTIFIERS  
1 AMPERE  
20, 30 and 40 VOLTS**



CASE 59-04

**MAXIMUM RATINGS**

Rating	Symbol	1N5817	1N5818	1N5819	Unit
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	$V_{RRM}$ $V_{RWM}$ $V_R$	20	30	40	V
Non-Repetitive Peak Reverse Voltage	$V_{RSM}$	24	36	48	V
RMS Reverse Voltage	$V_{R(RMS)}$	14	21	28	V
Average Rectified Forward Current (2) ( $V_{R(equiv)} \leq 0.2 V_R(dc)$ , $T_L = 90^\circ C$ , $R_{\theta JA} = 80^\circ C/W$ , P.C. Board Mounting, see Note 2, $T_A = 55^\circ C$ )	$I_O$	1.0			A
Ambient Temperature (Rated $V_R(dc)$ , $P_{F(AV)} = 0$ , $R_{\theta JA} = 80^\circ C/W$ )	$T_A$	85	80	75	$^\circ C$
Non-Repetitive Peak Surge Current (Surge applied at rated load conditions, half-wave, single phase 60 Hz, $T_L = 70^\circ C$ )	$I_{FSM}$	25 (for one cycle)			A
Operating and Storage Junction Temperature Range (Reverse Voltage applied)	$T_J, T_{stg}$	-65 to +125			$^\circ C$
Peak Operating Junction Temperature (Forward Current applied)	$T_{J(pk)}$	150			$^\circ C$

**THERMAL CHARACTERISTICS (2)**

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	80	$^\circ C/W$

**ELECTRICAL CHARACTERISTICS ( $T_L = 25^\circ C$  unless otherwise noted) (2)**

Characteristic	Symbol	1N5817	1N5818	1N5819	Unit
Maximum Instantaneous Forward Voltage (1) ( $i_F = 0.1 A$ ) ( $i_F = 1.0 A$ ) ( $i_F = 3.0 A$ )	$V_F$	0.32 0.45 0.75	0.33 0.55 0.875	0.34 0.6 0.9	V
Maximum Instantaneous Reverse Current @ Rated dc Voltage (1) ( $T_L = 25^\circ C$ ) ( $T_L = 100^\circ C$ )	$I_R$	1.0 10	1.0 10	1.0 10	mA

1N5820 1N5821 1N5822 SCHOTTKY 3A 20/30/40V



## 1N5820 - 1N5822

### 3.0 AMPS. Schottky Barrier Rectifiers

#### DO-201AD

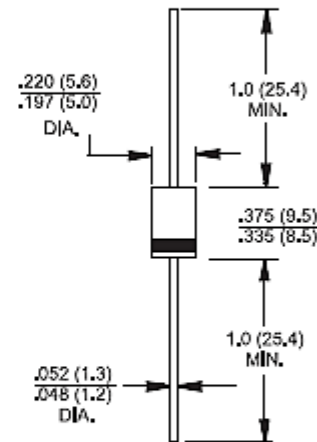


### Features

- ✦ Low power loss, high efficiency.
- ✦ High current capability, Low VF.
- ✦ High reliability
- ✦ High surge current capability.
- ✦ Epitaxial construction.
- ✦ Guard-ring for transient protection.
- ✦ For use in low voltage, high frequency inverter, free wheeling, and polarity protection application.

### Mechanical Data

- ✦ Cases: DO-201AD molded plastic
- ✦ Epoxy: UL 94V-0 rate flame retardant
- ✦ Lead: Pure tin plated, lead free, solderable per MIL-STD-202, Method 208 guaranteed
- ✦ Polarity: Color band denotes cathode end
- ✦ High temperature soldering guaranteed: 260°C/10 seconds/.375",(9.5mm) lead lengths at 5 lbs., (2.3kg) tension
- ✦ Weight: 1.10 grams



Dimensions in inches and (millimeters)

DIODO DUPLO F12C 12A (CATOPDO COMUM F12A (ANODO COMUM) /F12D (DUPLO)

Pb Free Plating Product

**F12C20C thru F12C60C**



12.0 Ampere Common Cathode Fast Recovery Rectifier Diode

**Feature**

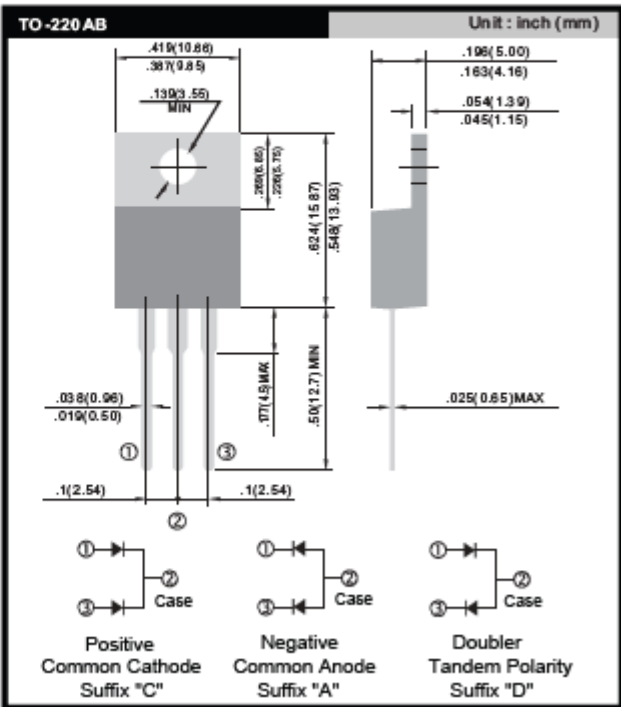
- \* Fast switching for high efficiency
- \* Low forward voltage drop
- \* High current capability
- \* Low reverse leakage current
- \* High surge current capability

**Application**

- \* Automotive Environment(Inverters/Converters)
- \* Plating Power Supply,Adaptor,SMPS and UPS
- \* Car Audio Amplifiers and Sound Device System

**Mechanical Data**

- \* Case:TO-220AB Heatsink
- \* Epoxy: UL 94V-0 rate flame retardant
- \* Terminals: Solderable per MIL-STD-202 method 208
- \* Polarity: As marked on diode body
- \* Mounting position: Any
- \* Weight: 2.2 gram approximately



**MAXIMUM RATINGS AND ELECTRICAL CHARACTERISTICS**

Rating at 25°C ambient temperature unless otherwise specified.

Single phase, half wave, 60Hz, resistive or inductive load.

For capacitive load, derate current by 20%.

	SYMBOL	<b>F12C20C</b> <b>F12C20A</b> <b>F12C20D</b>	<b>F12C40C</b> <b>F12C40A</b> <b>F12C40D</b>	<b>F12C60C</b> <b>F12C60A</b> <b>F12C60D</b>	UNIT
Maximum Recurrent Peak Reverse Voltage	VRRM	200	400	600	V
Maximum RMS Voltage	VRMS	140	280	420	V
Maximum DC Blocking Voltage	VDC	200	400	600	V
Maximum Average Forward Rectified Current Tc=100°C	IF(AV)	12.0			A
Peak Forward Surge Current, 8.3ms single Half sine-wave superimposed on rated load (JEDEC method)	IFSM	100			A
Maximum Instantaneous Forward Voltage @ 6.0 A	VF	0.98	1.3	1.7	V
Maximum DC Reverse Current @TJ=25°C At Rated DC Blocking Voltage @TJ=125°C	IR	10.0 250			uA uA
Maximum Reverse Recovery Time (Note 1)	Trr	35			nS
Typical junction Capacitance (Note 2)	CJ	65			pF
Typical Thermal Resistance (Note 3)	RθJC	2.2			°CW
Operating Junction and Storage Temperature Range	TJ, TSTG	-55 to +150			°C

- NOTES : (1) Reverse recovery test conditions IF = 0.5A, IR = 1.0A, Irr = 0.25A.  
 (2) Measured at 1.0 MHz and applied reverse voltage of 4.0 Volts DC.  
 (3) Thermal Resistance junction to case.

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DIODO PARA CORRENTES MAIORES QUE 1A

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DIODO 1N5404 1N5407 1N5408 (3A) 400V/700V/1000V

1N5404-SI-D 400V 3A

1N5407-SI-D 700V 3A

1N5408-SI-D 1000V 3A

(a corrente dos 1N54 é 3A o que muda é a tensão)



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DIODO 6A10 10A10

6A10 SI-D 1000V 6A

10A10 SI-D 1000V 10A

DIODO ZENER

## DIODO ZENER 1N TABELA

**TABELA DE DIODOS DE REFERÊNCIA - ZENER**

Tipo	Vz	/W	Tipo	Vz	/W	Tipo	Vz	/W	Tipo	Vz	/W
1N746	3,3	/0,4W	1N5221	2,4	/0,5W	1N4728	3,3	/1W	1N5333	3,3	/5W
1N747	3,6	/0,4W	1N5222	2,5	/0,5W	1N4729	3,6	/1W	1N5334	3,6	/5W
1N748	3,9	/0,4W	1N5223	2,7	/0,5W	1N4730	3,9	/1W	1N5335	3,9	/5W
1N749	4,3	/0,4W	1N5224	2,8	/0,5W	1N4731	4,3	/1W	1N5336	4,3	/5W
1N750	4,7	/0,4W	1N5225	3,0	/0,5W	1N4732	4,7	/1W	1N5337	4,7	/5W
1N751	5,1	/0,4W	1N5226	3,3	/0,5W	1N4733	5,1	/1W	1N5338	5,1	/5W
1N752	5,6	/0,4W	1N5227	3,6	/0,5W	1N4734	5,6	/1W	1N5339	5,6	/5W
1N753	6,2	/0,4W	1N5228	3,9	/0,5W	1N4735	6,2	/1W	1N5340	6,0	/5W
1N754	6,8	/0,4W	1N5229	4,3	/0,5W	1N4736	6,8	/1W	1N5341	6,2	/5W
1N755	7,5	/0,4W	1N5230	4,7	/0,5W	1N4737	7,5	/1W	1N5342	6,8	/5W
1N756	8,2	/0,4W	1N5231	5,1	/0,5W	1N4738	8,2	/1W	1N5343	7,5	/5W
1N757	9,1	/0,4W	1N5232	5,6	/0,5W	1N4739	9,1	/1W	1N5344	8,2	/5W
1N758	10	/0,4W	1N5234	6,2	/0,5W	1N4740	10	/1W	1N5345	8,7	/5W
1N759	12	/0,4W	1N5235	6,8	/0,5W	1N4742	12	/1W	1N5346	9,1	/5W
1N957	6,8	/0,4W	1N5236	7,5	/0,5W	1N4743	13	/1W	1N5347	10	/5W
1N958	7,5	/0,4W	1N5237	8,2	/0,5W	1N4744	15	/1W	1N5348	11	/5W
1N959	8,2	/0,4W	1N5239	9,1	/0,5W	1N4745	16	/1W	1N5349	12	/5W
1N960	9,1	/0,4W	1N5240	10	/0,5W	1N4746	18	/1W	1N5350	13	/5W
1N961	10	/0,4W	1N5242	12	/0,5W	1N4747	20	/1W	1N5351	14	/5W
1N962	11	/0,4W	1N5245	15	/0,5W	1N4748	22	/1W	1N5352	15	/5W
1N963	12	/0,4W	1N5246	16	/0,5W	1N4749	24	/1W	1N5353	16	/5W
1N964	13	/0,4W	1N5248	18	/0,5W	1N4750	27	/1W	1N5354	17	/5W
1N965	15	/0,4W	1N5250	20	/0,5W	1N4751	30	/1W	1N5355	18	/5W
1N966	16	/0,4W	1N5251	22	/0,5W	1N4752	33	/1W	1N5356	19	/5W
1N967	18	/0,4W	1N5252	24	/0,5W	1N4753	36	/1W	1N5357	20	/5W
1N968	20	/0,4W	1N5254	27	/0,5W	1N4754	39	/1W	1N5358	22	/5W

1N969 22 /0,4W	1N5256 30 /0,5W	1N4755 43 /1W	1N5359 24 /5W
1N970 24 /0,4W	1N5257 33 /0,5W	1N4756 47 /1W	1N5361 27 /5W
1N971 27 /0,4W	1N5258 36 /0,5W	1N4757 51 /1W	1N5362 28 /5W
1N972 30 /0,4W	1N5259 39 /0,5W	1N4758 56 /1W	1N5363 30 /5W
1N973 33 /0,4W	1N5260 43 /0,5W	1N4759 62 /1W	1N5364 33 /5W
1N974 36 /0,4W	1N5261 47 /0,5W	1N4760 68 /1W	1N5365 36 /5W
1N975 39 /0,4W	1N5262 51 /0,5W	1N4761 75 /1W	1N5366 39 /5W
1N976 43 /0,4W	1N5263 56 /0,5W	1N4762 82 /1W	1N5367 43 /5W
1N977 47 /0,4W	1N5265 62 /0,5W	1N4763 91 /1W	1N5368 47 /5W
1N978 51 /0,4W	1N5266 68 /0,5W	1N4764 100 /1W	1N5369 51 /5W
1N979 56 /0,4W	1N5267 75 /0,5W		1N5370 56 /5W
1N980 62 /0,4W	1N5268 82 /0,5W		1N5371 60 /5W
1N981 68 /0,4W	1N5270 91 /0,5W		1N5372 62 /5W
1N982 75 /0,4W	1N5271 100 /0,5W		1N5373 68 /5W
1N983 82 /0,4W			1N5374 75 /5W
1N984 91 /0,4W			
1N985 100 /0,4W			
1N986 110 /0,4W			
1N987 120 /0,4W			
1N988 130 /0,4W			
1N989 150 /0,4W			
1N990 160 /0,4W			
1N991 180 /0,4W			
1N992 200 /0,4W			

DIODO ZENER BZX TABELA DE SELEÇÃO

BZX55 Tensão Zéner desde 5,1V a 33V, Tolerância 5%, Potência 500mW					BZX85 Tensão Zéner desde 2,7V a 33V, Tolerância 5%, Potência 1,3W					
	Vz	Rz	Coefficiente	Izm						
	(IzT = 5mA)	(IzT = 5mA)	temperatura							
	V	Ω	típico	mA	Vz a	Izr	Rz a Izr	Izm		
			%/°C		V	mA	Ω mA	mA		
BZX55-C5V1	4,8 - 5,4	35	+0,015	80	BZX85-C2V7	2,7	80	20	2,7	370
BZX55-C5V6	5,2 - 6,0	25	+0,025	70	BZX85-C3V0	3,0	80	20	3,0	340
BZX55-C6V2	5,8 - 6,6	10	+0,035	64	BZX85-C3V3	3,3	80	20	3,3	320
BZX55-C6V8	6,4 - 7,2	8	+0,045	58	BZX85-C3V6	3,6	70	20	3,6	290
BZX55-C7V5	7,0 - 7,9	7	+0,050	53	BZX85-C3V9	3,9	60	15	3,9	280
BZX55-C8V2	7,7 - 8,7	7	+0,050	47	BZX85-C4V3	4,3	50	13	4,3	250
BZX55-C9V1	8,5 - 9,6	10	+0,060	43	BZX85-C4V7	4,7	45	13	4,7	215
BZX55-C10	9,4 - 10,6	15	+0,070	40	BZX85-C5V1	5,1	45	10	5,1	200
BZX55-C11	10,4 - 11,6	20	+0,070	38	BZX85-C5V6	5,6	45	7	5,6	190
BZX55-C12	11,4 - 12,7	20	+0,070	32	BZX85-C6V2	6,2	35	4	6,2	170
BZX55-C13	12,4 - 14,1	26	+0,070	29	BZX85-C6V8	6,8	35	3,5	6,8	155
BZX55-C15	13,8 - 15,6	30	+0,070	27	BZX85-C7V5	7,5	35	3	7,5	140
BZX55-C16	15,3 - 17,1	40	+0,070	24	BZX85-C8V2	8,2	25	5	8,2	130
BZX55-C18	16,8 - 19,1	50	+0,070	21	BZX85-C9V1	9,1	25	5	9,1	120
BZX55-C20	18,8 - 21,2	55	+0,070	20	BZX85-C10	10	25	7	10	105
BZX55-C22	20,8 - 23,3	55	+0,070	18	BZX85-C11	11	20	8	11	97
BZX55-C24	22,8 - 25,6	80	+0,080	16	BZX85-C12	12	20	9	12	88
BZX55-C27	25,1 - 28,9	80	+0,080	14	BZX85-C13	13	20	10	13	79
BZX55-C30	28 - 32	80	+0,080	13	BZX85-C15	15	15	15	15	71
BZX55-C33	31 - 35	80	+0,080	12	BZX85-C16	16	15	15	16	66
					BZX85-C18	18	15	20	18	62
					BZX85-C20	20	10	24	20	56
					BZX85-C22	22	10	25	22	52
					BZX85-C24	24	10	25	24	47
					BZX85-C27	27	8	30	27	41
					BZX85-C30	30	8	30	30	36
					BZX85-C33	33	8	35	33	33



## RESUMO DOS DIODOS ZENERES 1N E UZ258.

TABELA DE DIODOS ZENERES DAS FAMÍLIAS 1N e UZ - Prof. ÉLICE - elice@mgconecta.com.br								
Volt	0.4 Watt		0.5 Watt		1 Watt		5 Watt	
2.4			1N5221	1N4617		JZ87=UZ88		JZ51=52=53
2.5			1N5222			JZ81=UZ82		JZ57=58=59
2.7			1N5223	1N4618				
2.8			1N5224					
3.0			1N5225	1N4619				
3.3	1N746		1N5226	1N4620	1N4728		1N5333	
3.6	1N747		1N5227	1N4621	1N4729		1N5334	
3.9	1N748		1N5228	1N4622	1N4730		1N5335	
4.3	1N749		1N5229	1N4623	1N4731		1N5336	
4.7	1N750		1N5230	1N4624	1N4732		1N5337	
5.1	1N751		1N5231	1N4625	1N4733		1N5338	
5.6	1N752		1N5232	1N4626	1N4734		1N5339	
6.0			1N5233	1N469			1N5340	
6.2	1N753		1N5234	1N4627	1N4735		1N5341	
6.8	1N754	1N957	1N5235	1N4628	1N4736	UZ8806	1N5342	UZ5806
7.5	1N755	1N958	1N5236	1N4629	1N4737	UZ8807	1N5343	UZ5807
8.2	1N756	1N959	1N5237	1N4630	1N4738	UZ8808	1N5344	UZ5808
8.7			1N5238	1N4695			1N5345	
9.1	1N757	1N960	1N5239	1N4631	1N4739	UZ8809	1N5346	UZ5809
10.0	1N758	1N961	1N5240	1N4632	1N4740	UZ8810	1N5347	UZ5810
11.0		1N962	1N5241	1N4633	1N4741		1N5348	
12.0	1N759	1N963	1N5242	1N4634	1N4742	UZ8812	1N5349	UZ5812
13.0	1N717	1N964	1N5243	1N4635	1N4743	UZ8813	1N5350	UZ5813
14.0			1N5244				1N5351	UZ5814
15.0	1N718	1N965	1N5245	1N4636	1N4744	UZ8815	1N5352	UZ5815
16.0	1N719	1N966	1N5246	1N4637	1N4745	UZ8816	1N5353	UZ5816
17.0			1N5247				1N5354	
18.0	1N720	1N967	1N5248	1N4638	1N4746	UZ8818	1N5355	UZ5818
19.0			1N5249				1N5356	
20.0	1N721	1N968	1N5250	1N4639	1N4747	UZ8820	1N5357	UZ5820
22.0	1N722	1N969	1N5251	1N4640	1N4748	UZ8822	1N5358	UZ5822
24.0	1N723	1N970	1N5252	1N4641	1N4749	UZ8824	1N5359	UZ5824
25.0			1N5253				1N5360	
27.0	1N724	1N971	1N5254	1N4642	1N4750	UZ8827	1N5361	UZ5827
28.0			1N5255				1N5362	

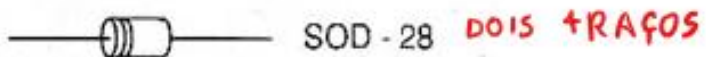
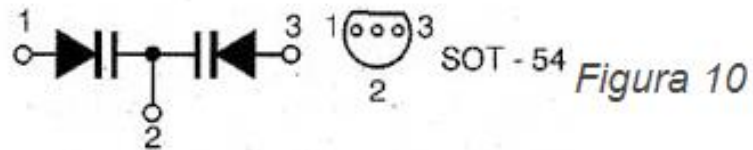
TABELA DE DIODOS ZENERES DAS FAMÍLIAS 1N e UZ - Prof. ÉLICE - elice@mgconecta.com.br								
Volt	0.4 Watt		0.5 Watt		1 Watt		5 Watt	
30.0	1N725	1N972	1N5256	1N4643	1N4751	UZ8830	1N5363	UZ5830
33.0	1N726	1N973	1N5257	1N4644	1N4752	UZ8833	1N5364	UZ5833
36.0	1N727	1N974	1N5258	1N4645	1N4753	UZ8836	1N5365	UZ5836
39.0	1N728	1N975	1N5259	1N4646	1N4754	UZ8840	1N5366	UZ5840
43.0	1N729	1N976	1N5260	1N4647	1N4755		1N5367	
47.0	1N730	1N977	1N5261	1N4648	1N4756	UZ8845	1N5368	
51.0	1N731	1N978	1N5262		1N4757	UZ8850	1N5369	UZ5850
56.0	1N732	1N979	1N5263		1N4758	UZ8856	1N5370	UZ5856
60.0			1N5264				1N5371	UZ5860
62.0	1N733	1N980	1N5265		1N4759	UZ8860	1N5372	
68.0	1N734	1N981	1N5266		1N4760	UZ8870	1N5373	
75.0	1N735	1N982	1N5267		1N4761	UZ8875	1N5374	UZ5875
82.0	1N736	1N983	1N5268		1N4762	UZ8880	1N5375	UZ5880
87.0			1N5269				1N5376	
91.0	1N737	1N984	1N5270		1N4763	UZ8890	1N5377	UZ5890
100.0	1N738	1N985	1N5271		1N4764	UZ8110	1N5378	UZ5310
110.0	1N739	1N986	1N5272			UZ8111	1N5379	UZ5311
120.0	1N740	1N987	1N5273			UZ8112	1N5380	UZ5312
130.0	1N741	1N988	1N5274			UZ8113	1N5381	UZ5313
140.0			1N5275			UZ8114	1N5382	UZ5314
150.0	1N742	1N989	1N5276			UZ8115	1N5383	UZ5315
160.0	1N743	1N990	1N5277			UZ8116	1N5384	UZ5316
170.0			1N5278			UZ8117	1N5385	UZ5317
180.0	1N744	1N991	1N5279			UZ8118	1N5386	UZ5318
190.0			1N5280			UZ8119	1N5387	UZ5319
200.0	1N745	1N992	1N5281			UZ8120	1N5388	UZ5320

DIODO VARICAP

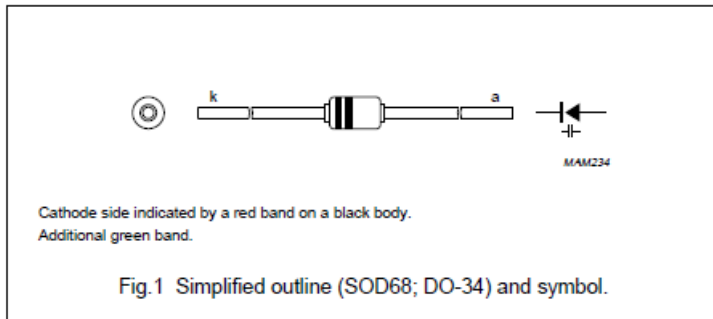


DIODO VARICAP BB TABELA DE SELEÇÃO BB119/204/405/809/909

Nº	Tipo	ENCAPS.	$V_R$ (V)	$I_F$ (mA)	$C_d$ (pF)	$V_R$ (V)	$C_{d1}/C_{d2}$	$V_{R1}/V_{R2}$ (V/V)	$r_D$ ( $\Omega$ )	Aplicações típicas
31	BB119	SOD-27	15	200	20-25	4	>1,3	4/10	<15	CAF em rádio e TV
32	BB204B	SOT-54	30	100	37-42	3	2,5-2,8	3/30	<0,4	Rádio FM
33	BB204G	SOT-54	30	100	34-39	3	2,5-2,8	3/30	<0,4	Rádio FM
34	BB405B	SOD-68A	30	20	<18	1	>7,6	1/28	<0,75	Televisão UHF
35	BB809	SOD-68A	28	20	39-46	1	8-10	1/28	<0,6	Televisão VHF
36	BB909A	SOD-68	32	20	>31	1	12-15	1/28	<0,9	Televisão VHF
37	BB909B	SOD-68	32	20	>33,5	1	12-15	1/28	<0,9	Televisão VHF

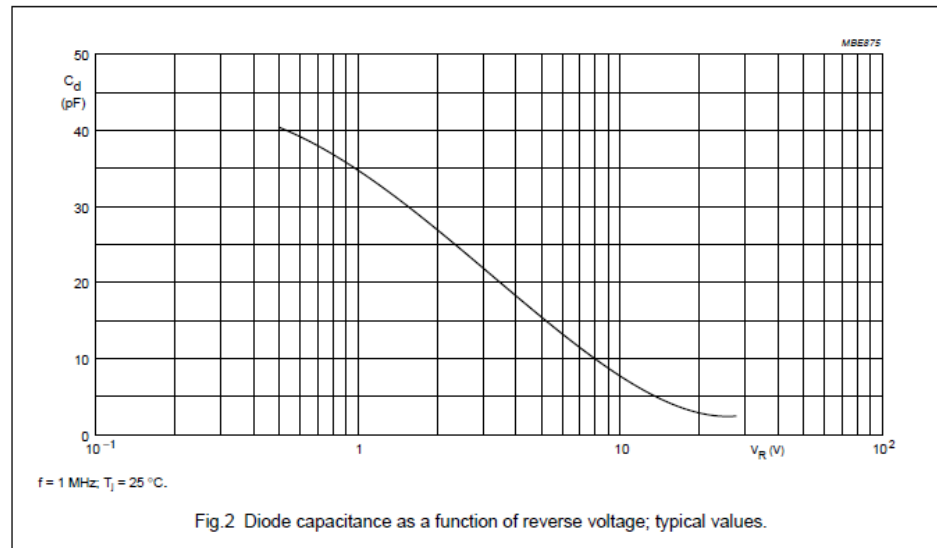


BB910

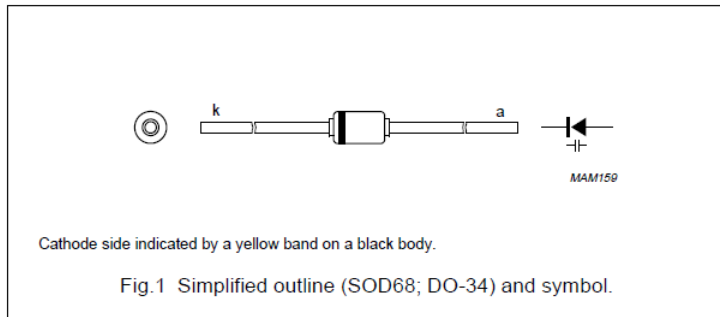


$C_d$	diode capacitance	$V_R = 0.5 \text{ V}; f = 1 \text{ MHz}; \text{ see Figs 2 and 4}$	38	-	-	pF
		$V_R = 28 \text{ V}; f = 1 \text{ MHz}; \text{ see Figs 2 and 4}$	2.3	-	2.7	pF

GRAPHICAL DATA

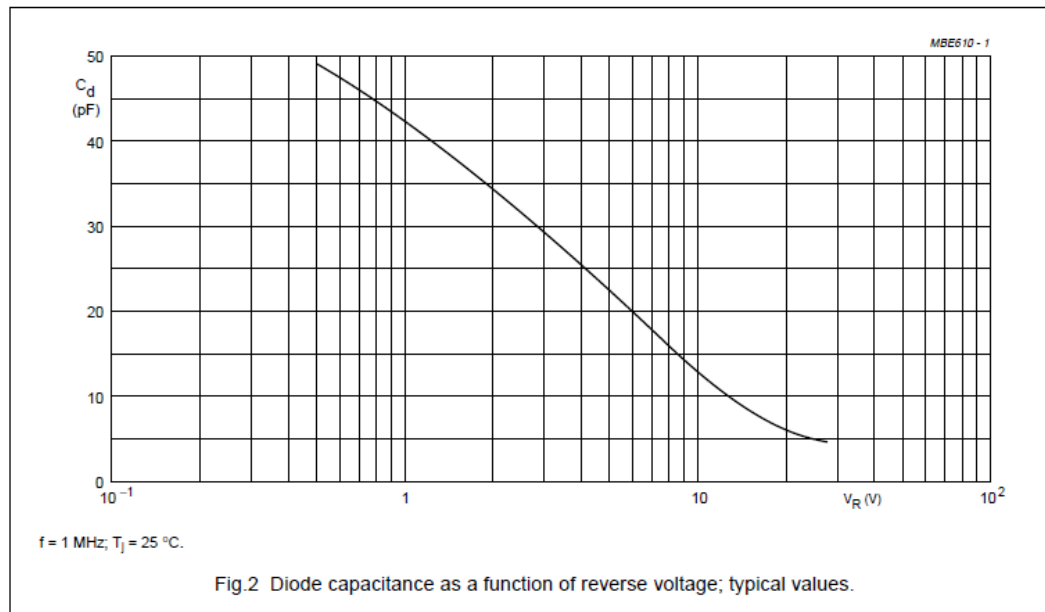


BB809

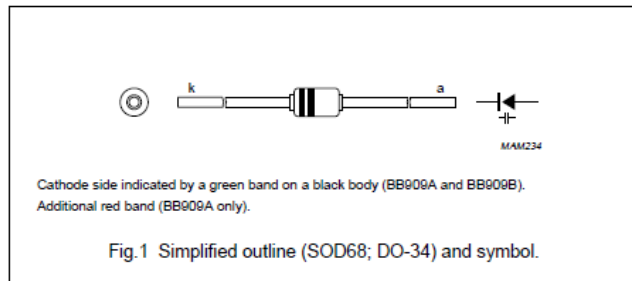


$r_s$	diode series resistance	$f = 200 \text{ MHz; note 1}$	–	–	0.6	$\Omega$
$C_d$	diode capacitance	$V_R = 1 \text{ V; } f = 1 \text{ MHz; see Figs 2 and 4}$	39	–	46	pF
		$V_R = 28 \text{ V; } f = 1 \text{ MHz; see Figs 2 and 4}$	4	–	5	pF

GRAPHICAL DATA

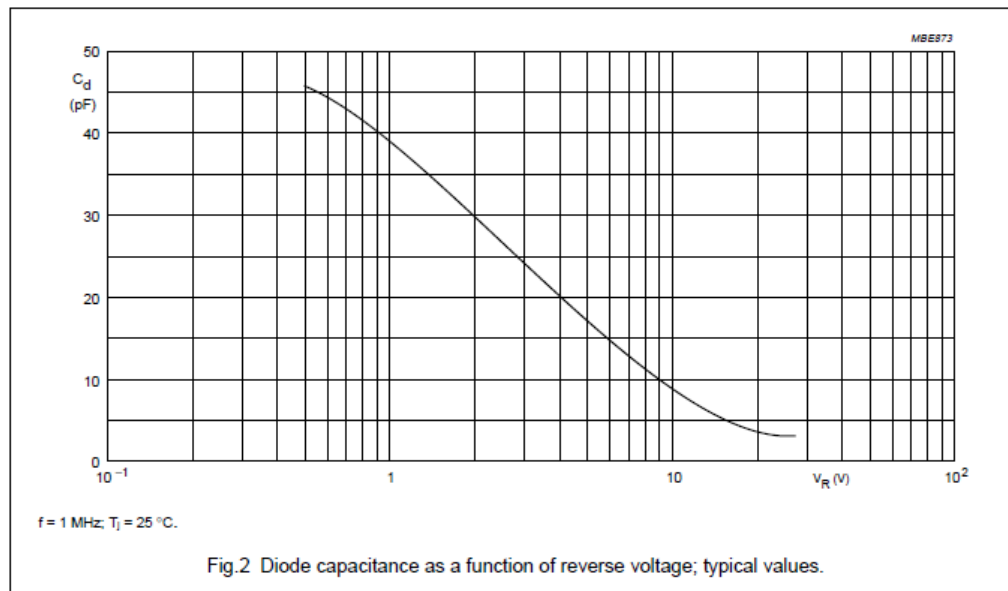


BB909A/ BB909B



$r_s$	diode series resistance	$f = 100 \text{ MHz; note 1}$	-	-	0.9	$\Omega$	
$C_d$	diode capacitance BB909A	see Figs 2 and 4					
		$V_R = 1 \text{ V; } f = 1 \text{ MHz}$	31	-	-	pF	
	$V_R = 3 \text{ V; } f = 1 \text{ MHz}$	-	23	-	-	pF	
	$V_R = 28 \text{ V; } f = 1 \text{ MHz}$	2.6	-	3	-	pF	
	BB909B	$V_R = 1 \text{ V; } f = 1 \text{ MHz}$	33.5	-	-	-	pF
		$V_R = 3 \text{ V; } f = 1 \text{ MHz}$	-	25	-	-	pF
$V_R = 28 \text{ V; } f = 1 \text{ MHz}$		2.8	-	3.2	-	pF	

GRAPHICAL DATA



BB204B/BB204G

PIN	DESCRIPTION
1	anode (a1)
2	common cathode
3	anode (a2)

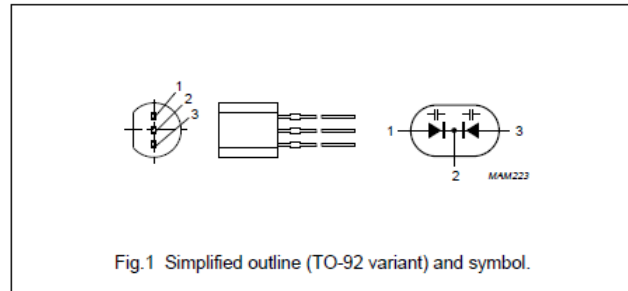


Fig.1 Simplified outline (TO-92 variant) and symbol.

$r_s$	diode series resistance	$f = 100 \text{ MHz; note 1}$	–	0.2	0.4	$\Omega$
$C_d$	diode capacitance	see Figs 2 and 4				
	BB204B	$V_R = 3 \text{ V; } f = 1 \text{ MHz}$	37	–	42	pF
		$V_R = 8 \text{ V; } f = 1 \text{ MHz}$	24	–	29	pF
	BB204G	$V_R = 30 \text{ V; } f = 1 \text{ MHz}$	–	14	–	pF
		$V_R = 3 \text{ V; } f = 1 \text{ MHz}$	34	–	39	pF
		$V_R = 8 \text{ V; } f = 1 \text{ MHz}$	22	–	27	pF
		$V_R = 30 \text{ V; } f = 1 \text{ MHz}$	–	14	–	pF

GRAPHICAL DATA

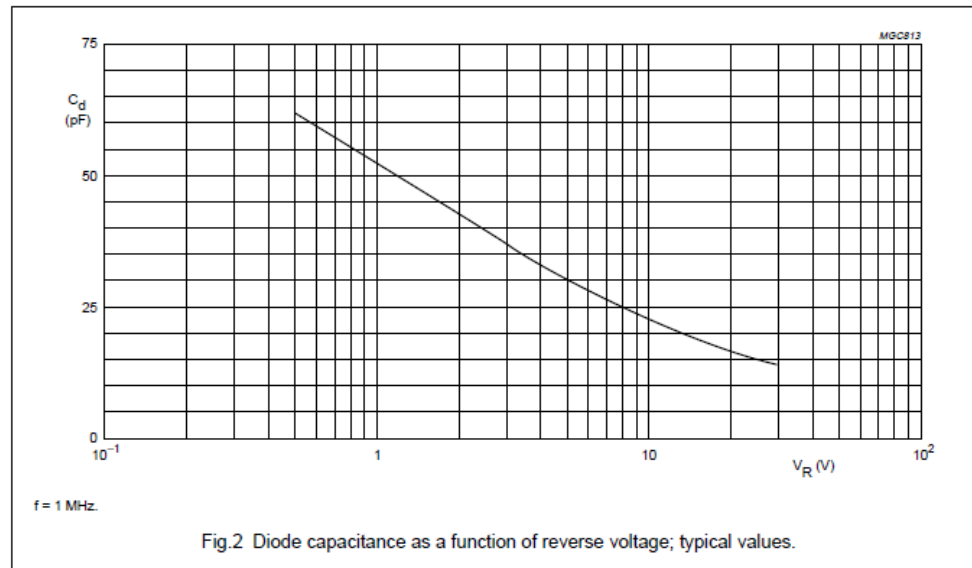
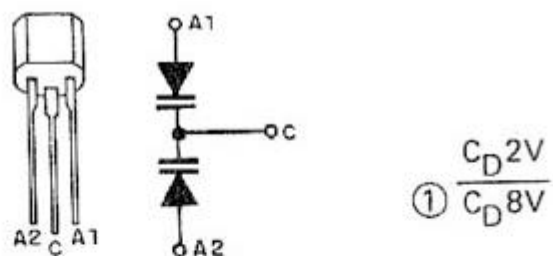


Fig.2 Diode capacitance as a function of reverse voltage; typical values.

BB204/ BB304



	BB204	BB304	
$C_D$	14	42 a 47,5	pF
Para $U_N$	30	2	V
$\frac{C_D 3V}{C_D 30V}$	2,4 a 2,8	1,65 a 1,75 <sup>①</sup>	
$r_s$	0,2	0,2	$\Omega$
Para $C_D$	38	38	pF

PONTES RETIFICADORAS

KPC200 ATÉ KBPC210 PONTE RETIFICADORA 2A



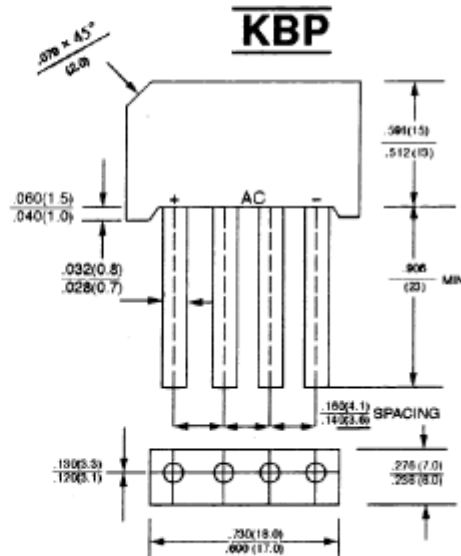
**KBP200 THRU KBP210**  
**SINGLE PHASE 2.0 AMPS. SILICON BRIDGE RECTIFIERS**



**FEATURES**

- \* Ideal for printed circuit board
- \* High Surge Current Capability
- \* Reliable low cost construction
- \* Small size, simple installation

**VOLTAGE RANGE**  
 50 to 1000 Volts  
**CURRENT**  
 2.0 Amperes



Dimensions in inches and (millimeters)



**MAXIMUM RATINGS AND ELECTRICAL CHARACTERISTICS**

Rating at 25°C ambient temperature unless otherwise specified.  
 Single phase, half wave, 60 Hz, resistive or inductive load.  
 For capacitive load, derate current by 20%

TYPE NUMBER	SYMBOLS	KBP 200	KBP 201	KBP 202	KBP 204	KBP 206	KBP 208	KBP 210	UNITS
Maximum Recurrent Peak Reverse Voltage	$V_{RRM}$	50	100	200	400	600	800	1000	V
Maximum RMS Bridge Input Voltage	$V_{RMS}$	35	70	140	280	420	560	700	V
Maximum D. C Blocking Voltage	$V_{DC}$	50	100	200	400	600	800	1000	V
Maximum Average Forward Rectified Current @ $T_A = 50^\circ\text{C}$	$I_{F(AV)}$	2.0							A
Peak Forward Surge Current, 8.3 ms single half sine-wave superimposed on rated load (JEDEC method)	$I_{FSM}$	50							A
Maximum Forward Voltage Drop per element @ 1.0A (Note)	$V_F$	1.10							V
Maximum Reverse Current at Rated @ $T_A = 25^\circ\text{C}$ D. C. Blocking Voltage per element @ $T_A = 100^\circ\text{C}$	$I_R$	10 500							$\mu\text{A}$ $\mu\text{A}$
Operating Temperature Range	$T_J$	- 55 to + 125							$^\circ\text{C}$
Storage Temperature Range	$T_{STG}$	- 55 to + 150							$^\circ\text{C}$

NOTE: Mounted on glass - epoxy P. C. B, Soldering land  $\phi 3\text{mm}$ .

FIG. 1-MAXIMUM NON - REPETITIVE PEAK FORWARD SURGE CURRENT - PER ELEMENT

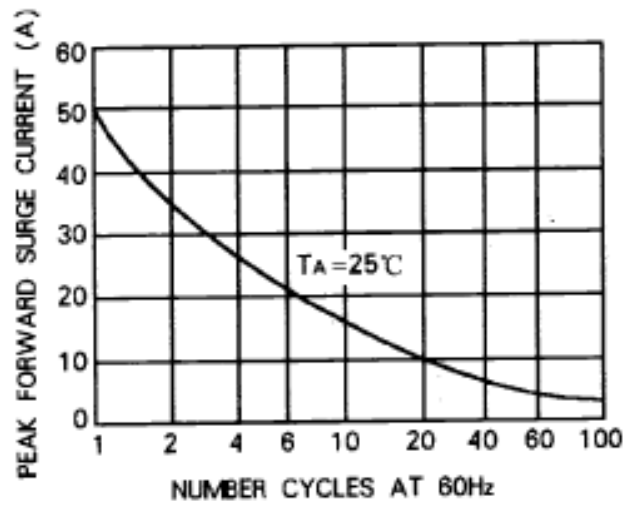


FIG. 2-TYPICAL FORWARD OUTPUT CURRENT DERATING CURVE

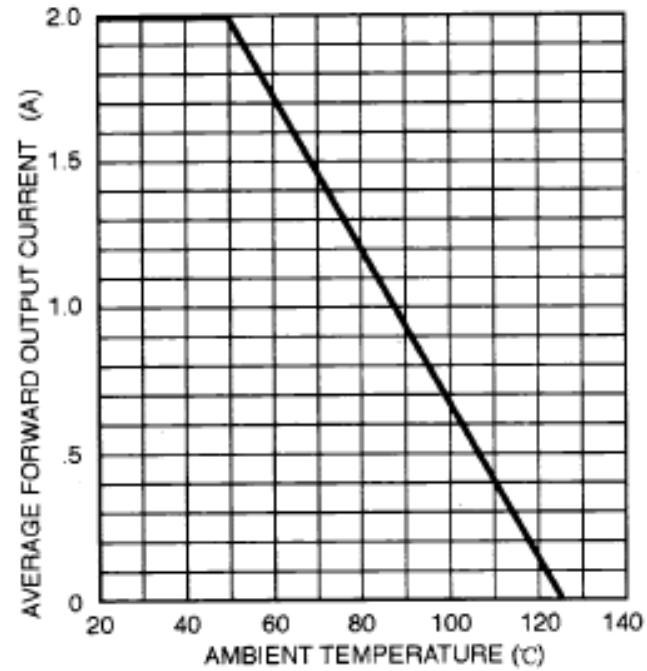


FIG. 3-TYPICAL INSTANTANEOUS FORWARD CHARACTERISTICS – PER ELEMENT

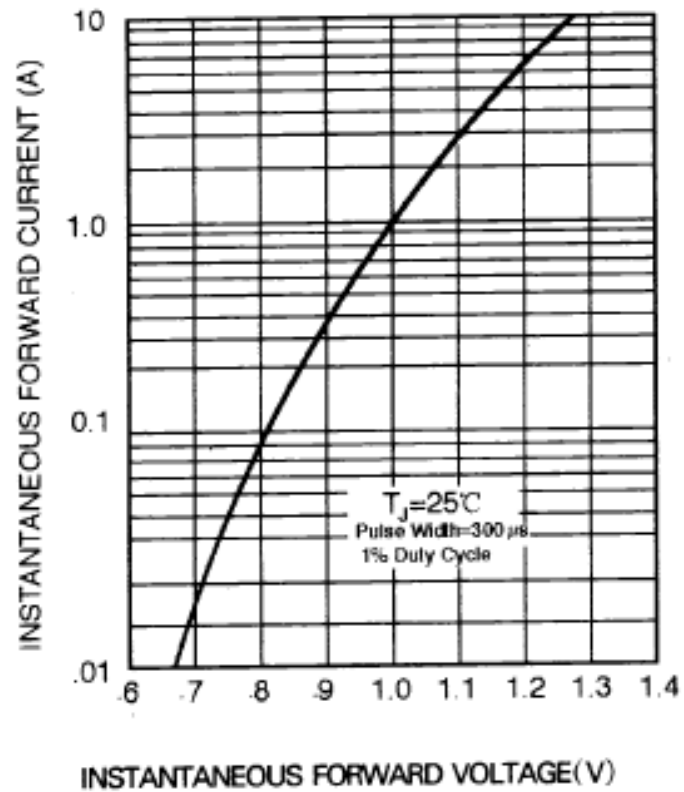
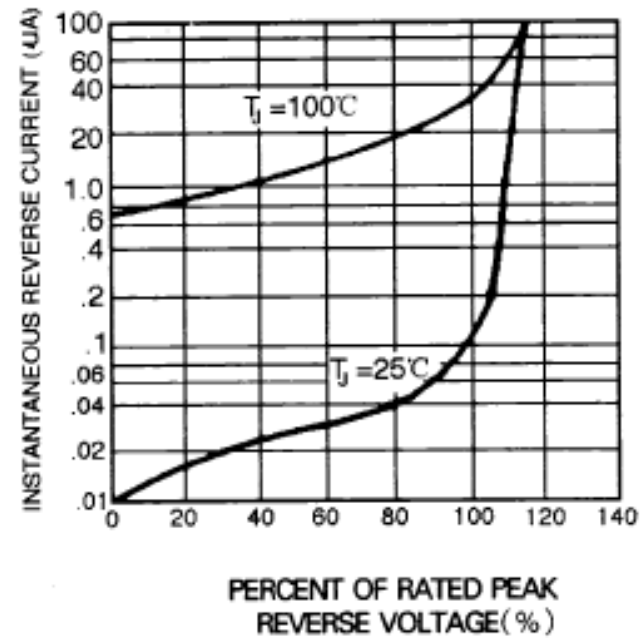


FIG. 4-TYPICAL REVERSE CHARACTERISTICS PER ELEMENT



KBPC15 KBPC25 KBPC35 / W SERIE



# KBPC15, 25, 35/W SERIES

## 15, 25, 35A HIGH CURRENT BRIDGE RECTIFIER

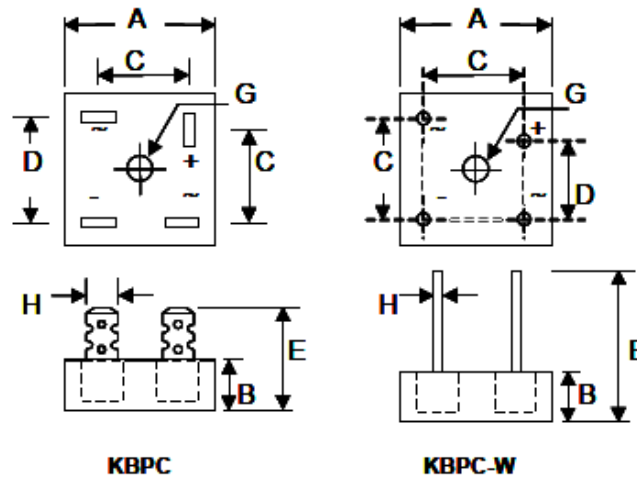
### Features

- Diffused Junction
- Low Reverse Leakage Current
- Low Power Loss, High Efficiency
- Electrically Isolated Metal Case for Maximum Heat Dissipation
- Case to Terminal Isolation Voltage 2500V
- UL Recognized File # E157705

### Mechanical Data

- Case: Metal Case with Electrically Isolated Epoxy
- Terminals: Plated Leads Solderable per MIL-STD-202, Method 208
- Polarity: Symbols Marked on Case
- Mounting: Through Hole for #10 Screw
- Weight: KBPC 31.6 grams (approx.)  
KBPC-W 28.5 grams (approx.)
- Marking: Type Number

\*W\* Suffix Designates Wire Leads  
No Suffix Designates Faston Terminals



Dim	KBPC		KBPC-W	
	Min	Max	Min	Max
A	28.40	28.70	28.40	28.70
B	10.97	11.23	10.97	11.23
C	15.70	16.70	17.10	19.10
D	17.50	18.50	10.90	11.90
E	22.86	25.40	30.50	—
G	Hole for #10 screw, 5.08Ø Nominal			
H	6.35 Typical		0.97Ø	1.07Ø
All Dimension in mm				

**Maximum Ratings and Electrical Characteristics** @ $T_A = 25^\circ\text{C}$  unless otherwise specified

Single Phase, half wave, 60Hz, resistive or inductive load.

For capacitive load, derate current by 20%.

Characteristics	Symbol	-00/W	-01/W	-02/W	-04/W	-06/W	-08/W	-10/W	Unit
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	VRRM VRRM VR	50	100	200	400	600	800	1000	V
RMS Reverse Voltage	VR(RMS)	35	70	140	280	420	560	700	V
Average Rectifier Output Current @ $T_C = 60^\circ\text{C}$	KBPC15 KBPC25 KBPC35  Io				15 25 35				A
Non-Repetitive Peak Forward Surge Current 8.3ms single half sine-wave Superimposed on rated load (JEDEC Method)	KBPC15 KBPC25 KBPC35  IFSM				300 400 400				A
Forward Voltage Drop (per element)	KBPC15 @ $I_F = 7.5\text{A}$ KBPC25 @ $I_F = 12.5\text{A}$ KBPC35 @ $I_F = 17.5\text{A}$  VFM				1.2				V
Peak Reverse Current At Rated DC Blocking Voltage	@ $T_C = 25^\circ\text{C}$ @ $T_C = 125^\circ\text{C}$  IRM				10 1.0				$\mu\text{A}$ mA
I <sup>2</sup> t Rating for Fusing (t < 8.3ms) (Note 1)	KBPC15 KBPC25 KBPC35  I <sup>2</sup> t				373 373 664				A <sup>2</sup> s

**Maximum Ratings and Electrical Characteristics** @ $T_A=25^{\circ}\text{C}$  unless otherwise specified

Typical Junction Capacitance (per element) (Note 2)	$C_j$	300	pF
Typical Thermal Resistance Junction to Case (per element) (Note 3)	$R_{\theta JC}$	6.3 3.8 2.7	K/W
RMS Isolation Voltage from Case to Lead	Viso	2500	V
Operating and Storage Temperature Range	$T_j, T_{STG}$	-65 to +150	$^{\circ}\text{C}$

\* Glass passivated forms are available upon request.

- Note: 1. Measured at non-repetitive, for  $t > 1\text{ms}$  and  $< 8.3\text{ms}$ .  
 2. Measured at 1.0 MHz and applied reverse voltage of 4.0V D.C.  
 3. Thermal resistance junction to case mounted on heatsink.

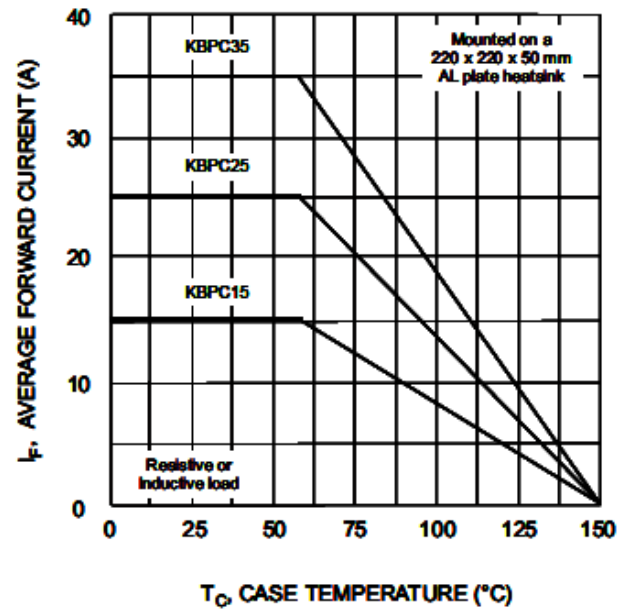


Fig. 1 Forward Current Derating Curve

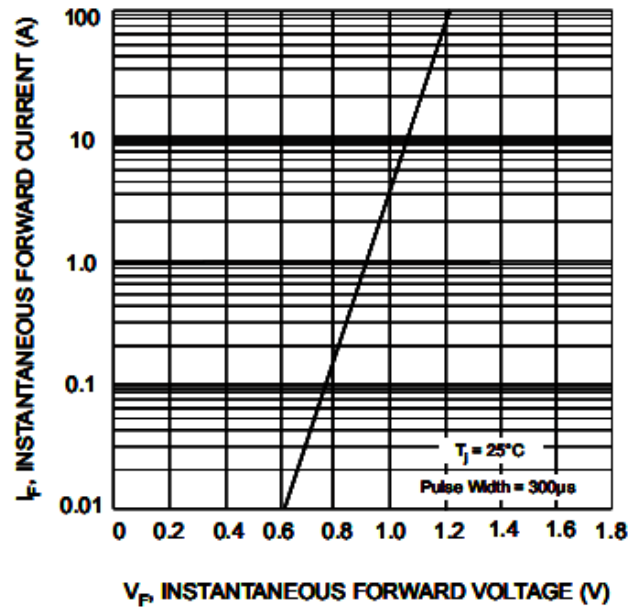


Fig. 2 Typical Forward Characteristics (per element)

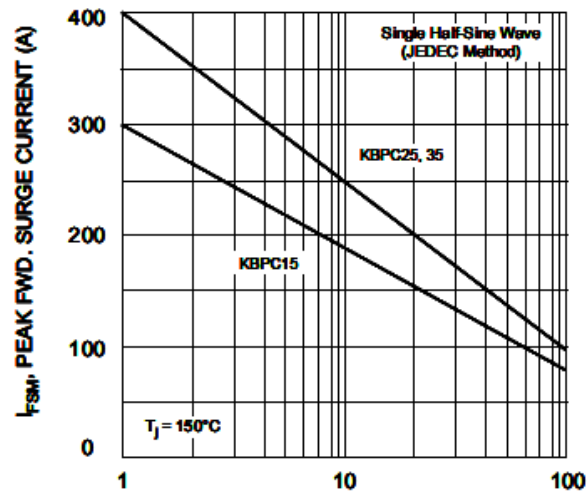


Fig. 3 Max Non-Repetitive Surge Current

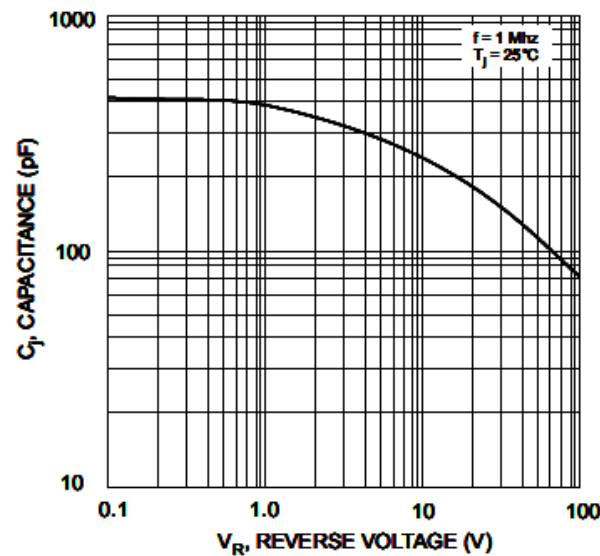


Fig. 4 Typical Junction Capacitance (per element)

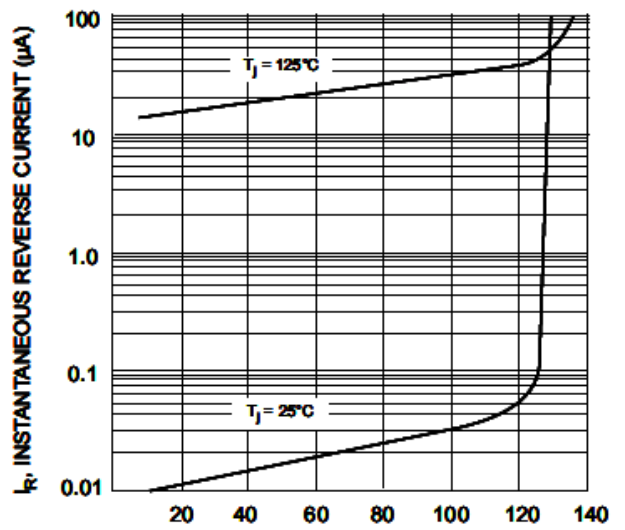


Fig. 5 Typical Reverse Characteristics (per element)

KBP005/01/02/03/04/06/08 PONTE RETIFICADORA 2A

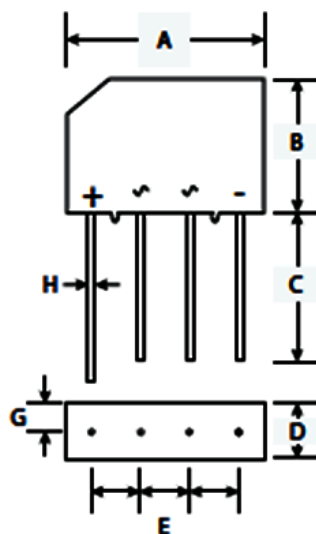
# DEC

**KBP005 THRU KBP10**

**CURRENT 2.0 Amperes  
VOLTAGE 50 to 1000 Volts**

### Features

- Glass Passivated Die Construction
- High Case Dielectric Strength of 1500V<sub>RMS</sub>
- Low Reverse Leakage Current
- Surge Overload Rating to 40A Peak
- Ideal for Printed Circuit Board Applications
- Plastic Material - UL Flammability Classification 94V-0



KBP		
Dim	Min	Max
A	14.00	15.00
B	10.50	11.50
C	15.00	—
D	4.70	5.00
E	3.50	4.00
G	2.30	2.50
H	0.70 Typical	
All Dimensions in mm		

### Mechanical Data

- Case: Molded Plastic
- Terminals: Plated Leads, Solderable per MIL-STD-202, Method 208
- Polarity: As Marked on Body
- Approx. Weight: 1.52 grams
- Mounting Position: Any
- Marking: Type Number



## Maximum Ratings And Electrical Characteristics

(Ratings at 25 °C ambient temperature unless otherwise specified, Single phase, half wave 60Hz, resistive or inductive load. For capacitive load, derate by 20%)

	Symbols	KBP 005	KBP 01	KBP 02	KBP 04	KBP 06	KBP 08	KBP 10	Units
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	$V_{RRM}$ $V_{WRM}$ $V_R$	50	100	200	400	600	800	1000	Volts
RMS Reverse voltage	$V_{R(RMS)}$	35	70	140	280	420	560	700	Volts
Average Rectified Output Current @ $T_C=105^\circ\text{C}$	$I_o$	2.0							Amps
Non-Repetitive Peak Forward Surge Current, 8.3ms single half-sine-wave superimposed on rated load per element (JEDEC method)	$I_{FSM}$	40							Amps
Forward Voltage (per element) @ $I_F=2.0\text{ A}$	$V_{FM}$	1.1							Volts
Peak Reverse Current at Rated DC Blocking Voltage	@ $T_C=25^\circ\text{C}$	5.0							$\mu\text{A}$
	@ $T_C=125^\circ\text{C}$	500							
Typical Junction Capacitance (Note 1)	$C_j$	20							pF
Typical Thermal Resistance, Junction to Case (Note 2)	$R_{\theta JC}$	30							$^\circ\text{C}/\text{W}$
Operating and Storage Temperature Range	$T_j$ $T_{STG}$	-65 to +150							$^\circ\text{C}$

### Notes:

(1) Thermal resistance from junction to case per element. Unit mounted on 300 x 300 x 16mm aluminum plate heat sink.

(2) Measured at 1.0MHz and Applied Reverse Voltage of 4.0V DC.

KBU800/ KBU801/ KBU802/ KBU804/ KBU806/ KBU808/ KBU810 PONTE RETIFICADORA 8A



# KBU800 – KBU810

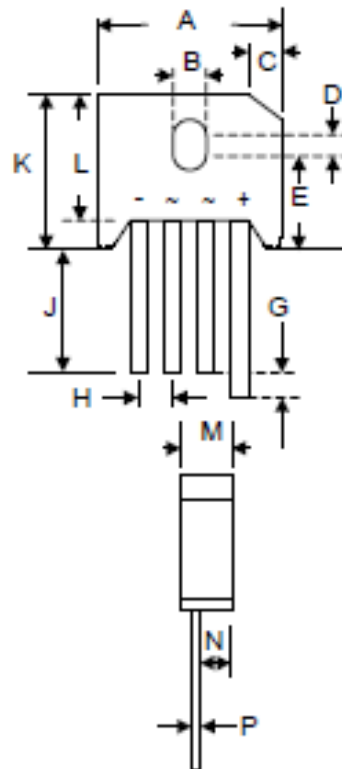
## 8.0A BRIDGE RECTIFIER

### Features

- Diffused Junction
- Low Forward Voltage Drop
- High Current Capability
- High Reliability
- High Surge Current Capability
- Ideal for Printed Circuit Boards
- UL Recognized File # E157705

### Mechanical Data

- Case: Molded Plastic
- Terminals: Plated Leads Solderable per MIL-STD-202, Method 208
- Polarity: As Marked on Body
- Weight: 8.0 grams (approx.)
- Mounting Position: Any
- Marking: Type Number



KBU		
Dim	Min	Max
A	22.70	23.70
B	3.80	4.10
C	4.20	4.70
D	1.70	2.20
E	10.30	11.30
G	4.50	6.80
H	4.60	5.60
J	25.40	—
K	—	19.30
L	16.80	17.80
M	6.60	7.10
N	4.70	5.20
P	1.20	1.30
All Dimensions in mm		

**Maximum Ratings and Electrical Characteristics** @ $T_A=25^{\circ}\text{C}$  unless otherwise specified

Single Phase, half wave, 60Hz, resistive or inductive load.  
For capacitive load, derate current by 20%.

Characteristic	Symbol	KBU 800	KBU 801	KBU 802	KBU 804	KBU 806	KBU 808	KBU 810	Unit
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	VRRM VRWM VR	50	100	200	400	600	800	1000	V
RMS Reverse Voltage	VR(RMS)	35	70	140	280	420	560	700	V
Average Rectified Output Current @ $T_C = 100^{\circ}\text{C}$	IO	8.0							A
Non-Repetitive Peak Forward Surge Current 8.3ms Single half sine-wave superimposed on rated load (JEDEC Method)	IFSM	300							A
Forward Voltage (per element) @ $I_F = 4.0\text{A}$	VFM	1.0							V
Peak Reverse Current @ $T_C = 25^{\circ}\text{C}$ At Rated DC Blocking Voltage @ $T_C = 100^{\circ}\text{C}$	IR	10 1.0							$\mu\text{A}$ mA
Rating for Fusing ( $t < 8.3\text{ms}$ ) (Note 1)	$I^2t$	373							$\text{A}^2\text{s}$
Typical Thermal Resistance (Note 2)	$R_{\theta JC}$	7.5							K/W
Operating and Storage Temperature Range	$T_j, T_{STG}$	-65 to +150							$^{\circ}\text{C}$

Note: 1. Non-repetitive for  $t > 1\text{ms}$  and  $< 8.3\text{ms}$ .

2. Thermal resistance junction to case per element mounted on PC board with 13.0x13.0x0.03mm thick land areas.

KBU1000 -KBU101 PONTE RETIFICADORA 10A



## KBU1000 – KBU1010

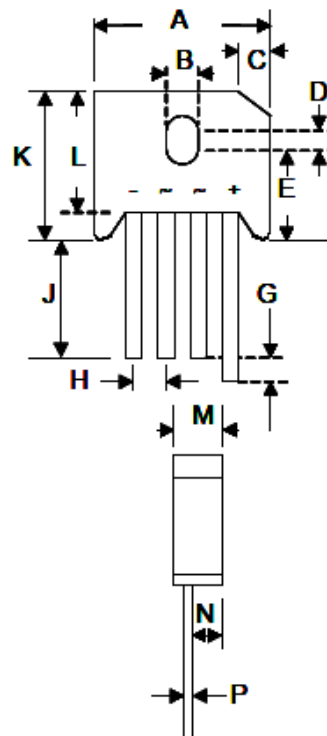
### 10A BRIDGE RECTIFIER

#### Features

- Diffused Junction
- Low Forward Voltage Drop
- High Current Capability
- High Reliability
- High Surge Current Capability
- Ideal for Printed Circuit Boards

#### Mechanical Data

- Case: Molded Plastic
- Terminals: Plated Leads Solderable per MIL-STD-202, Method 208
- Polarity: As Marked on Body
- Weight: 8.0 grams (approx.)
- Mounting Position: Any
- Marking: Type Number



KBU		
Dim	Min	Max
A	22.70	23.70
B	3.80	4.10
C	4.20	4.70
D	1.70	2.20
E	10.30	11.30
G	4.50	6.80
H	4.60	5.60
J	25.40	—
K	—	19.30
L	16.80	17.80
M	6.60	7.10
N	4.70	5.20
P	1.20	1.30
All Dimensions in mm		

**Maximum Ratings and Electrical Characteristics** @ $T_A=25^\circ\text{C}$  unless otherwise specified

Single Phase, half wave, 60Hz, resistive or inductive load.

For capacitive load, derate current by 20%.

Characteristic	Symbol	KBU 1000	KBU 1001	KBU 1002	KBU 1004	KBU 1006	KBU 1008	KBU 1010	Unit
Peak Repetitive Reverse Voltage	VRRM								
Working Peak Reverse Voltage	VRWM	50	100	200	400	600	800	1000	V
DC Blocking Voltage	VR								
RMS Reverse Voltage	VR(RMS)	35	70	140	280	420	560	700	V
Average Rectified Output Current @ $T_C = 100^\circ\text{C}$	IO	10							A
Non-Repetitive Peak Forward Surge Current 8.3ms Single half sine-wave superimposed on rated load (JEDEC Method)	IFSM	300							A
Forward Voltage (per element) @ $I_F = 5.0\text{A}$	VFM	1.0							V
Peak Reverse Current @ $T_C = 25^\circ\text{C}$	IR	10							$\mu\text{A}$
At Rated DC Blocking Voltage @ $T_C = 100^\circ\text{C}$		1.0							mA
Rating for Fusing ( $t < 8.3\text{ms}$ ) (Note 1)	$I^2t$	373							$\text{A}^2\text{s}$
Typical Thermal Resistance (Note 2)	$R_{\theta JC}$	8.0							K/W
Operating and Storage Temperature Range	$T_j, T_{STG}$	-65 to +150							$^\circ\text{C}$

 Note: 1. Non-repetitive for  $t > 1\text{ms}$  and  $< 8.3\text{ms}$ .

2. Thermal resistance junction to case per element mounted on PC board with 13.0x13.0x0.03mm thick land areas.

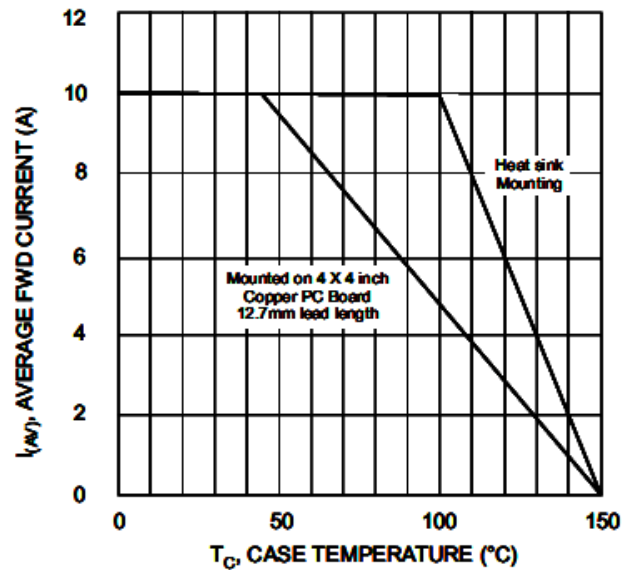


Fig. 1 Forward Current Derating Curve

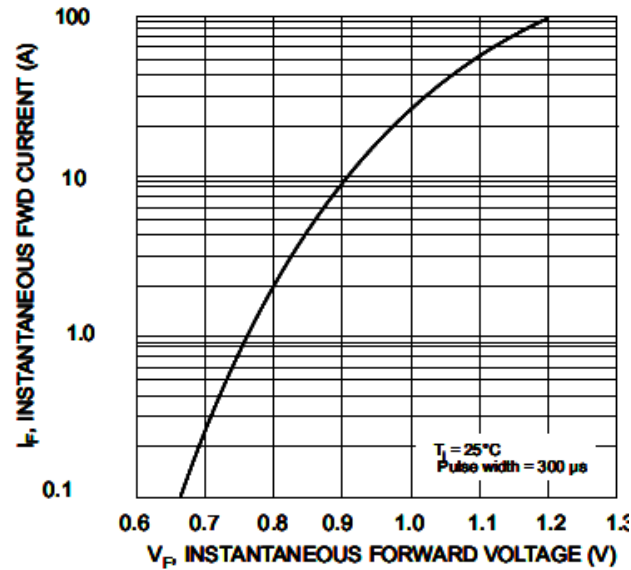


Fig. 2 Typical Forward Characteristics, per element

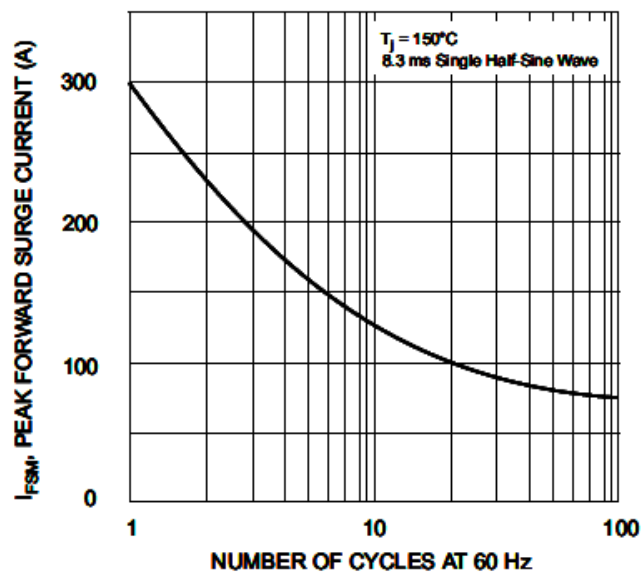


Fig. 3 Max Non-Repetitive Peak Fwd Surge Current

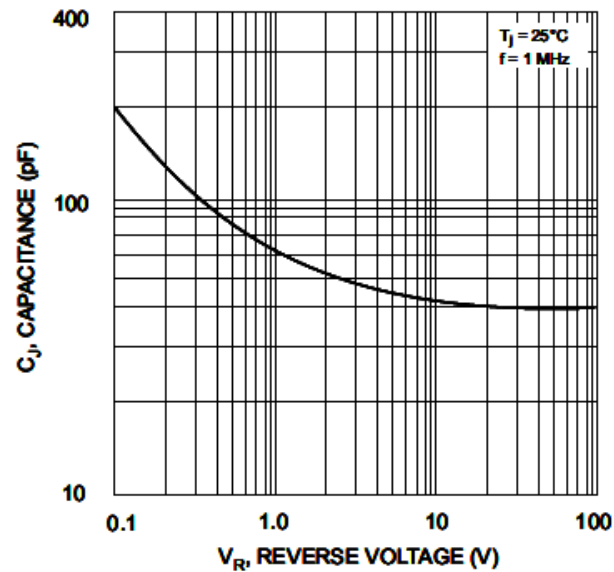


Fig. 4 Typical Junction Capacitance per element

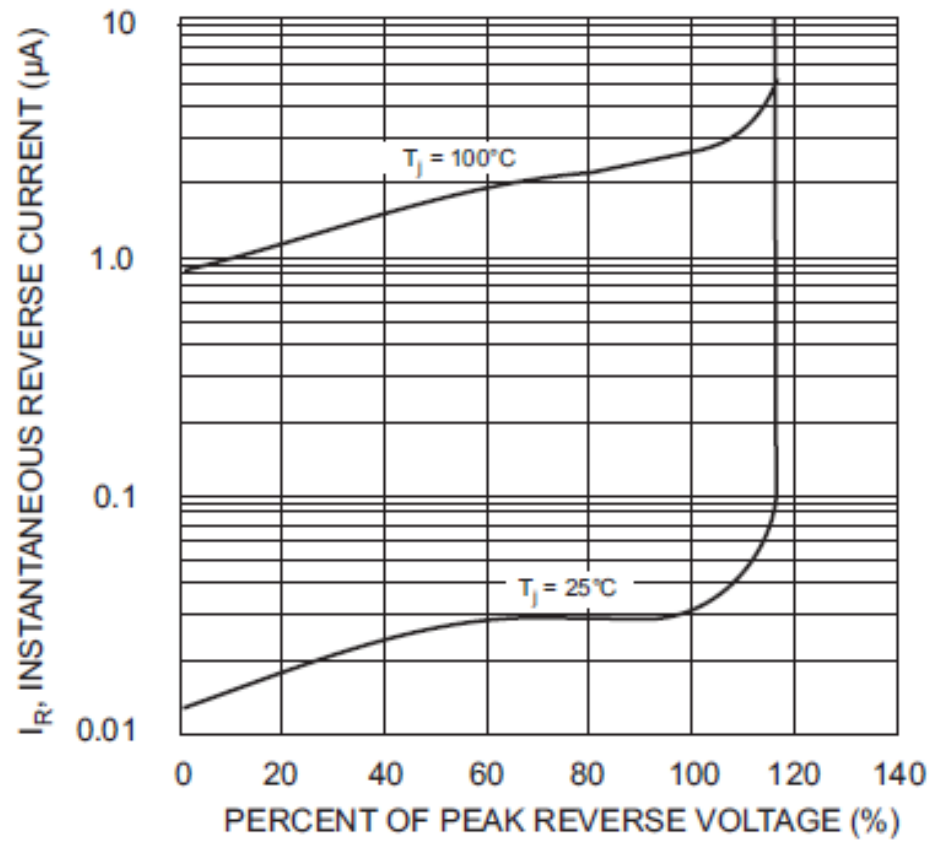


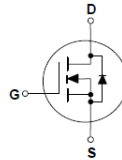
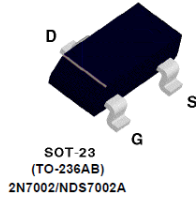
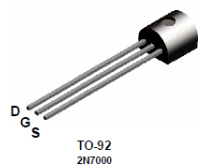
Fig. 5 Typical Reverse Characteristics

MOSFET



2N7000/2N7002 MOSFET N BAIXA POTÊNCIA 200MA BAIXO VGS .

MOSFET prático para usar em baixas potências, mas tem a vantagem de ter baixo vgs



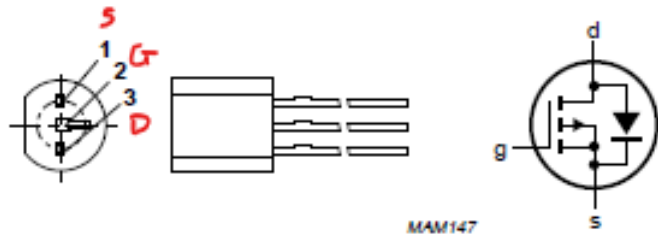
**Absolute Maximum Ratings**  $T_A = 25^\circ\text{C}$  unless otherwise noted

Symbol	Parameter	2N7000	2N7002	NDS7002A	Units
$V_{DSS}$	Drain-Source Voltage	60			V
$V_{DGR}$	Drain-Gate Voltage ( $R_{GS} \leq 1\text{ M}\Omega$ )	60			V
$V_{GSS}$	Gate-Source Voltage - Continuous	$\pm 20$			V
	- Non Repetitive ( $t_p < 50\mu\text{s}$ )	$\pm 40$			
$I_D$	Maximum Drain Current - Continuous	200	115	280	mA
	- Pulsed	500	800	1500	
$P_D$	Maximum Power Dissipation	400	200	300	mW
	Derated above $25^\circ\text{C}$	3.2	1.6	2.4	
$T_J, T_{STG}$	Operating and Storage Temperature Range	-55 to 150		-65 to 150	$^\circ\text{C}$
$T_L$	Maximum Lead Temperature for Soldering Purposes, 1/16" from Case for 10 Seconds	300			$^\circ\text{C}$
<b>THERMAL CHARACTERISTICS</b>					
$R_{\theta JA}$	Thermal Resistance, Junction-to-Ambient	312.5	625	417	$^\circ\text{C/W}$

**ELECTRICAL CHARACTERISTICS** ( $T_A=25^\circ\text{C}$  unless otherwise specified)

Parameter	Symbol	Min.	Typ.	Max.	Unit	Test Condition
Drain-Source Breakdown Voltage	$V_{(BR)DSS}$	60	-	-	V	$V_{GS}=0, I_D=10\mu\text{A}$
Gate Threshold Voltage <sup>1</sup>	$V_{GS(th)}$	0.8	-	3	V	$V_{DS}=V_{GS}, I_D=1\text{mA}$
Gate-Source Leakage Current	$I_{GSS}$	-	-	$\pm 10$	nA	$V_{DS}=0, V_{GS}=\pm 15\text{V}$
Drain-Source Leakage Current	$I_{DSS}$	-	-	1	$\mu\text{A}$	$V_{DS}=60\text{V}, V_{GS}=0$
On-State Drain Current	$I_{D(ON)}$	75	-	-	mA	$V_{DS}=10\text{V}, V_{GS}=4.5\text{V}$
Static Drain-Source On-Resistance <sup>1</sup>	$R_{DS(ON)}$	-	-	6	$\Omega$	$V_{GS}=4.5\text{V}, I_D=75\text{mA}$
		-	-	5		$V_{GS}=10\text{V}, I_D=500\text{mA}$
Forward Transconductance <sup>1</sup>	$g_{fs}$	100	-	-	mS	$V_{DS}=10\text{V}, I_D=200\text{mA}$
Drain-Source On-Voltage <sup>1</sup>	$V_{DS(ON)}$	-	-	0.45	V	$V_{GS}=4.5\text{V}, I_D=75\text{mA}$
		-	-	2.5		$V_{GS}=10\text{V}, I_D=500\text{mA}$
Input Capacitance <sup>2</sup>	$C_{iss}$	-	60	-	pF	$V_{GS}=0$ $V_{DS}=25\text{V}$ $f=1\text{MHz}$
Output Capacitance <sup>2</sup>	$C_{oss}$	-	25	-		
Reverse Transfer Capacitance <sup>2</sup>	$C_{rss}$	-	5	-		
Turn-on Delay Time <sup>2</sup>	$T_{d(on)}$	-	10	-	nS	$V_{DD}=15\text{V}, V_{GEN}=10\text{V}$ $R_L=30\Omega, R_G=25\Omega, I_D=500\text{mA}$
Turn-off Delay Time <sup>2</sup>	$T_{d(off)}$	-	10	-		

BS250 P MOSFET 250MA



QUICK REFERENCE DATA

Drain-source voltage	$-V_{DS}$	max.	45 V
Gate-source voltage (open drain)	$\pm V_{GSO}$	max.	20 V
Drain current (DC)	$-I_D$	max.	0.25 A
Total power dissipation up to $T_{amb} = 25\text{ }^\circ\text{C}$	$P_{tot}$	max.	0.83 W
Drain-source ON-resistance $-I_D = 200\text{ mA}; -V_{GS} = 10\text{ V}$	$R_{DS(on)}$	typ. max.	9 $\Omega$ 14 $\Omega$
Transfer admittance $-I_D = 200\text{ mA}; -V_{DS} = 15\text{ V}$	$ Y_{fs} $	typ.	125 mS

IRF510 MOSFET CANAL N 5,6A 100V RDS 0,540 OHM



IRF510

Data Sheet

November 1999

File Number 1573.4

**5.6A, 100V, 0.540 Ohm, N-Channel Power MOSFET**

This N-Channel enhancement mode silicon gate power field effect transistor is an advanced power MOSFET designed, tested, and guaranteed to withstand a specified level of energy in the breakdown avalanche mode of operation. All of these power MOSFETs are designed for applications such as switching regulators, switching convertors, motor drivers, relay drivers, and drivers for high power bipolar switching transistors requiring high speed and low gate drive power. These types can be operated directly from integrated circuits.

Formerly developmental type TA17441.

**Ordering Information**

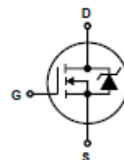
PART NUMBER	PACKAGE	BRAND
IRF510	TO-220AB	IRF510

NOTE: When ordering, include the entire part number.

**Features**

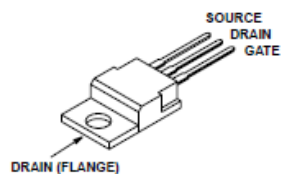
- 5.6A, 100V
- $r_{DS(ON)} = 0.540\Omega$
- Single Pulse Avalanche Energy Rated
- SOA is Power Dissipation Limited
- Nanosecond Switching Speeds
- Linear Transfer Characteristics
- High Input Impedance
- Related Literature
  - TB334 "Guidelines for Soldering Surface Mount Components to PC Boards"

**Symbol**



**Packaging**

JEDEC TO-220AB



**Absolute Maximum Ratings**  $T_C = 25^{\circ}\text{C}$ , Unless Otherwise Specified

	IRF510	UNITS
Drain to Source Voltage (Note 1) . . . . .	100	V
Drain to Gate Voltage ( $R_{GS} = 20\text{k}\Omega$ ) (Note 1) . . . . .	100	V
Continuous Drain Current . . . . .	5.6	A
$T_C = 100^{\circ}\text{C}$ . . . . .	4	A
Pulsed Drain Current (Note 3) . . . . .	20	A
Gate to Source Voltage . . . . .	$\pm 20$	V
Maximum Power Dissipation . . . . .	43	W
Linear Derating Factor . . . . .	0.29	W/ $^{\circ}\text{C}$
Single Pulse Avalanche Energy Rating (Note 4) . . . . .	19	mJ
Operating and Storage Temperature Range . . . . .	-55 to 175	$^{\circ}\text{C}$
Maximum Temperature for Soldering		
Leads at 0.063in (1.6mm) from Case for 10s . . . . .	300	$^{\circ}\text{C}$
Package Body for 10s, See Techbrief 334 . . . . .	260	$^{\circ}\text{C}$

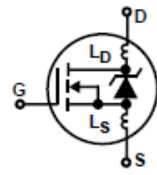
*CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.*

## NOTE:

1.  $T_J = 25^{\circ}\text{C}$  to  $150^{\circ}\text{C}$ .

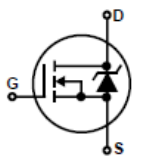
**Electrical Specifications**  $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS	
Drain to Source Breakdown Voltage	$BV_{DSS}$	$V_{GS} = 0V, I_D = 250\mu A$ , (Figure 10)	100	-	-	V	
Gate to Threshold Voltage	$V_{GS(TH)}$	$V_{GS} = V_{DS}, I_D = 250\mu A$	2.0	-	4.0	V	
Zero-Gate Voltage Drain Current	$I_{DSS}$	$V_{DS} = 95V, V_{GS} = 0V$	-	-	25	$\mu A$	
		$V_{DS} = 0.8 \times \text{Rated } BV_{DSS}, V_{GS} = 0V, T_J = 150^\circ\text{C}$	-	-	250	$\mu A$	
On-State Drain Current (Note 2)	$I_{D(ON)}$	$V_{DS} > I_{D(ON)} \times r_{DS(ON)MAX}, V_{GS} = 10V$ (Figure 7)	5.6	-	-	A	
Gate to Source Leakage Current	$I_{GSS}$	$V_{GS} = \pm 20V$	-	-	$\pm 100$	nA	
Drain to Source On Resistance (Note 2)	$r_{DS(ON)}$	$V_{GS} = 10V, I_D = 3.4A$ (Figures 8, 9)	-	0.4	0.54	$\Omega$	
Forward Transconductance (Note 2)	$g_{fs}$	$V_{GS} = 50V, I_D = 3.4A$ (Figure 12)	1.3	2.0	-	S	
Turn-On Delay Time	$t_{d(ON)}$	$I_D = 5.6A, R_{GS} = 24\Omega, V_{DD} = 50V, R_L = 9\Omega,$ $V_{DD} = 50V, V_{GS} = 10V$ MOSFET switching times are essentially independent of operating temperature	-	8	12	ns	
Rise Time	$t_r$		-	25	63	ns	
Turn-Off Delay Time	$t_{d(OFF)}$		-	15	7	ns	
Fall Time	$t_f$		-	12	59	ns	
Total Gate Charge (Gate to Source + Gate to Drain)	$Q_g(TOT)$	$V_{GS} = 10V, I_D = 5.6A, V_{DS} = 0.8 \times \text{Rated } BV_{DSS},$ $I_{G(REF)} = 1.5mA$ (Figure 14)	-	5.0	30	nC	
Gate to Source Charge	$Q_{gs}$	Gate charge is essentially independent of operating temperature.	-	2.0	-	nC	
Gate to Drain "Miller" Charge	$Q_{gd}$		-	3.0	-	nC	
Input Capacitance	$C_{ISS}$	$V_{GS} = 0V, V_{DS} = 25V, f = 1.0MHz$ (Figure 11)	-	135	-	pF	
Output Capacitance	$C_{OSS}$		-	80	-	pF	
Reverse-Transfer Capacitance	$C_{RSS}$		-	20	-	pF	
Internal Drain Inductance	$L_D$	Measured From the Contact Screw On Tab To Center of Die	Modified MOSFET Symbol Showing the Internal Devices Inductances	-	3.5	-	nH
		Measured From the Drain Lead, 6mm (0.25in) From Package to Center of Die		-	4.5	-	nH
Internal Source Inductance	$L_S$	Measured From The Source Lead, 6mm (0.25in) From Header to Source Bonding Pad		-	7.5	-	nH
Junction to Case	$R_{\theta JC}$		-	-	3.5	$^\circ\text{C/W}$	
Junction to Ambient	$R_{\theta JA}$	Free air operation	-	-	80	$^\circ\text{C/W}$	



**IRF510**

**Source to Drain Diode Specifications**

PARAMETER	SYMBOL	Test Conditions	MIN	TYP	MAX	UNITS
Continuous Source to Drain Current	$I_{SD}$	Modified MOSFET Symbol Showing the Integral Reverse P-N Junction Diode 	-	-	5.6	A
Pulse Source to Drain Current (Note 3)	$I_{SDM}$		-	-	20	A
Source to Drain Diode Voltage (Note 2)	$V_{SD}$	$T_J = 25^{\circ}\text{C}$ , $I_{SD} = 5.6\text{A}$ , $V_{GS} = 0\text{V}$ (Figure 13)	-	-	2.5	V
Reverse Recovery Time	$t_{rr}$	$T_J = 25^{\circ}\text{C}$ , $I_{SD} = 5.6\text{A}$ , $dI_{SD}/dt = 100\text{A}/\mu\text{s}$	4.6	96	200	ns
Reverse Recovered Charge	$Q_{RR}$	$T_J = 25^{\circ}\text{C}$ , $I_{SD} = 5.6\text{A}$ , $dI_{SD}/dt = 100\text{A}/\mu\text{s}$	0.17	0.4	0.83	$\mu\text{C}$

NOTES:

2. Pulse test: pulse width  $\leq 300\mu\text{s}$ , duty cycle  $\leq 2\%$ .
3. Repetitive rating: pulse width limited by max junction temperature. See Transient Thermal Impedance curve (Figure 3).
4.  $V_{DD} = 25\text{V}$ , start  $T_J = 25^{\circ}\text{C}$ ,  $L = 910\mu\text{H}$ ,  $R_G = 25\Omega$ , peak  $I_{AS} = 5.6\text{A}$ .

**Typical Performance Curves** Unless Otherwise Specified

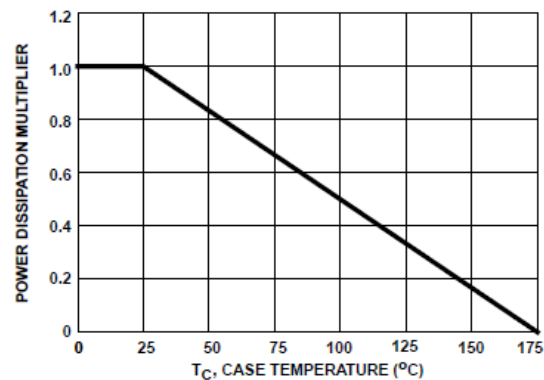


FIGURE 1. NORMALIZED POWER DISSIPATION vs CASE TEMPERATURE

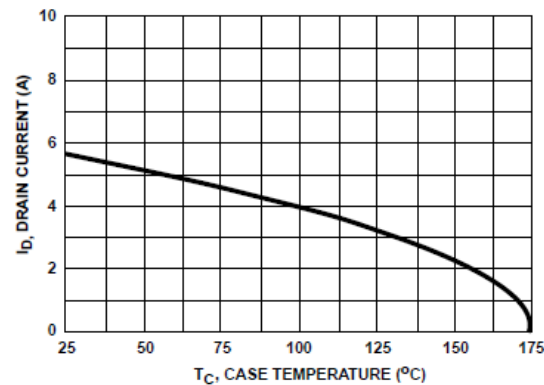


FIGURE 2. MAXIMUM CONTINUOUS DRAIN CURRENT vs CASE TEMPERATURE

IR520 MOSFET CANAL N 10A 100V RDS 0,270HM



**IRF520**  
**IRF520FI**

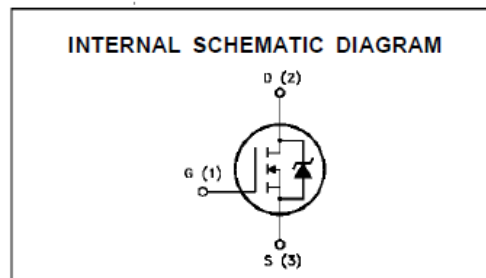
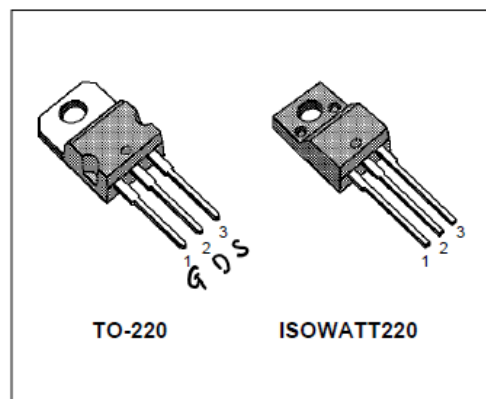
**N - CHANNEL ENHANCEMENT MODE  
POWER MOS TRANSISTORS**

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	I <sub>D</sub>
IRF520	100 V	< 0.27 Ω	10 A
IRF520FI	100 V	< 0.27 Ω	7 A

- TYPICAL R<sub>DS(on)</sub> = 0.23 Ω
- AVALANCHE RUGGED TECHNOLOGY
- 100% AVALANCHE TESTED
- REPETITIVE AVALANCHE DATA AT 100°C
- LOW GATE CHARGE
- HIGH CURRENT CAPABILITY
- 175°C OPERATING TEMPERATURE

**APPLICATIONS**

- HIGH CURRENT, HIGH SPEED SWITCHING
- SOLENOID AND RELAY DRIVERS
- REGULATORS
- DC-DC & DC-AC CONVERTERS
- MOTOR CONTROL, AUDIO AMPLIFIERS
- AUTOMOTIVE ENVIRONMENT (INJECTION, ABS, AIR-BAG, LAMPDRIVERS, Etc.)





**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value		Unit
		IRF520	IRF520FI	
$V_{DS}$	Drain-source Voltage ( $V_{GS} = 0$ )	100		V
$V_{DGR}$	Drain- gate Voltage ( $R_{GS} = 20 \text{ k}\Omega$ )	100		V
$V_{GS}$	Gate-source Voltage	$\pm 20$		V
$I_D$	Drain Current (cont.) at $T_c = 25 \text{ }^\circ\text{C}$	10	7	A
$I_D$	Drain Current (cont.) at $T_c = 100 \text{ }^\circ\text{C}$	7	5	A
$I_{DM}(\bullet)$	Drain Current (pulsed)	40	40	A
$P_{tot}$	Total Dissipation at $T_c = 25 \text{ }^\circ\text{C}$	70	35	W
	Derating Factor	0.47	0.23	W/ $^\circ\text{C}$
$V_{ISO}$	Insulation Withstand Voltage (DC)	—	2000	V
$T_{stg}$	Storage Temperature	-65 to 175		$^\circ\text{C}$
$T_j$	Max. Operating Junction Temperature	175		$^\circ\text{C}$

( $\bullet$ ) Pulse width limited by safe operating area

**IRF520/FI****THERMAL DATA**

			TO-220	ISOWATT220	
R <sub>thj-case</sub>	Thermal Resistance Junction-case	Max	2.14	4.29	°C/W
R <sub>thj-amb</sub>	Thermal Resistance Junction-ambient	Max	62.5		°C/W
R <sub>thc-s</sub>	Thermal Resistance Case-sink	Typ	0.5		°C/W
T <sub>l</sub>	Maximum Lead Temperature For Soldering Purpose		300		°C

**AVALANCHE CHARACTERISTICS**

Symbol	Parameter	Max Value	Unit
I <sub>AR</sub>	Avalanche Current, Repetitive or Not-Repetitive (pulse width limited by T <sub>j</sub> max, δ < 1%)	10	A
E <sub>AS</sub>	Single Pulse Avalanche Energy (starting T <sub>j</sub> = 25 °C, I <sub>D</sub> = I <sub>AR</sub> , V <sub>DD</sub> = 25 V)	36	mJ
E <sub>AR</sub>	Repetitive Avalanche Energy (pulse width limited by T <sub>j</sub> max, δ < 1%)	9	mJ
I <sub>AR</sub>	Avalanche Current, Repetitive or Not-Repetitive (T <sub>c</sub> = 100 °C, pulse width limited by T <sub>j</sub> max, δ < 1%)	7	A

**ELECTRICAL CHARACTERISTICS** ( $T_{\text{case}} = 25^{\circ}\text{C}$  unless otherwise specified)

OFF

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$V_{(\text{BR})\text{DSS}}$	Drain-source Breakdown Voltage	$I_{\text{D}} = 250 \mu\text{A}$ $V_{\text{GS}} = 0$	100			V
$I_{\text{DSS}}$	Zero Gate Voltage Drain Current ( $V_{\text{GS}} = 0$ )	$V_{\text{DS}} = \text{Max Rating}$ $V_{\text{DS}} = \text{Max Rating} \times 0.8$ $T_{\text{c}} = 125^{\circ}\text{C}$			250 1000	$\mu\text{A}$ $\mu\text{A}$
$I_{\text{GSS}}$	Gate-body Leakage Current ( $V_{\text{DS}} = 0$ )	$V_{\text{GS}} = \pm 20 \text{ V}$			$\pm 100$	nA

ON (\*)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$V_{\text{GS(th)}}$	Gate Threshold Voltage	$V_{\text{DS}} = V_{\text{GS}}$ $I_{\text{D}} = 250 \mu\text{A}$	2	2.9	4	V
$R_{\text{DS(on)}}$	Static Drain-source On Resistance	$V_{\text{GS}} = 10\text{V}$ $I_{\text{D}} = 5 \text{ A}$		0.23	0.27	$\Omega$
$I_{\text{D(on)}}$	On State Drain Current	$V_{\text{DS}} > I_{\text{D(on)}} \times R_{\text{DS(on)max}}$ $V_{\text{GS}} = 10 \text{ V}$	10			A

DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$g_{\text{fs}}$ (*)	Forward Transconductance	$V_{\text{DS}} > I_{\text{D(on)}} \times R_{\text{DS(on)max}}$ $I_{\text{D}} = 5 \text{ A}$	2.7	4.5		S
$C_{\text{iss}}$	Input Capacitance	$V_{\text{DS}} = 25 \text{ V}$ $f = 1 \text{ MHz}$ $V_{\text{GS}} = 0$		330	450	pF
$C_{\text{oss}}$	Output Capacitance			90	120	pF
$C_{\text{rss}}$	Reverse Transfer Capacitance			25	40	pF

**ELECTRICAL CHARACTERISTICS** (continued)

## SWITCHING RESISTIVE LOAD

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on Time	$V_{DD} = 50\text{ V}$ $I_D = 5\text{ A}$ $R_{GS} = 4.7\ \Omega$ $V_{GS} = 10\text{ V}$ (see test circuit)		10	15	ns
$t_r$	Rise Time			50	75	ns
$t_{d(off)}$	Turn-off Delay Time			25	40	ns
$t_f$	Fall Time			20	30	ns
$Q_g$	Total Gate Charge	$I_D = 10\text{ A}$ $V_{GS} = 10\text{ V}$ $V_{DD} = \text{Max Rating} \times 0.8$ (see test circuit)		15	25	nC
$Q_{gs}$	Gate-Source Charge			7		nC
$Q_{gd}$	Gate-Drain Charge			4		nC

## SOURCE DRAIN DIODE

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$I_{SD}$	Source-drain Current				10	A
$I_{SDM}(\bullet)$	Source-drain Current (pulsed)				40	A
$V_{SD} (*)$	Forward On Voltage	$I_{SD} = 10\text{ A}$ $V_{GS} = 0$			1.6	V
$t_{rr}$	Reverse Recovery Time	$I_{SD} = 10\text{ A}$ $di/dt = 100\text{ A}/\mu\text{s}$ $V_{DD} = 20\text{ V}$ $T_j = 150\text{ }^\circ\text{C}$		80		ns
$Q_{rr}$	Reverse Recovery Charge				0.22	

(\*) Pulsed: Pulse duration = 300  $\mu\text{s}$ , duty cycle 1.5 %

(\bullet) Pulse width limited by safe operating area

IRF540 MOSFET CANAL N 23A 100V RDS 0,077 OHM

Philips Semiconductors

Product specification

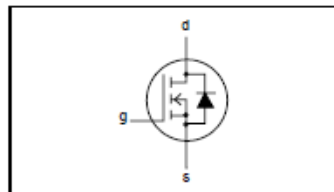
**N-channel TrenchMOS™ transistor**

**IRF540, IRF540S**

**FEATURES**

- 'Trench' technology
- Low on-state resistance
- Fast switching
- Low thermal resistance

**SYMBOL**



**QUICK REFERENCE DATA**

$V_{DS} = 100\text{ V}$
$I_D = 23\text{ A}$
$R_{DS(ON)} \leq 77\text{ m}\Omega$

**GENERAL DESCRIPTION**

N-channel enhancement mode field-effect power transistor in a plastic envelope using 'trench' technology.

**Applications:-**

- d.c. to d.c. converters
- switched mode power supplies
- T.V. and computer monitor power supplies

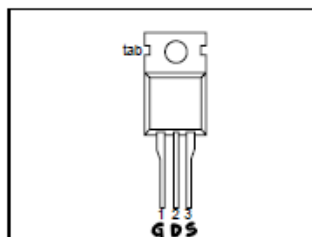
The IRF540 is supplied in the SOT78 (TO220AB) conventional leaded package.

The IRF540S is supplied in the SOT404 (D<sup>2</sup>PAK) surface mounting package.

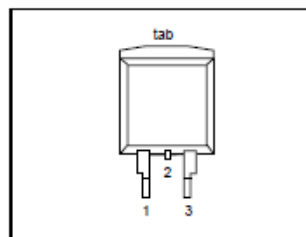
**PINNING**

PIN	DESCRIPTION
1	gate
2	drain <sup>1</sup>
3	source
tab	drain

**SOT78 (TO220AB)**



**SOT404 (D<sup>2</sup>PAK)**



**LIMITING VALUES**

Limiting values in accordance with the Absolute Maximum System (IEC 134)

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_{DSS}$	Drain-source voltage	$T_j = 25\text{ }^\circ\text{C}$ to $175\text{ }^\circ\text{C}$	-	100	V
$V_{DGR}$	Drain-gate voltage	$T_j = 25\text{ }^\circ\text{C}$ to $175\text{ }^\circ\text{C}$ ; $R_{GS} = 20\text{ k}\Omega$	-	100	V
$V_{GS}$	Gate-source voltage		-	$\pm 20$	V
$I_D$	Continuous drain current	$T_{mb} = 25\text{ }^\circ\text{C}$ ; $V_{GS} = 10\text{ V}$	-	23	A
		$T_{mb} = 100\text{ }^\circ\text{C}$ ; $V_{GS} = 10\text{ V}$	-	16	A
$I_{DM}$	Pulsed drain current	$T_{mb} = 25\text{ }^\circ\text{C}$	-	92	A
$P_D$	Total power dissipation	$T_{mb} = 25\text{ }^\circ\text{C}$	-	100	W
$T_j, T_{stg}$	Operating junction and storage temperature		- 55	175	$^\circ\text{C}$

**AVALANCHE ENERGY LIMITING VALUES**

Limiting values in accordance with the Absolute Maximum System (IEC 134)

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$E_{AS}$	Non-repetitive avalanche energy	Unclamped inductive load, $I_{AS} = 10\text{ A}$ ; $t_p = 350\text{ }\mu\text{s}$ ; $T_j$ prior to avalanche = $25\text{ }^\circ\text{C}$ ; $V_{DD} \leq 25\text{ V}$ ; $R_{GS} = 50\text{ }\Omega$ ; $V_{GS} = 10\text{ V}$ ; refer to fig:14	-	230	mJ
$I_{AS}$	Peak non-repetitive avalanche current		-	23	A

**THERMAL RESISTANCES**

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$R_{th\text{-}j\text{-}mb}$	Thermal resistance junction to mounting base		-	-	1.5	K/W
$R_{th\text{-}j\text{-}a}$	Thermal resistance junction to ambient	SOT78 package, in free air	-	60	-	K/W
		SOT404 package, pcb mounted, minimum footprint	-	50	-	K/W

**ELECTRICAL CHARACTERISTICS** $T_j = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{(BR)DSS}$	Drain-source breakdown voltage	$V_{GS} = 0\text{ V}; I_D = 0.25\text{ mA};$ $T_j = -55^\circ\text{C}$	100 89	- -	- -	V V
$V_{GS(TO)}$	Gate threshold voltage	$V_{DS} = V_{GS}; I_D = 1\text{ mA}$ $T_j = 175^\circ\text{C}$ $T_j = -55^\circ\text{C}$	2 1 -	3 -	4 -	V V V
$R_{DS(ON)}$	Drain-source on-state resistance	$V_{GS} = 10\text{ V}; I_D = 17\text{ A}$ $T_j = 175^\circ\text{C}$	- -	49 132	77 193	m $\Omega$ m $\Omega$
$g_{fs}$	Forward transconductance	$V_{DS} = 25\text{ V}; I_D = 17\text{ A}$	8.7	15.5	-	S
$I_{GSS}$	Gate source leakage current	$V_{GS} = \pm 20\text{ V}; V_{DS} = 0\text{ V}$	-	10	100	nA
$I_{DSS}$	Zero gate voltage drain current	$V_{DS} = 100\text{ V}; V_{GS} = 0\text{ V}$ $V_{DS} = 80\text{ V}; V_{GS} = 0\text{ V}; T_j = 175^\circ\text{C}$	- -	0.05 -	10 250	$\mu\text{A}$ $\mu\text{A}$
$Q_{g(tot)}$	Total gate charge	$I_D = 17\text{ A}; V_{DD} = 80\text{ V}; V_{GS} = 10\text{ V}$	-	-	65	nC
$Q_{gs}$	Gate-source charge		-	-	10	nC
$Q_{gd}$	Gate-drain (Miller) charge		-	-	29	nC
$t_{d\ on}$	Turn-on delay time	$V_{DD} = 50\text{ V}; R_D = 2.2\ \Omega;$	-	8	-	ns
$t_r$	Turn-on rise time	$V_{GS} = 10\text{ V}; R_G = 5.6\ \Omega$	-	39	-	ns
$t_{d\ off}$	Turn-off delay time	Resistive load	-	26	-	ns
$t_f$	Turn-off fall time		-	24	-	ns
$L_d$	Internal drain inductance	Measured tab to centre of die	-	3.5	-	nH
$L_d$	Internal drain inductance	Measured from drain lead to centre of die (SOT78 package only)	-	4.5	-	nH
$L_s$	Internal source inductance	Measured from source lead to source bond pad	-	7.5	-	nH
$C_{iss}$	Input capacitance	$V_{GS} = 0\text{ V}; V_{DS} = 25\text{ V}; f = 1\text{ MHz}$	-	890	1187	pF
$C_{oss}$	Output capacitance		-	139	167	pF
$C_{rss}$	Feedback capacitance		-	83	109	pF

IRF630 9A 200V RDS 0,34 OHM

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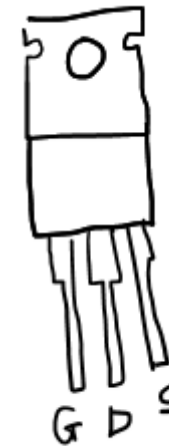
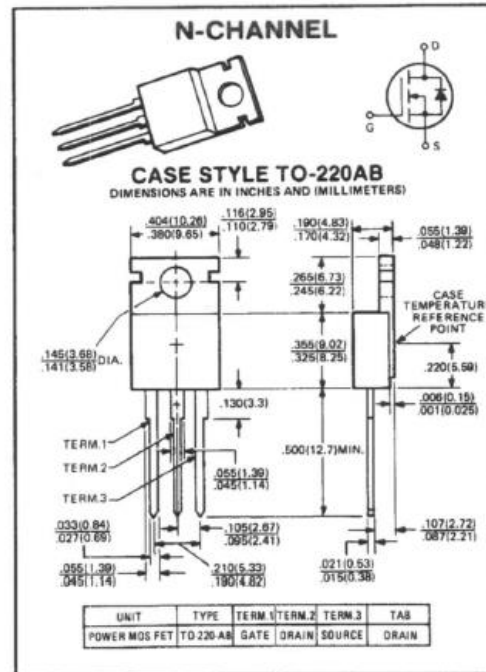
<b>IRF630,631</b> <b>D84DN2,M2</b>
9.0 AMPERES 200, 150 VOLTS RDS(ON) = 0.4 Ω

# POWER-MOS FET

## FIELD EFFECT POWER TRANSISTOR

**Features**

- Polysilicon gate — Improved stability and reliability
- No secondary breakdown — Excellent ruggedness
- Ultra-fast switching — Independent of temperature
- Voltage controlled — High transconductance
- Low input capacitance — Reduced drive requirement
- Excellent thermal stability — Ease of paralleling





maximum ratings ( $T_C = 25^\circ\text{C}$ ) (unless otherwise specified)

RATING	SYMBOL	IRF630/D84DN2	IRF631/D84DM2	UNITS
Drain-Source Voltage	$V_{DSS}$	200	150	Volts
Drain-Gate Voltage, $R_{GS} = 1M\Omega$	$V_{DGR}$	200	150	Volts
Continuous Drain Current @ $T_C = 25^\circ\text{C}$ @ $T_C = 100^\circ\text{C}$	$I_D$	9.0	9.0	A
		6.0	6.0	A
Pulsed Drain Current <sup>(1)</sup>	$I_{DM}$	36	36	A
Gate-Source Voltage	$V_{GS}$	$\pm 20$	$\pm 20$	Volts
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ Derate Above $25^\circ\text{C}$	$P_D$	75	75	Watts
		0.6	0.6	W/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{STG}$	-55 to 150	-55 to 150	$^\circ\text{C}$

thermal characteristics

Thermal Resistance, Junction to Case	$R_{\theta JC}$	1.67	1.67	$^\circ\text{C/W}$
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	80	80	$^\circ\text{C/W}$
Maximum Lead Temperature for Soldering Purposes: $\frac{1}{8}$ " from Case for 5 Seconds	$T_L$	260	260	$^\circ\text{C}$

(1) Repetitive Rating: Pulse width limited by max. junction temperature.

electrical characteristics ( $T_C = 25^\circ C$ ) (unless otherwise specified)

CHARACTERISTIC	SYMBOL	MIN	TYP	MAX	UNIT
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off characteristics

Drain-Source Breakdown Voltage ( $V_{GS} = 0V, I_D = 250 \mu A$ )	IRF630/D84DN2 IRF631/D84DM2	$BV_{DSS}$	200 150	— —	— —	Volts
Zero Gate Voltage Drain Current ( $V_{DS} = \text{Max Rating}, V_{GS} = 0V, T_C = 25^\circ C$ ) ( $V_{DS} = \text{Max Rating}, * 0.8, V_{GS} = 0V, T_C = 125^\circ C$ )		$I_{DSS}$	— —	— —	250 1000	$\mu A$
Gate-Source Leakage Current ( $V_{GS} = \pm 20V$ )		$I_{GSS}$	—	—	$\pm 500$	nA

on characteristics\*

Gate Threshold Voltage ( $V_{DS} = V_{GS}, I_D = 250 \mu A$ )	$T_C = 25^\circ C$	$V_{GS(TH)}$	2.0	—	4.0	Volts
On-State Drain Current ( $V_{GS} = 10V, V_{DS} = 10V$ )		$I_{D(ON)}$	9.0	—	—	A
Static Drain-Source On-State Resistance ( $V_{GS} = 10V, I_D = 5.0A$ )		$R_{DS(ON)}$	—	0.34	0.4	Ohms
Forward Transconductance ( $V_{DS} = 10V, I_D = 5.0A$ )		$g_{fs}$	2.4	3.0	—	mhos

dynamic characteristics

Input Capacitance	$V_{GS} = 0V$	$C_{iss}$	—	650	800	pF
Output Capacitance	$V_{DS} = 25V$	$C_{oss}$	—	150	450	pF
Reverse Transfer Capacitance	$f = 1 \text{ MHz}$	$C_{rss}$	—	30	150	pF

switching characteristics\*

Turn-on Delay Time	$V_{DS} = 90V$	$t_{d(on)}$	—	15	—	ns
Rise Time	$I_D = 5.0A, V_{GS} = 15V$	$t_r$	—	25	—	ns
Turn-off Delay Time	$R_{GEN} = 50\Omega, R_{GS} = 12.5\Omega$	$t_{d(off)}$	—	30	—	ns
Fall Time	$(R_{GS} \text{ (EQUIV.)} = 10\Omega)$	$t_f$	—	20	—	ns

source-drain diode ratings and characteristics\*

Continuous Source Current		$I_S$	—	—	9.0	A
Pulsed Source Current		$I_{SM}$	—	—	36.0	A
Diode Forward Voltage ( $T_C = 25^\circ C, V_{GS} = 0V, I_S = 9.0A$ )		$V_{SD}$	—	1.0	2.0	Volts
Reverse Recovery Time ( $I_S = 9.0A, di_s/dt = 100A/\mu\text{sec}, T_C = 125^\circ C$ )		$t_{rr}$ $Q_{RR}$	— —	300 2.5	— —	ns $\mu C$

IRF640 MOSFET CANAL N 200V 18A RDS 0,165 OHM VGS 10V

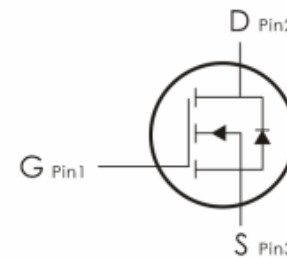
## Description:

This N-Channel MOSFET uses advanced trench technology and design to provide excellent  $R_{DS(on)}$  with low gate charge.

It can be used in a wide variety of applications.

## Features:

- 1)  $V_{DS}=200V, I_D=18A, R_{DS(on)} < 165m\Omega @ V_{GS}=10V$
- 2) Low gate charge.
- 3) Green device available.
- 4) Advanced high cell density trench technology for ultra low  $R_{DS(on)}$ .
- 5) Excellent package for good heat dissipation.



**Absolute Maximum Ratings:** ( $T_A=25^\circ\text{C}$  unless otherwise noted)

Symbol	Parameter	Ratings	Units
$V_{DS}$	Drain-Source Voltage	200	V
$V_{GS}$	Gate-Source Voltage	$\pm 20$	V
$I_D$	Continuous Drain Current- $T_C=25^\circ\text{C}^G$	18	A
	Continuous Drain Current- $T_C=100^\circ\text{C}$	13	
$E_{AS}$	Single Pulse Avalanche Energy	125	mJ
$I_{DM}$	Drain Current - Pulsed <sup>C</sup>	45	A
$I_{AS}$	Avalanche Current <sup>C</sup> (L=10mH)	9.5	A
$P_D$	Power Dissipation, $T_C=25^\circ\text{C}^B$	212	W
$T_J, T_{STG}$	Operating and Storage Junction Temperature Range	-55 to +150	$^\circ\text{C}$

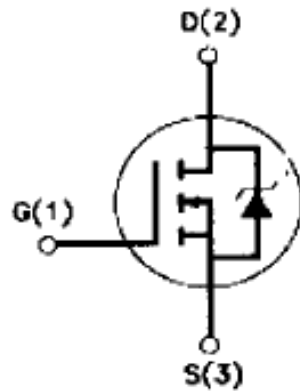
**Thermal Characteristics:**

Symbol	Parameter	Max	Units
$R_{\theta JC}$	Thermal Resistance, Junction to Case (Steady-State)	0.59	$^\circ\text{C}/\text{W}$

Electrical Characteristics: ( $T_A=25^\circ\text{C}$  unless otherwise noted)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
<b>Off Characteristics</b>						
$BV_{DSS}$	Drain-Source Breakdown Voltage	$V_{GS}=0V, I_D=250 \mu A$	200	---	---	V
$I_{DSS}$	Zero Gate Voltage Drain Current	$V_{GS}=0V, V_{DS}=200V$	---	---	1	$\mu A$
$I_{GSS}$	Gate-Source Leakage Current	$V_{GS}=\pm 20V, V_{DS}=0A$	---	---	$\pm 100$	nA
<b>On Characteristics</b>						
$V_{GS(th)}$	GATE-Source Threshold Voltage	$V_{GS}=V_{DS}, I_D=250 \mu A$	1	---	3	V
$R_{DS(on)}$	Drain-Source On Resistance	$V_{GS}=10V, I_D=1A$	---	120	165	m $\Omega$
		$V_{GS}=4.5V, I_D=1A$	---	150	180	m $\Omega$
<b>Dynamic Characteristics</b>						
$R_g$	Gate resistance	$V_{DS}=0V, V_{GS}=0V, f=1\text{MHz}$	---	6.5	---	$\Omega$
$C_{iss}$	Input Capacitance	$V_{DS}=25V, V_{GS}=0V, f=1\text{MHz}$	---	800	---	pF
$C_{oss}$	Output Capacitance		---	100	---	
$C_{rss}$	Reverse Transfer Capacitance		---	60	---	
<b>Switching Characteristics</b>						
$t_{d(on)}$	Turn-On Delay Time	$V_{GS}=10V, V_{DS}=100V,$ $R_L=5.5\Omega, R_{GEN}=3\Omega$	---	8	---	ns
$t_r$	Rise Time		---	10	---	ns
$t_{d(off)}$	Turn-Off Delay Time		---	30	---	ns
$t_f$	Fall Time		---	4	---	ns
$Q_g$	Total Gate Charge	$V_{GS}=10V, V_{DS}=100V,$ $I_D=18A$	---	27	40	nC
$Q_{gs}$	Gate-Source Charge		---	7	---	nC
$Q_{gd}$	Gate-Drain "Miller" Charge		---	3	---	nC
<b>Drain-Source Diode Characteristics</b>						

IRF840 MOSFET CANAL N 8A 500V 0,85 OHM



PIN 1.Gate  
2.Drain  
3.Source  
TO-220C package

SYMBOL	PARAMETER	VALUE	UNIT
$V_{DSS}$	Drain-Source Voltage	500	V
$V_{GS}$	Gate-Source Voltage-Continuous	$\pm 20$	V
$I_D$	Drain Current-Continuous	8	A
$I_{DM}$	Drain Current-Single Plused	32	A
$P_D$	Total Dissipation @ $T_C=25^\circ C$	125	W
$T_j$	Max. Operating Junction Temperature	150	$^\circ C$
$T_{stg}$	Storage Temperature	-55~150	$^\circ C$

• THERMAL CHARACTERISTICS

SYMBOL	PARAMETER
$R_{th j-c}$	Thermal Resistance, Junction to case
$R_{th j-a}$	Thermal Resistance, Junction to ambient

## IRLZ44 47A 47A BAIXA TENSÃO DE GATE

International  
**IR** Rectifier

PD - 9.1346B

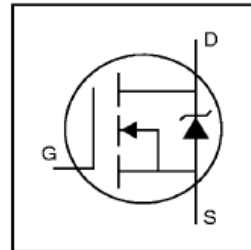
**IRLZ44N**HEXFET<sup>®</sup> Power MOSFET

- Logic-Level Gate Drive
- Advanced Process Technology
- Dynamic dv/dt Rating
- 175°C Operating Temperature
- Fast Switching
- Fully Avalanche Rated

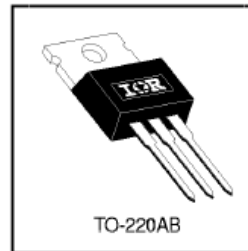
**Description**

Fifth Generation HEXFETs from International Rectifier utilize advanced processing techniques to achieve the lowest possible on-resistance per silicon area. This benefit, combined with the fast switching speed and ruggedized device design that HEXFET Power MOSFETs are well known for, provides the designer with an extremely efficient device for use in a wide variety of applications.

The TO-220 package is universally preferred for all commercial-industrial applications at power dissipation levels to approximately 50 watts. The low thermal resistance and low package cost of the TO-220 contribute to its wide acceptance throughout the industry.



$V_{DSS} = 55V$
$R_{DS(on)} = 0.022\Omega$
$I_D = 47A$



**Absolute Maximum Ratings**

	Parameter	Max.	Units
$I_D @ T_C = 25^\circ\text{C}$	Continuous Drain Current, $V_{GS} @ 10\text{V}$	47	A
$I_D @ T_C = 100^\circ\text{C}$	Continuous Drain Current, $V_{GS} @ 10\text{V}$	33	
$I_{DM}$	Pulsed Drain Current ①	160	
$P_D @ T_C = 25^\circ\text{C}$	Power Dissipation	110	W
	Linear Derating Factor	0.71	W/°C
$V_{GS}$	Gate-to-Source Voltage	$\pm 16$	V
$E_{AS}$	Single Pulse Avalanche Energy ②	210	mJ
$I_{AR}$	Avalanche Current ①	25	A
$E_{AR}$	Repetitive Avalanche Energy ①	11	mJ
$dv/dt$	Peak Diode Recovery $dv/dt$ ③	5.0	V/ns
$T_J$	Operating Junction and	-55 to + 175	°C
$T_{STG}$	Storage Temperature Range		
	Soldering Temperature, for 10 seconds	300 (1.6mm from case)	
	Mounting torque, 6-32 or M3 screw.	10 lbf•in (1.1N•m)	

**Thermal Resistance**

	Parameter	Min.	Typ.	Max.	Units
$R_{\theta JC}$	Junction-to-Case	—	—	1.4	°C/W
$R_{\theta CS}$	Case-to-Sink, Flat, Greased Surface	—	0.50	—	
$R_{\theta JA}$	Junction-to-Ambient	—	—	62	



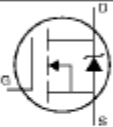
## IRLZ44N

International  
TOR RectifierElectrical Characteristics @  $T_J = 25^\circ\text{C}$  (unless otherwise specified)

	Parameter	Min.	Typ.	Max.	Units	Conditions
$V_{(BR)DSS}$	Drain-to-Source Breakdown Voltage	55	—	—	V	$V_{GS} = 0V, I_D = 250\mu A$
$\Delta V_{(BR)DSS}/\Delta T_J$	Breakdown Voltage Temp. Coefficient	—	0.070	—	V/ $^\circ\text{C}$	Reference to $25^\circ\text{C}, I_D = 1\text{mA}$
$R_{DS(on)}$	Static Drain-to-Source On-Resistance	—	—	0.022	$\Omega$	$V_{GS} = 10V, I_D = 25A$ ③
		—	—	0.025		$V_{GS} = 5.0V, I_D = 25A$ ④
		—	—	0.035		$V_{GS} = 4.0V, I_D = 21A$ ④
$V_{GS(th)}$	Gate Threshold Voltage	1.0	—	2.0	V	$V_{DS} = V_{GS}, I_D = 250\mu A$
$g_s$	Forward Transconductance	21	—	—	S	$V_{DS} = 25V, I_D = 25A$
$I_{DSS}$	Drain-to-Source Leakage Current	—	—	25	$\mu A$	$V_{DS} = 55V, V_{GS} = 0V$
		—	—	250		$V_{DS} = 44V, V_{GS} = 0V, T_J = 150^\circ\text{C}$
$I_{GSS}$	Gate-to-Source Forward Leakage	—	—	100	nA	$V_{GS} = 16V$
	Gate-to-Source Reverse Leakage	—	—	-100		$V_{GS} = -16V$
$Q_g$	Total Gate Charge	—	—	48	nC	$I_D = 25A$
$Q_{gs}$	Gate-to-Source Charge	—	—	8.6		$V_{DS} = 44V$
$Q_{gd}$	Gate-to-Drain ("Miller") Charge	—	—	25		$V_{GS} = 5.0V$ , See Fig. 6 and 13 ④
$t_{d(on)}$	Turn-On Delay Time	—	11	—	ns	$V_{DD} = 28V$
$t_r$	Rise Time	—	84	—		$I_D = 25A$
$t_{d(off)}$	Turn-Off Delay Time	—	26	—		$R_G = 3.4\Omega, V_{GS} = 5.0V$
$t_f$	Fall Time	—	15	—		$R_D = 1.1\Omega$ , See Fig. 10 ④
$L_D$	Internal Drain Inductance	—	4.5	—	nH	Between lead, 6mm (0.25in.) from package and center of die contact
$L_S$	Internal Source Inductance	—	7.5	—		
$C_{iss}$	Input Capacitance	—	1700	—	pF	$V_{GS} = 0V$
$C_{oss}$	Output Capacitance	—	400	—		$V_{DS} = 25V$
$C_{rss}$	Reverse Transfer Capacitance	—	150	—		$f = 1.0\text{MHz}$ , See Fig. 5



### Source-Drain Ratings and Characteristics

	Parameter	Min.	Typ.	Max.	Units	Conditions
$I_S$	Continuous Source Current (Body Diode)	—	—	47	A	MOSFET symbol showing the integral reverse p-n junction diode. 
$I_{SM}$	Pulsed Source Current (Body Diode) ①	—	—	160		
$V_{SD}$	Diode Forward Voltage	—	—	1.3	V	$T_J = 25^\circ\text{C}$ , $I_S = 25\text{A}$ , $V_{GS} = 0\text{V}$ ②
$t_{rr}$	Reverse Recovery Time	—	80	120	ns	$T_J = 25^\circ\text{C}$ , $I_F = 25\text{A}$
$Q_{rr}$	Reverse Recovery Charge	—	210	320	nC	$di/dt = 100\text{A}/\mu\text{s}$ ③
$t_{on}$	Forward Turn-On Time	Intrinsic turn-on time is negligible (turn-on is dominated by $L_S+L_D$ )				

IRFZ44N CNAL N BAIXÍSSIMO RDS 0,022OHM BAIXA TENSÃO 55V ALTA CORRENTE 49 A ISOLAÇÃO 2KV

Philips Semiconductors

Product specification

**N-channel enhancement mode  
TrenchMOS™ transistor**

**IRFZ44N**

**GENERAL DESCRIPTION**

N-channel enhancement mode standard level field-effect power transistor in a plastic envelope using 'trench' technology. The device features very low on-state resistance and has integral zener diodes giving ESD protection up to 2kV. It is intended for use in switched mode power supplies and general purpose switching applications.

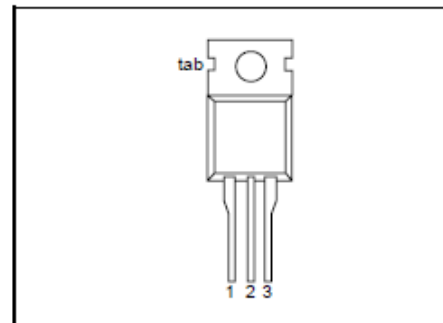
**QUICK REFERENCE DATA**

SYMBOL	PARAMETER	MAX.	UNIT
$V_{DS}$	Drain-source voltage	55	V
$I_D$	Drain current (DC)	49	A
$P_{tot}$	Total power dissipation	110	W
$T_j$	Junction temperature	175	°C
$R_{DS(ON)}$	Drain-source on-state resistance $V_{GS} = 10 V$	22	mΩ

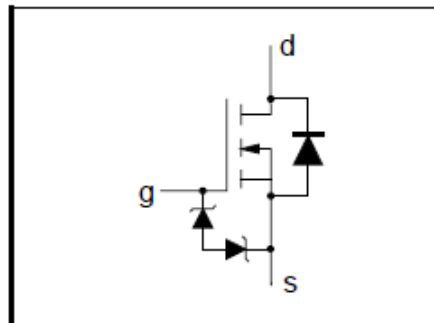
**PINNING - TO220AB**

PIN	DESCRIPTION
1	gate
2	drain
3	source
tab	drain

**PIN CONFIGURATION**



**SYMBOL**



**LIMITING VALUES**

Limiting values in accordance with the Absolute Maximum System (IEC 134)

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_{DS}$	Drain-source voltage	-	-	55	V
$V_{DGR}$	Drain-gate voltage	$R_{GS} = 20 \text{ k}\Omega$	-	55	V
$\pm V_{GS}$	Gate-source voltage	-	-	20	V
$I_D$	Drain current (DC)	$T_{mb} = 25 \text{ }^\circ\text{C}$	-	49	A
$I_D$	Drain current (DC)	$T_{mb} = 100 \text{ }^\circ\text{C}$	-	35	A
$I_{DM}$	Drain current (pulse peak value)	$T_{mb} = 25 \text{ }^\circ\text{C}$	-	160	A
$P_{tot}$	Total power dissipation	$T_{mb} = 25 \text{ }^\circ\text{C}$	-	110	W
$T_{stg}, T_j$	Storage & operating temperature	-	- 55	175	$^\circ\text{C}$

**ESD LIMITING VALUE**

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_C$	Electrostatic discharge capacitor voltage, all pins	Human body model (100 pF, 1.5 k $\Omega$ )	-	2	kV

**THERMAL RESISTANCES**

SYMBOL	PARAMETER	CONDITIONS	TYP.	MAX.	UNIT
$R_{th\ j-mb}$	Thermal resistance junction to mounting base	-	-	1.4	K/W
$R_{th\ j-a}$	Thermal resistance junction to ambient	in free air	60	-	K/W

**STATIC CHARACTERISTICS**

$T_j = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{(BR)DSS}$	Drain-source breakdown voltage	$V_{GS} = 0\text{ V}; I_D = 0.25\text{ mA}; T_j = -55^\circ\text{C}$	55	-	-	V
$V_{GS(TO)}$	Gate threshold voltage	$V_{DS} = V_{GS}; I_D = 1\text{ mA}; T_j = 175^\circ\text{C}$	2.0	3.0	4.0	V
		$T_j = -55^\circ\text{C}$	1.0	-	-	V
		$T_j = 175^\circ\text{C}$	-	-	4.4	-
$I_{DSS}$	Zero gate voltage drain current	$V_{DS} = 55\text{ V}; V_{GS} = 0\text{ V}; T_j = 175^\circ\text{C}$	-	0.05	10	$\mu\text{A}$
$I_{GSS}$	Gate source leakage current	$V_{GS} = \pm 10\text{ V}; V_{DS} = 0\text{ V}; T_j = 175^\circ\text{C}$	-	0.04	1	$\mu\text{A}$
		$T_j = 175^\circ\text{C}$	-	-	20	$\mu\text{A}$
$\pm V_{(BR)GSS}$	Gate source breakdown voltage	$I_G = \pm 1\text{ mA}; V_{GS} = 10\text{ V}; I_D = 25\text{ A}; T_j = 175^\circ\text{C}$	16	-	-	V
$R_{DS(ON)}$	Drain-source on-state resistance	$T_j = 175^\circ\text{C}$	-	15	22	$\text{m}\Omega$
			-	-	42	$\text{m}\Omega$

**DYNAMIC CHARACTERISTICS**

$T_{mb} = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$g_{fs}$	Forward transconductance	$V_{DS} = 25\text{ V}; I_D = 25\text{ A}$	6	-	-	S
$C_{iss}$	Input capacitance	$V_{GS} = 0\text{ V}; V_{DS} = 25\text{ V}; f = 1\text{ MHz}$	-	1350	1800	pF
$C_{oss}$	Output capacitance		-	330	400	pF
$C_{rss}$	Feedback capacitance		-	155	215	pF
$Q_g$	Total gate charge	$V_{DD} = 44\text{ V}; I_D = 50\text{ A}; V_{GS} = 10\text{ V}$	-	-	62	nC
$Q_{gs}$	Gate-source charge		-	-	15	nC
$Q_{gd}$	Gate-drain (miller) charge		-	-	26	nC
$t_{d\text{ on}}$	Turn-on delay time	$V_{DD} = 30\text{ V}; I_D = 25\text{ A}; V_{GS} = 10\text{ V}; R_G = 10\ \Omega$ Resistive load	-	18	26	ns
$t_r$	Turn-on rise time		-	50	75	ns
$t_{d\text{ off}}$	Turn-off delay time		-	40	50	ns
$t_f$	Turn-off fall time		-	30	40	ns
$L_d$	Internal drain inductance	Measured from contact screw on tab to centre of die	-	3.5	-	nH
$L_d$	Internal drain inductance	Measured from drain lead 6 mm from package to centre of die	-	4.5	-	nH
$L_s$	Internal source inductance	Measured from source lead 6 mm from package to source bond pad	-	7.5	-	nH

**REVERSE DIODE LIMITING VALUES AND CHARACTERISTICS** $T_j = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{DR}$	Continuous reverse drain current		-	-	49	A
$I_{DRM}$	Pulsed reverse drain current		-	-	160	A
$V_{SD}$	Diode forward voltage	$I_F = 25\text{ A}; V_{GS} = 0\text{ V}$ $I_F = 40\text{ A}; V_{GS} = 0\text{ V}$	-	0.95 1.0	1.2 -	V
$t_{rr}$	Reverse recovery time	$I_F = 40\text{ A}; -dI_F/dt = 100\text{ A}/\mu\text{s};$ $V_{GS} = -10\text{ V}; V_R = 30\text{ V}$	-	47	-	ns
$Q_{rr}$	Reverse recovery charge		-	0.15	-	$\mu\text{C}$

GT40T 40A 1500V PARA ALTAS POTENCIAS

**TOSHIBA**

**GT40T101**

TOSHIBA INSULATED GATE BIPOLAR TRANSISTOR SILICON N CHANNEL MOS TYPE

**GT40T101**

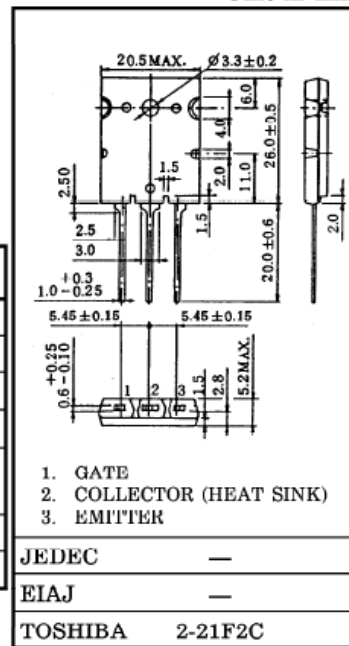
HIGH POWER SWITCHING APPLICATIONS.

- Enhancement-Mode
- High Speed :  $t_f=0.4\mu s$  (Max.) ( $I_C=40A$ )
- Low Saturation :  $V_{CE(sat)}=5.0V$  (Max.) ( $I_C=40A$ )

MAXIMUM RATINGS ( $T_a = 25^\circ C$ )

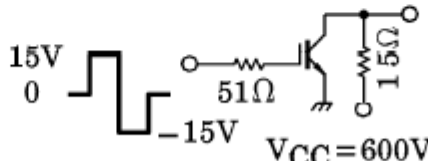
CHARACTERISTIC	SYMBOL	RATING	UNIT
Collector-Emitter Voltage	$V_{CES}$	1500	V
Gate-Emitter Voltage	$V_{GES}$	$\pm 25$	V
Collector Current	DC	$I_C$	40 A
Collector Current	1ms	$I_{CP}$	80 A
Collector Power Dissipation	( $T_c = 25^\circ C$ )	$P_C$	200 W
Junction Temperature	$T_j$	150	$^\circ C$
Storage Temperature Range	$T_{stg}$	-55~150	$^\circ C$

Unit in mm



Weight : 9.75g

ELECTRICAL CHARACTERISTICS (Ta = 25°C)

CHARACTERISTIC	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.	UNIT	
Gate Leakage Current	$I_{GES}$	$V_{GE} = \pm 25V, V_{CE} = 0$	—	—	$\pm 500$	nA	
Collector Cut-off Current	$I_{CES}$	$V_{CE} = 1500V, V_{GE} = 0$	—	—	1.0	mA	
Gate-Emitter Cut-off Voltage	$V_{GE(OFF)}$	$I_C = 40mA, V_{CE} = 5V$	3.0	—	6.0	V	
Collector-Emitter Saturation Voltage	$V_{CE(sat)}$	$I_C = 40A, V_{GE} = 15V$	—	4.0	5.0	V	
Input Capacitance	$C_{ies}$	$V_{CE} = 10V, V_{GE} = 0, f = 1MHz$	—	3600	—	pF	
Switching Time	Rise Time	$t_r$		—	0.6	1.0	$\mu s$
	Turn-on Time	$t_{on}$		—	0.7	1.1	
	Fall Time	$t_f$		—	0.2	0.4	
	Turn-off Time	$t_{off}$		—	0.5	1.0	
Thermal Resistance	$R_{th(j-c)}$	—	—	—	0.625	$^{\circ}C / W$	

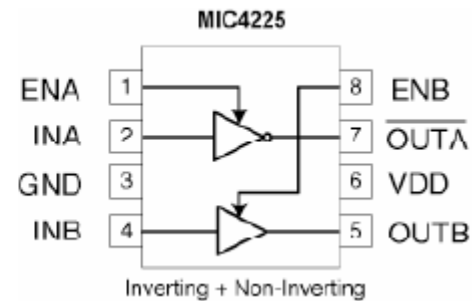
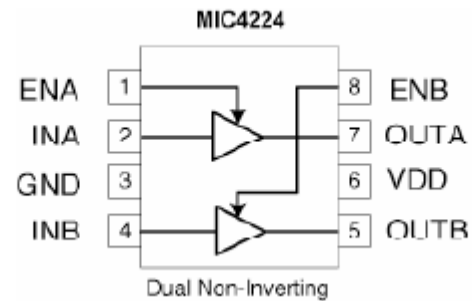
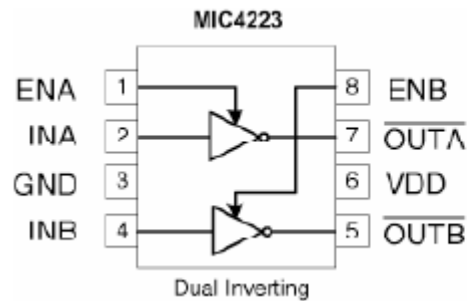
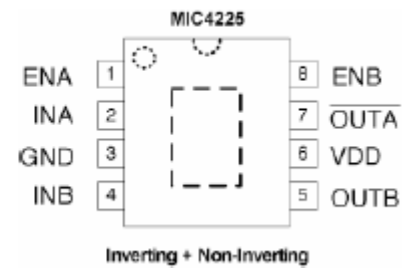
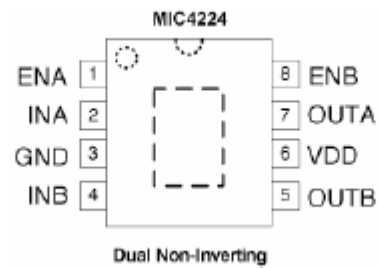
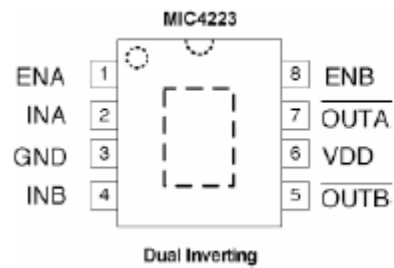


DRIVER MOSFET IGBT.

MIC4223/MIC4224/MIC4225 DRIVER SIMPLES

**MIC4223/MIC4224/MIC4225**

Dual 4A, 4.5V to 18V, 15ns Switch Time,  
Low-Side MOSFET Drivers with Enable

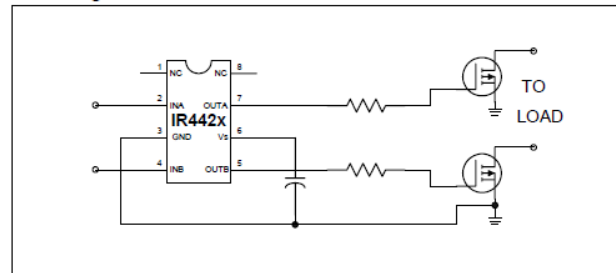


IR4426/IR4427/IR4448 DRIVER SIMPLES COM DUAS UNIDADES

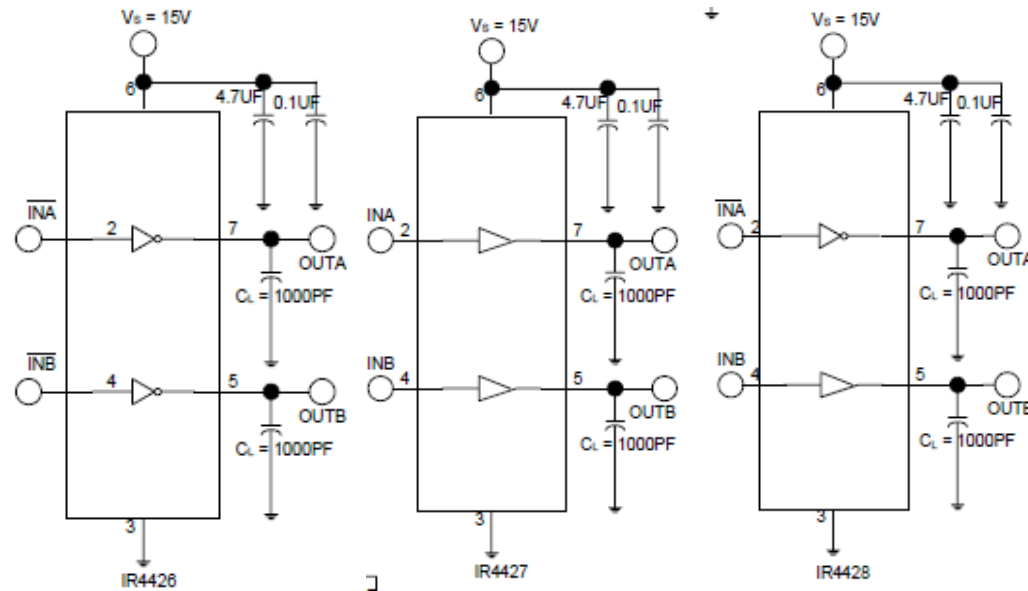
DUAL LOW SIDE DRIVER

Product Summary

$I_{O+/-}$	1.5A / 1.5A
$V_{OUT}$	6V - 20V
$t_{on/off}$ (typ.)	85 & 65 ns

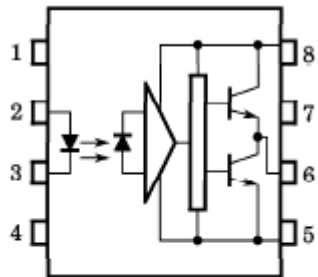


<p>8 Lead PDIP</p> <p>IR4426</p>	<p>8 Lead PDIP</p> <p>IR4427</p>	<p>8 Lead PDIP</p> <p>IR4428</p>
Part Number		

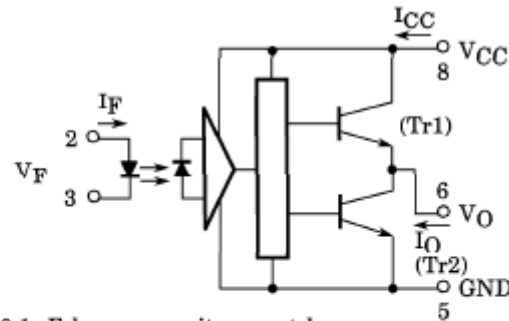


TLP251 OPTO DRIVER SIMPLES OPTOACOPLADO

PIN CONFIGURATION (TOP VIEW)



- 1 : N.C.
- 2 : ANODE
- 3 : CATHODE
- 4 : N.C.
- 5 : GND
- 6 :  $V_O$  (OUTPUT)
- 7 : N.C.
- 8 :  $V_{CC}$



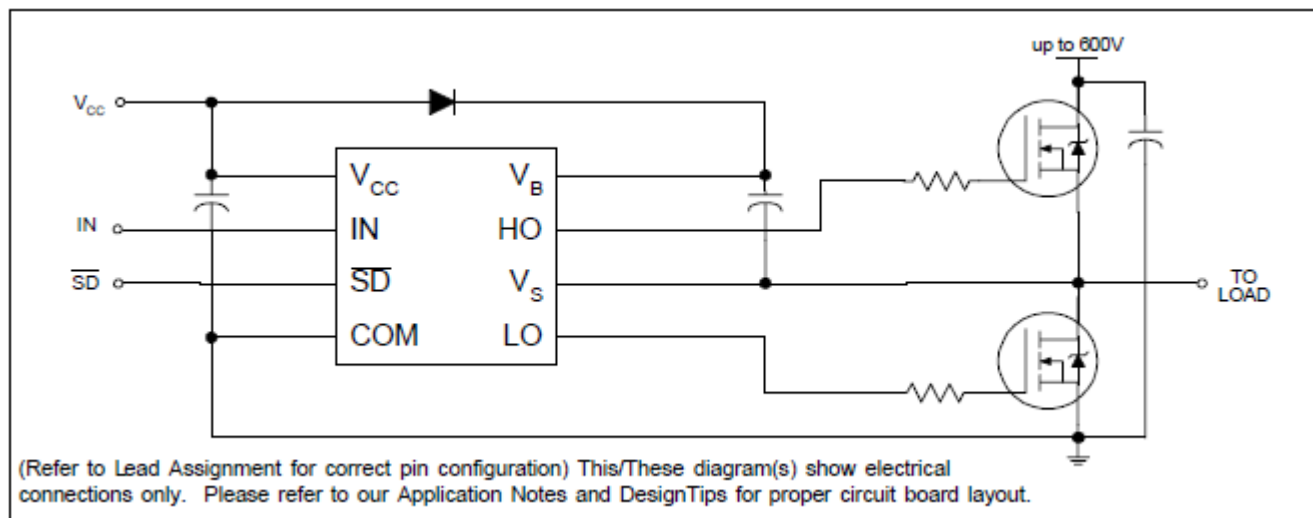
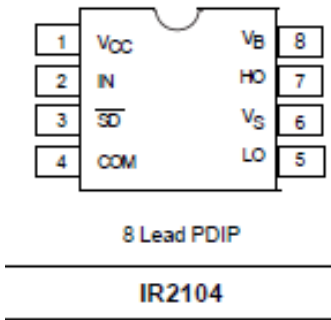
A  $0.1\mu F$  bypass capacitor must be connected between pin 8 and 5 (See Note 5).

TRUTH TABLE

		Tr1	Tr2
		ON	OFF
Input LED	ON	ON	OFF
	OFF	OFF	ON

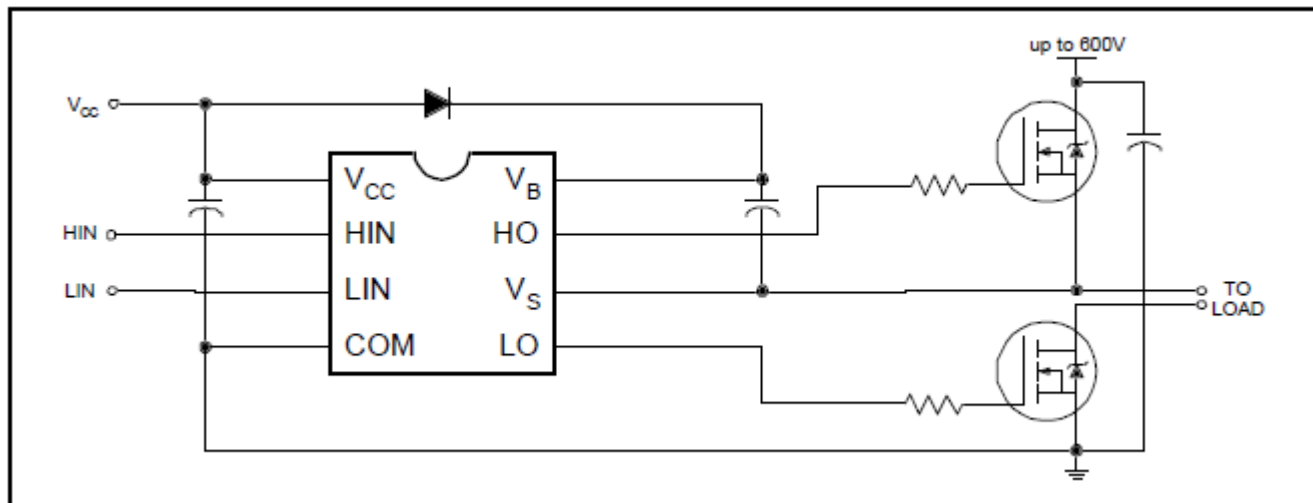
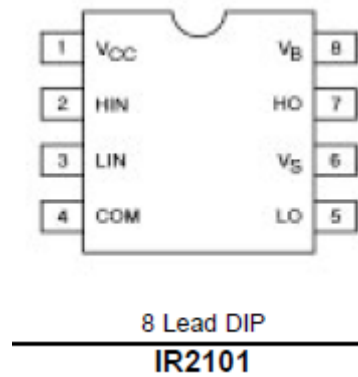
IR2104 DRIVER DUPLO COM BOOTSTRAPE ENTRADAS IN/SD

$V_{OFFSET}$	600V max.
$I_{O+/-}$	130 mA / 270 mA
$V_{OUT}$	10 - 20V
$t_{on/off}$ (typ.)	680 & 150 ns
Deadtime (typ.)	520 ns



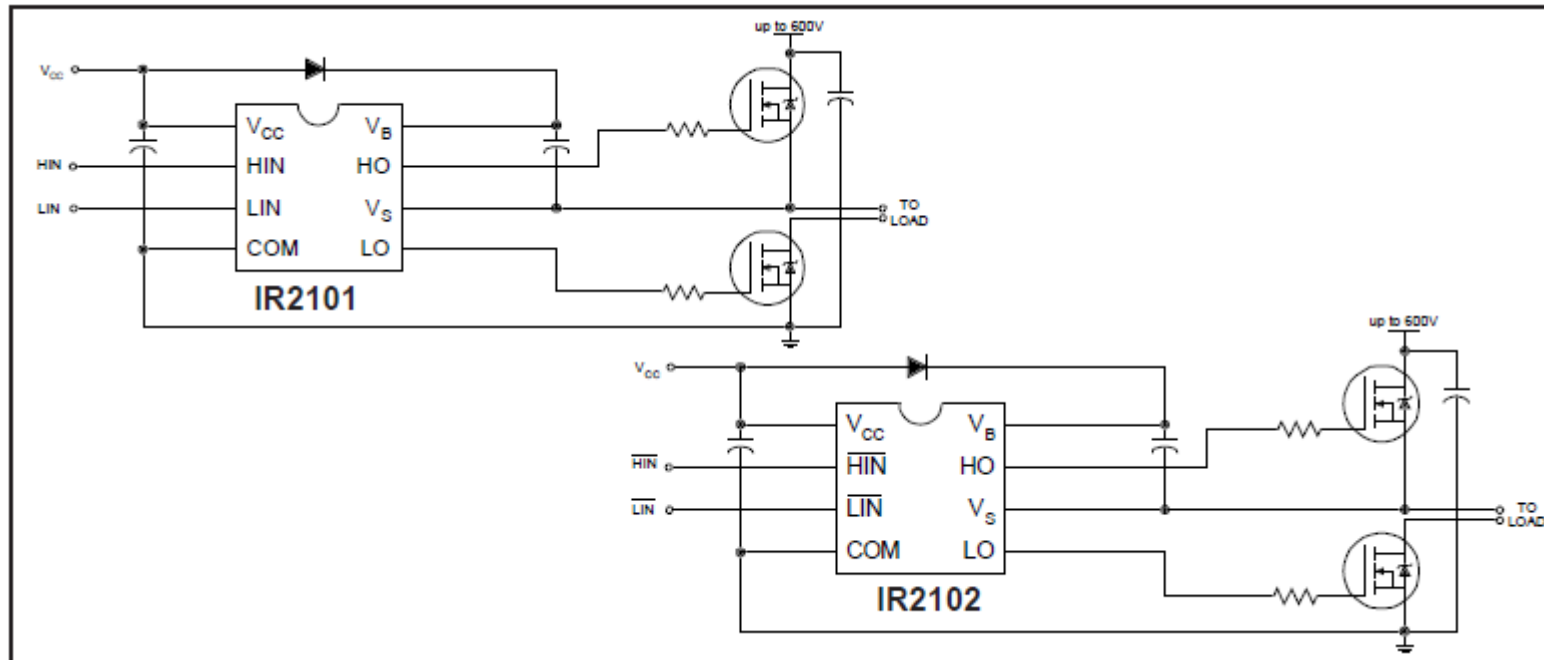
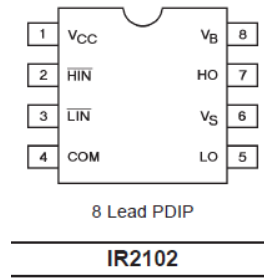
IR2101 DRIVER DUPLO COM BOOTSTRAPE ENTRADAS HIN LIN

<b>V<sub>OFFSET</sub></b>	<b>600V max.</b>
<b>I<sub>O+/-</sub></b>	<b>100 mA / 210 mA</b>
<b>V<sub>OUT</sub></b>	<b>10 - 20V</b>
<b>t<sub>on/off</sub> (typ.)</b>	<b>130 &amp; 90 ns</b>
<b>Delay Matching</b>	<b>30 ns</b>

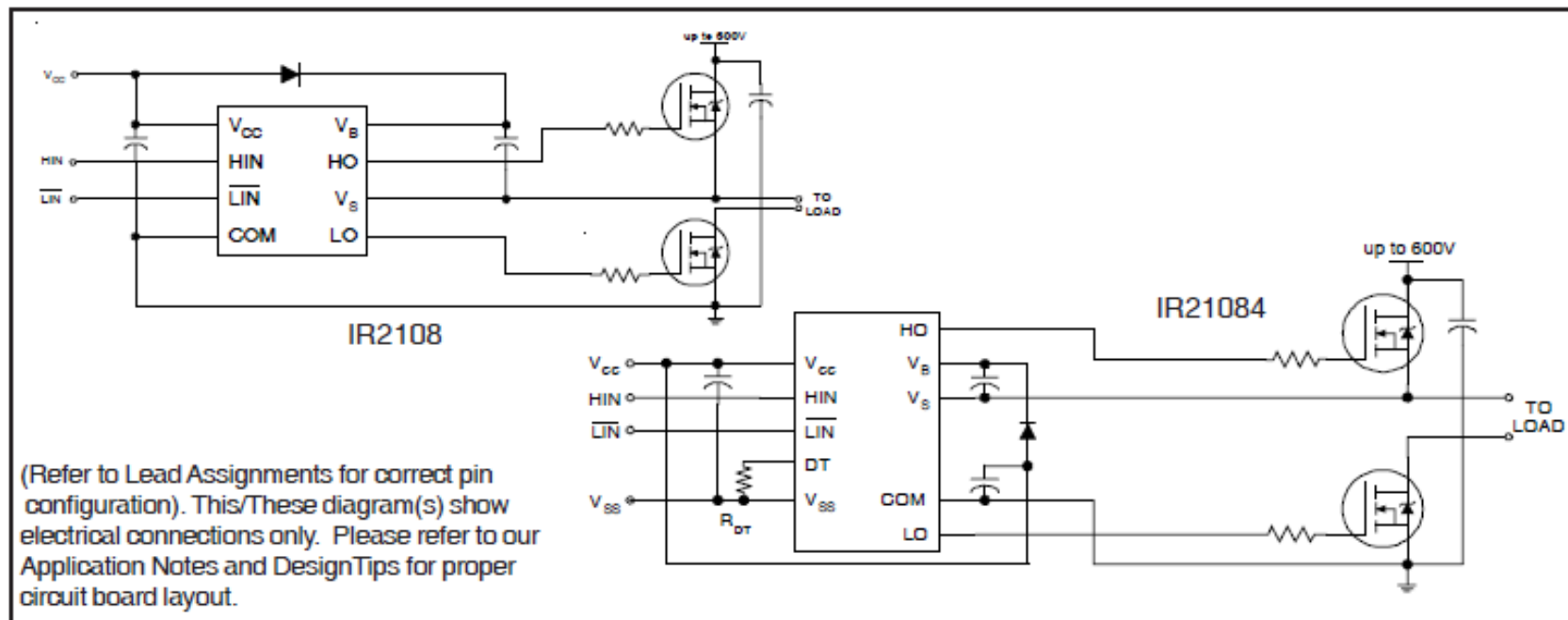
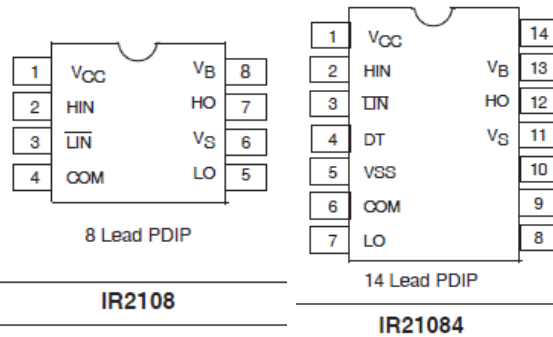


IR2102 DRIVER DUPLO COM BOOTSTRAPE ENTRADAS HIN/LIN INVERTIDAS

V <sub>OFFSET</sub>	600V max.
I <sub>O+/-</sub>	130 mA / 270 mA
V <sub>OUT</sub>	10 - 20V
t <sub>on/off (typ.)</sub>	160 & 150 ns
Delay Matching	50 ns



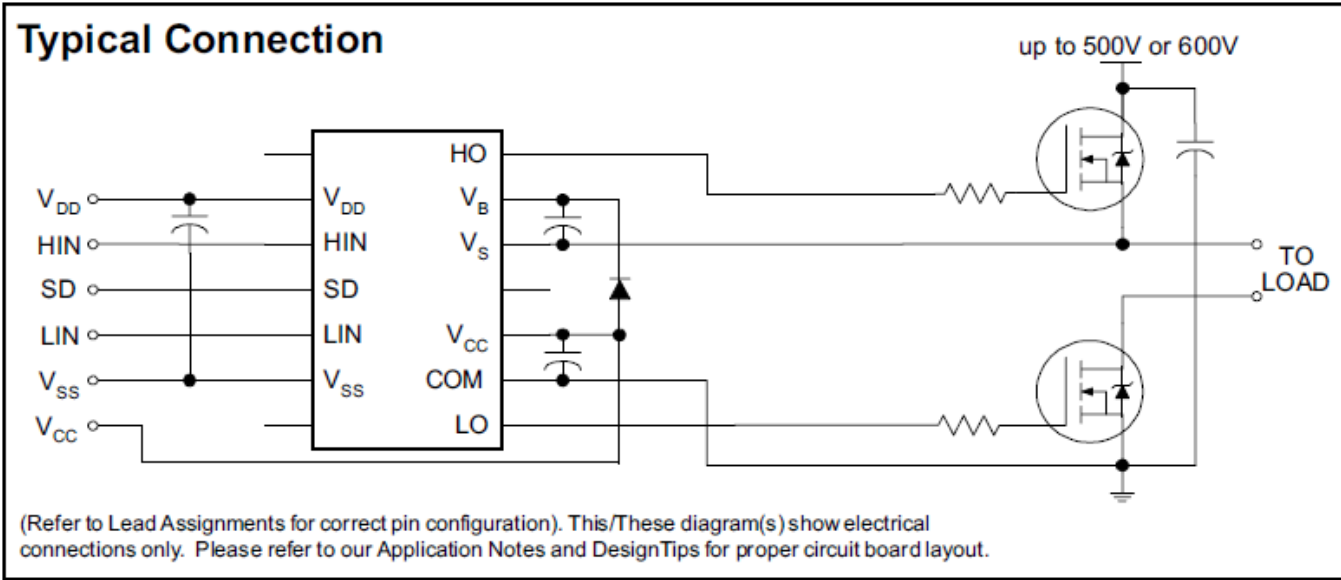
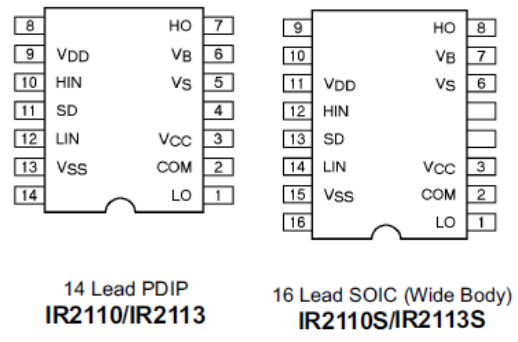
IR2108/IR21084 DRIVER DUPLO COM BOOTSTRAPE COM DEAD TIME ENTRADAS HIN/LIN O 1084 INVERTIDA





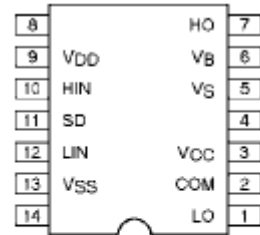
IR2110/IR2113 DRIVER DUPLO COM BOOTSTRAP ENTRADAS HIN/LIN/SD TERRAS SEPARADOS O MAIS USADO

$V_{OFFSET}$ (IR2110)	500V max.
(IR2113)	600V max.
$I_{O+/-}$	2A / 2A
$V_{OUT}$	10 - 20V
$t_{on/off}$ (typ.)	120 & 94 ns
Delay Matching (IR2110)	10 ns max.
(IR2113)	20ns max.



IR2112 DRIVER DUPLO COM BOOTSTRAPE SIMILAR AO 2110 MAS COM BAIXA CORRENTE DE GATE

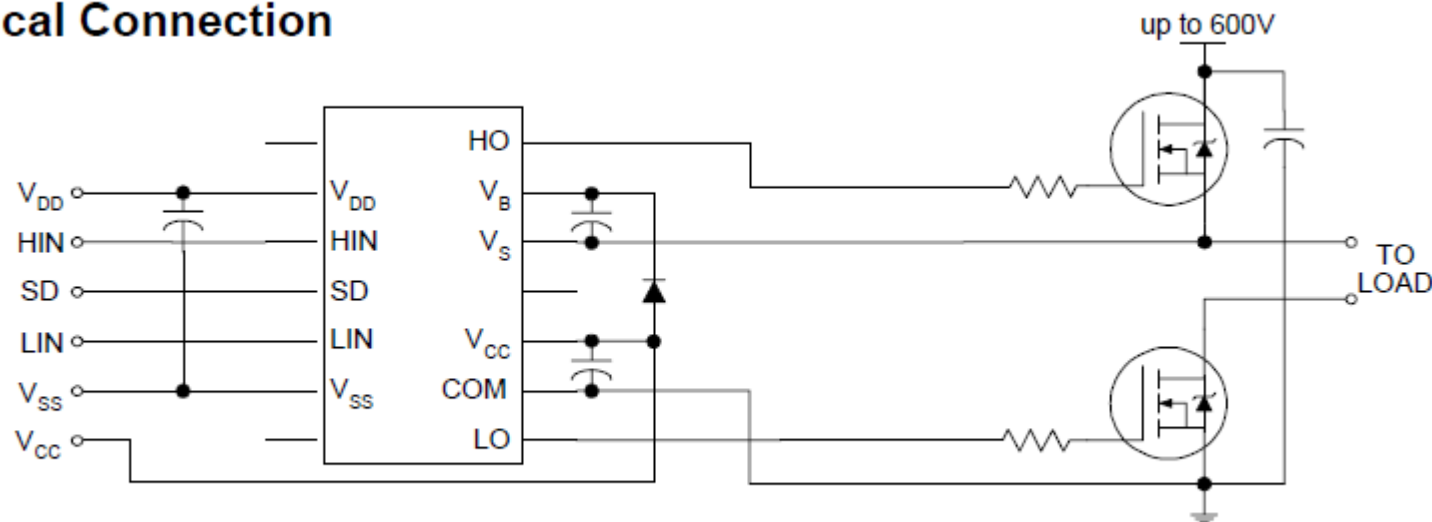
$V_{OFFSET}$	600V max.
$I_{O+/-}$	200 mA / 420 mA
$V_{OUT}$	10 - 20V
$t_{on/off}$ (typ.)	125 & 105 ns
Delay Matching	30 ns



14 Lead DIP

**IR2112**

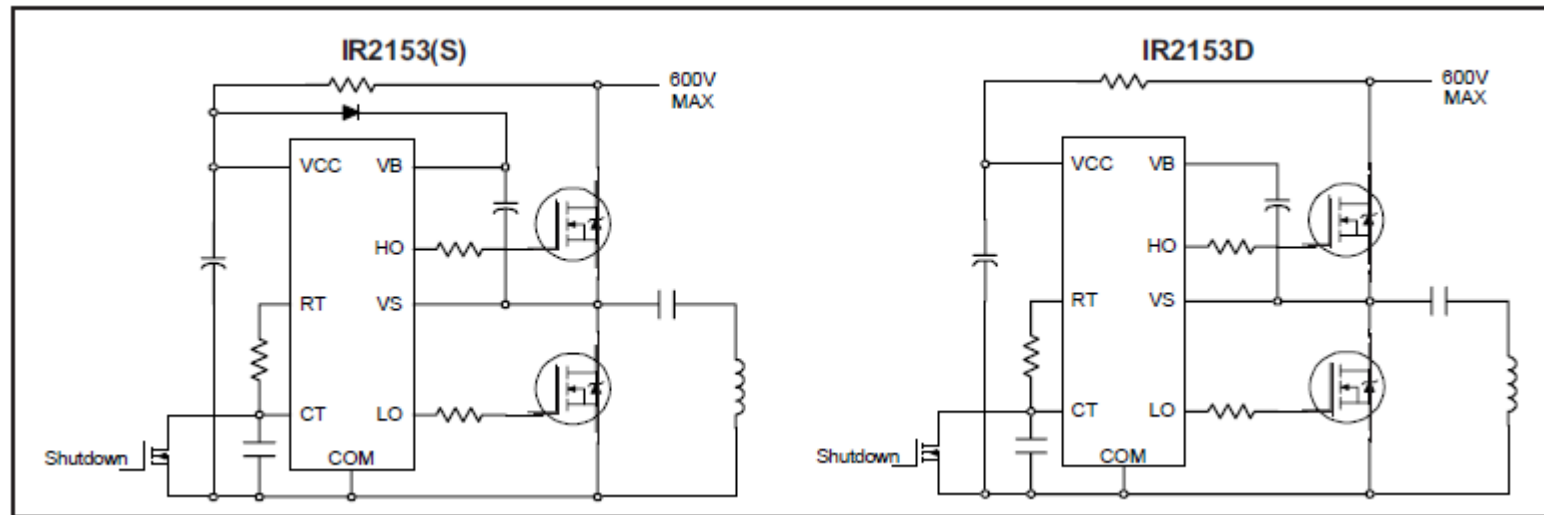
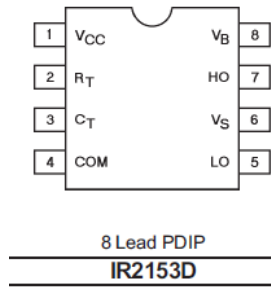
**Typical Connection**



(Refer to Lead Assignments for correct pin configuration). This/These diagram(s) show electrical connections only. Please refer to our Application Notes and Design Tips for proper circuit board layout.

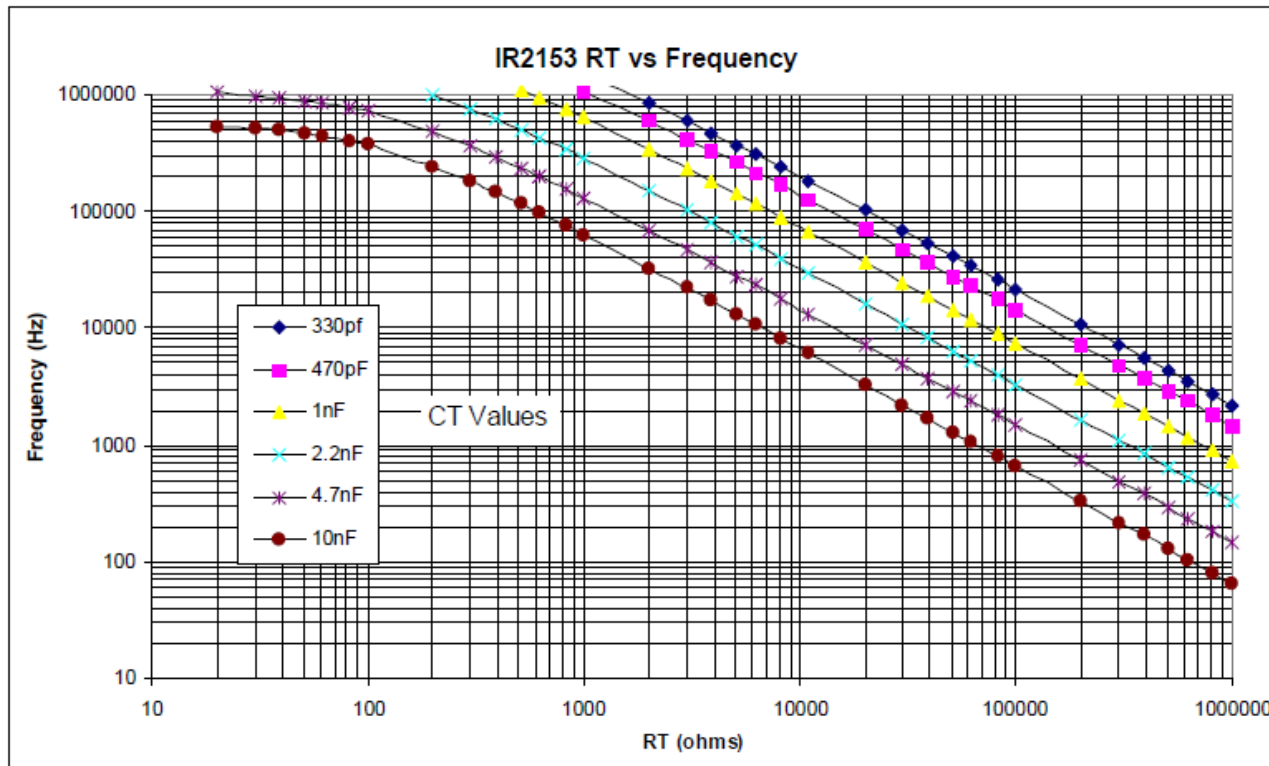
IR2153 DRIVER DUPLO COM BOOTSTRAPE SELF OSCILATION MELHORADO DO 2155

V <sub>OFFSET</sub>	600V max.
Duty Cycle	50%
T <sub>r</sub> /T <sub>p</sub>	80/40ns
V <sub>clamp</sub>	15.6V
Deadtime (typ.)	1.2 μs



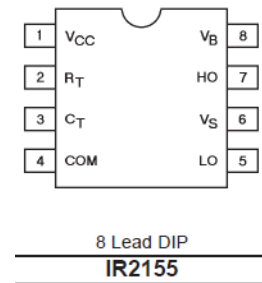
### Recommended Component Values

Symbol	Component	Min.	Max.	Units
$R_T$	Timing resistor value	10	—	$k\Omega$
$C_T$	$C_T$ pin capacitor value	330	—	$\mu F$



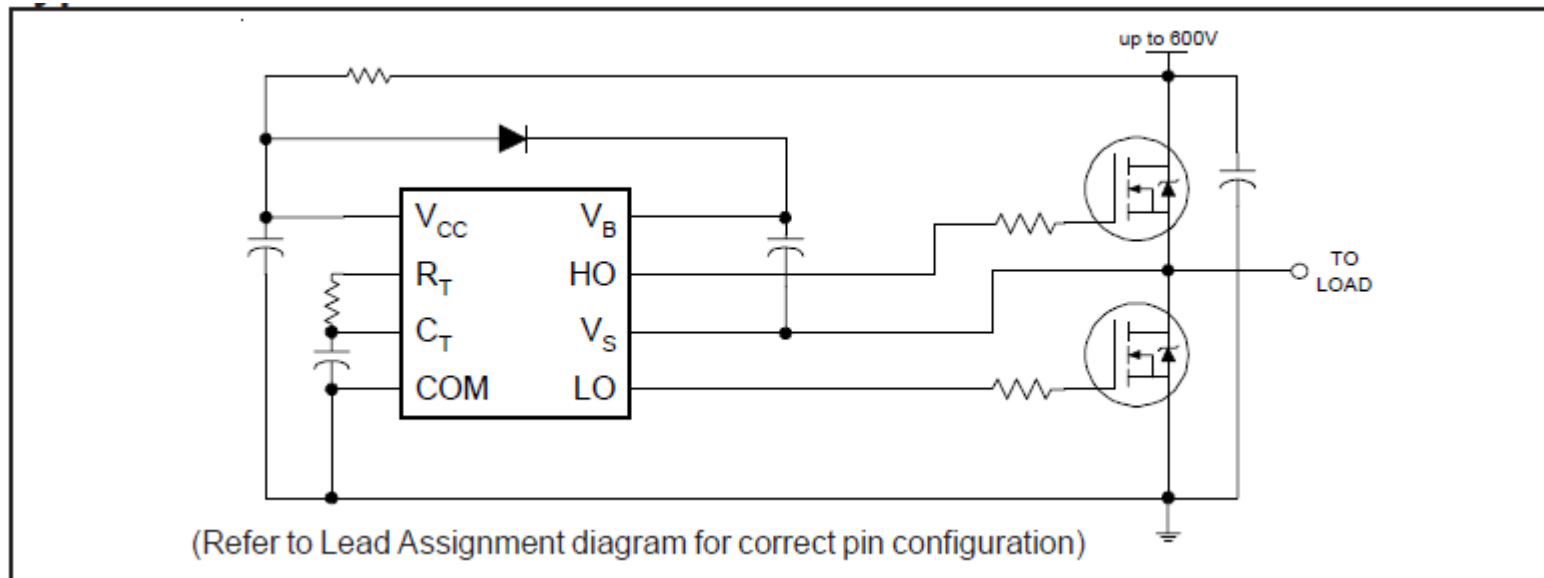
IR2155 DRIVER DUPLO COM BOOTSTRAPE SELF OSCILATION ANTIGO AGORA USAR O IR2153

V <sub>OFFSET</sub>	600V max.
Duty Cycle	50%
I <sub>O+/-</sub>	210 mA / 420 mA
V <sub>OUT</sub>	10 - 20V
Deadtime (typ.)	1.2 μs

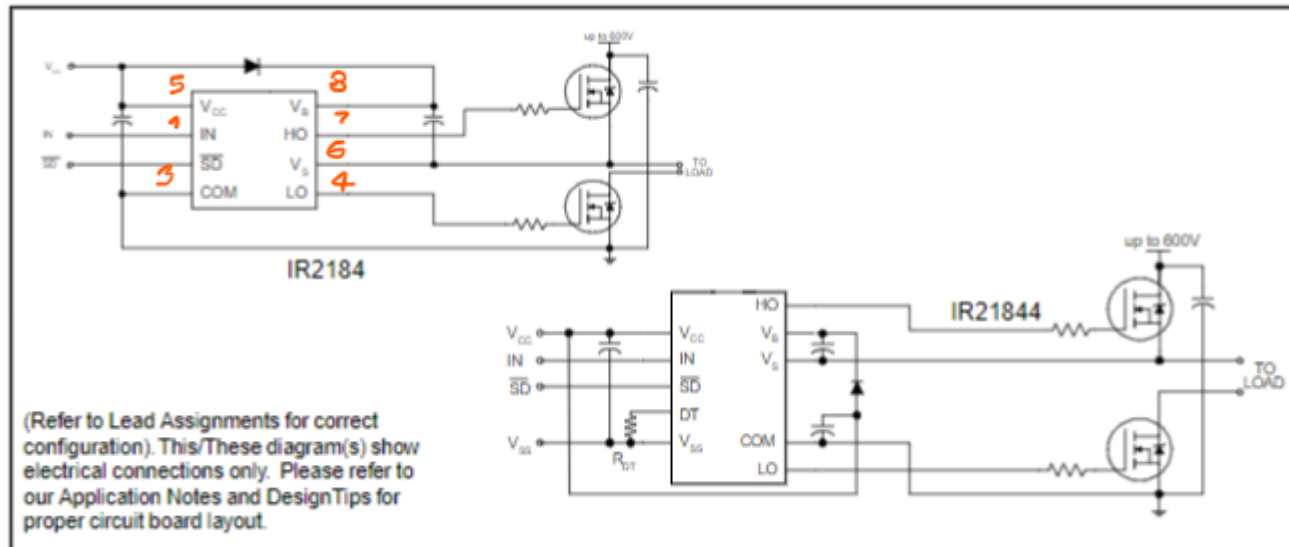


● Programmable oscillator frequency

$$f = \frac{1}{1.4 \times (R_T + 150\Omega) \times C_T}$$



IR2184



(Refer to Lead Assignments for correct configuration). This/These diagram(s) show electrical connections only. Please refer to our Application Notes and DesignTips for proper circuit board layout.

IGBT

DRIVER IRGB14C40L IRGSL14C40L IRGS14C40L IGNITION IGBT 14A

PD - 93891A

International  
**IR** Rectifier

**Ignition IGBT**

IRGS14C40L  
IRGSL14C40L  
IRGB14C40L

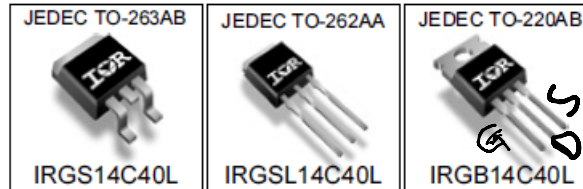
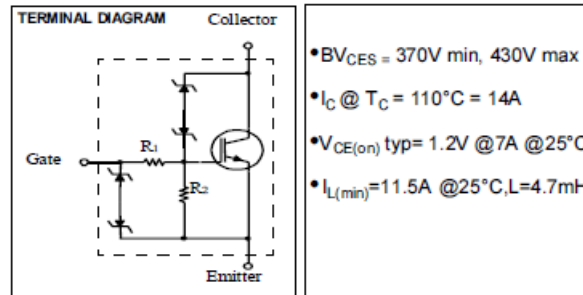
IGBT with on-chip Gate-Emitter and Gate-Collector clamps

**Features**

- Most Rugged in Industry
- Logic-Level Gate Drive
- > 6KV ESD Gate Protection
- Low Saturation Voltage
- High Self-clamped Inductive Switching Energy

**Description**

The advanced IGBT process family includes a MOS gated, N-channel logic level device which is intended for coil-on-plug automotive ignition applications and small-engine ignition circuits. Unique features include on-chip active voltage clamps between the Gate-Emitter and Gate-Collector which provide over voltage protection capability in ignition circuits.



NOTE: IRGS14C40L is available in tape and reel. Add a suffix of TRR or TRL to the part number to determine the orientation of the device in the pocket, i.e, IRGS14C40LTRR or IRGS14C40LTRL.



**Absolute Maximum Ratings**

	Parameter	Max	Unit	Condition
$V_{CES}$	Collector-to-Emitter Voltage	Clamped	V	$R_G = 1K \text{ ohm}$
$I_C @ T_C = 25^\circ\text{C}$	Continuous Collector Current	20	A	$V_{GE} = 5V$
$I_C @ T_C = 110^\circ\text{C}$	Continuous Collector Current	14	A	$V_{GE} = 5V$
$I_G$	Continuous Gate Current	1	mA	
$I_{Gp}$	Peak Gate Current	10	mA	$t_{PK} = 1\text{ms}, f = 100\text{Hz}$
$V_{GE}$	Gate-to-Emitter Voltage	Clamped	V	
$P_D @ T_C = 25^\circ\text{C}$	Maximum Power Dissipation	125	W	
$P_D @ T = 110^\circ\text{C}$	Maximum Power Dissipation	54	W	
$T_J$	Operating Junction and	- 40 to 175	$^\circ\text{C}$	
$T_{STG}$	Storage Temperature Range	- 40 to 175	$^\circ\text{C}$	
$V_{ESD}$	Electrostatic Voltage	6	KV	$C = 100\text{pF}, R = 1.5K \text{ ohm}$
$I_L$	Self-clamped Inductive Switching Current	11.5	A	$L = 4.7\text{mH}, T = 25^\circ\text{C}$

**Thermal Resistance**

	Parameter	Min	Typ	Max	Unit
$R_{\theta JC}$	Thermal Resistance, Junction-to-Case			1.2	$^\circ\text{C/W}$
$R_{\theta JA}$	Thermal Resistance, Junction-to-Ambient			40	
	(PCB Mounted, Steady State)				
$Z_{\theta JC}$	Transient Thermal Impedance, Junction-to-Case (Fig.11)				

**Off-State Electrical Characteristics @  $T_J = 25^\circ\text{C}$  (unless otherwise specified)**

	Parameter	Min	Typ	Max	Unit	Conditions	Fig
$BV_{CES}$	Collector-to-Emitter Breakdown Voltage	370	400	430	V	$R_G = 1\text{K ohm}, I_C = 7\text{A}, V_{GE} = 0\text{V}$	
$BV_{GES}$	Gate-to-Emitter Breakdown Voltage	10	12		V	$I_G = 2\text{m A}$	
$I_{CES}$	Collector-to-Emitter Leakage Current			15	$\mu\text{A}$	$R_G = 1\text{K ohm}, V_{CE} = 250\text{V}$	
				100	$\mu\text{A}$	$R_G = 1\text{K ohm}, V_{CE} = 250\text{V}, T_J = 150^\circ\text{C}$	
$BV_{CER}$	Emitter-to-Collector Breakdown Voltage	24	28		V	$I_C = -10\text{m A}$	
$R_1$	Gate Series Resistance		75		ohm		
$R_2$	Gate-to-Emitter Resistance	10	20	30	K ohm		

**On-State Electrical Characteristics @  $T_J = 25^\circ\text{C}$  (unless otherwise specified)**

	Parameter	Min	Typ	Max	Unit	Conditions	Fig	
$V_{CE(on)}$	Collector-to-Emitter Saturation Voltage		1.2	1.40	V	$I_C = 7\text{A}, V_{GE} = 4.5\text{V}$	1	
			1.35	1.55		$I_C = 10\text{A}, V_{GE} = 4.5\text{V}$		
			1.35	1.55		$I_C = 10\text{A}, V_{GE} = 4.5\text{V}, T_C = -40^\circ\text{C}$		2
			1.5	1.7		$I_C = 14\text{A}, V_{GE} = 5.0\text{V}, T_C = -40^\circ\text{C}$		4
			1.55	1.75		$I_C = 14\text{A}, V_{GE} = 5.0\text{V}$		
			1.6	1.8		$I_C = 14\text{A}, V_{GE} = 5.0\text{V}, T_C = 150^\circ\text{C}$		
$V_{GE(th)}$	Gate Threshold Voltage	1.3	1.8	2.2	V	$V_{CE} = V_{GE}, I_C = 1\text{ m A}, T_C = 25^\circ\text{C}$	3, 5	
		0.75		1.8		$V_{CE} = V_{GE}, I_C = 1\text{ m A}, T_C = 150^\circ\text{C}$		8
$g_{fs}$	Transconductance	10	15	19	S	$V_{CE} = 25\text{V}, I_C = 10\text{A}, T_C = 25^\circ\text{C}$		
$I_C$	Collector Current	20			A	$V_{CE} = 10\text{V}, V_{GE} = 4.5\text{V}$		

**Switching Characteristics @  $T_J = 25^\circ\text{C}$  (unless otherwise specified)**

	Parameter	Min	Typ	Max	Unit	Conditions	Fig	
$Q_g$	Total Gate charge		27			$I_C = 10\text{A}, V_{CE}=12\text{V}, V_{GE}=5\text{V}$	7	
$Q_{ge}$	Gate - Emitter Charge		2.5		nC	$I_C = 10\text{A}, V_{CE}=12\text{V}, V_{GE}=5\text{V}$	15	
$Q_{gc}$	Gate - Collector Charge		10			$I_C = 10\text{A}, V_{CE}=12\text{V}, V_{GE}=5\text{V}$		
$t_{d(on)}$	Turn - on delay time	0.6	0.9	1.35		$V_{GE}=5\text{V}, R_G=1\text{K ohm}, L=1\text{mH}, V_{CE}=14\text{V}$	12	
$t_r$	Rise time	1.6	2.8	4	$\mu\text{s}$	$V_{GE}=5\text{V}, R_G=1\text{K ohm}, L=1\text{mH}, V_{CE}=14\text{V}$	14	
$t_{d(off)}$	Turn - off delay time	3.7	6	8.3		$V_{GE}=5\text{V}, R_G=1\text{K ohm}, L=1\text{mH}, V_{CE}=300\text{V}$		
$C_{ies}$	Input Capacitance		550	825		$V_{GE}=0\text{V}, V_{CE}=25\text{V}, f=1\text{M H z}$	6	
$C_{oss}$	Output Capacitance		100	150	pF	$V_{GE}=0\text{V}, V_{CE}=25\text{V}, f=1\text{M H z}$		
$C_{res}$	Reverse Transfer Capacitance		12	18		$V_{GE}=0\text{V}, V_{CE}=25\text{V}, f=1\text{M H z}$		
$I_L$	Self-Clamped Inductive Switching Current	25			A	$L=0.7\text{m H}, T_C=25^\circ\text{C}$	9	
		15.5				$L=2.2\text{m H}, T_C=25^\circ\text{C}$		
		11.5				$L=4.7\text{m H}, T_C=25^\circ\text{C}$		10
		16.5				$L=1.5\text{m H}, T_C=150^\circ\text{C}$		13
		7.5				$L=4.7\text{m H}, T_C=150^\circ\text{C}$		14
		6				$L=8.7\text{m H}, T_C=150^\circ\text{C}$		
$t_{sc}$	Short Circuit Withstand Time		120		$\mu\text{s}$	$T_J = 150^\circ\text{C},$ $V_{CC} = 16\text{V}, L = 10\mu\text{H}$ $R_G = 1\text{K ohm}, V_{GE} = 5\text{V}$	14	

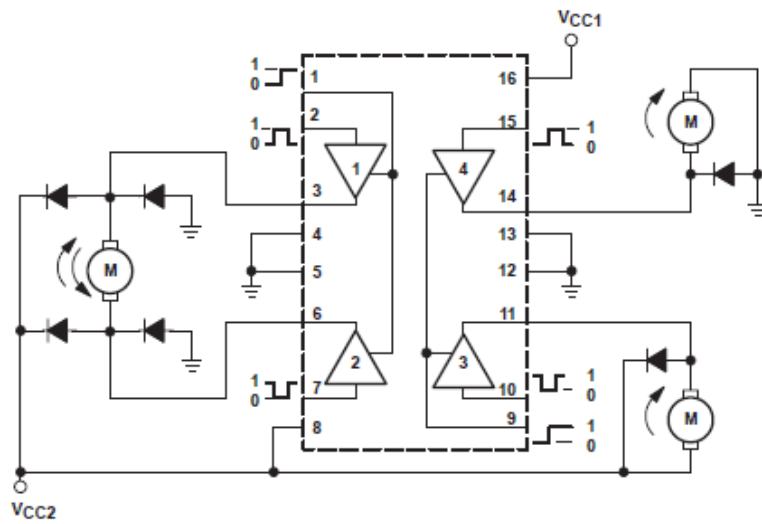
DRIVER FULL BRIDGE L297 STEP MOTOR CONTROLLER

L298N 4A 48 V

L293 1A 4,5 a 36V

L298 Step motor controller para operar junto com o L298

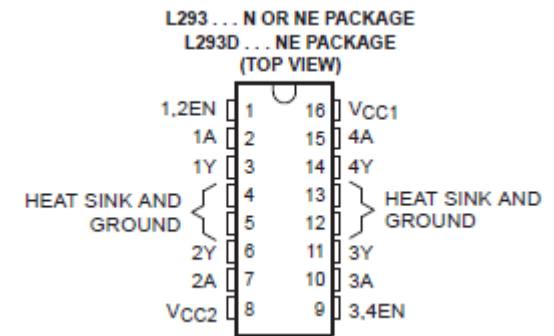
L293 4,5 A 36 V 1A 2A DE PICO 4 AMP



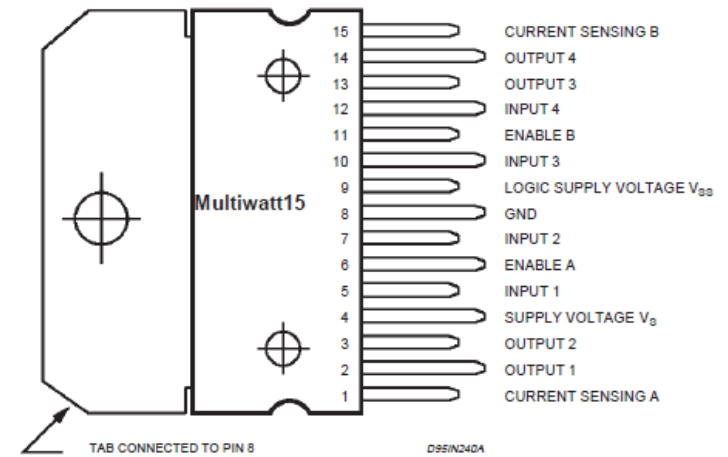
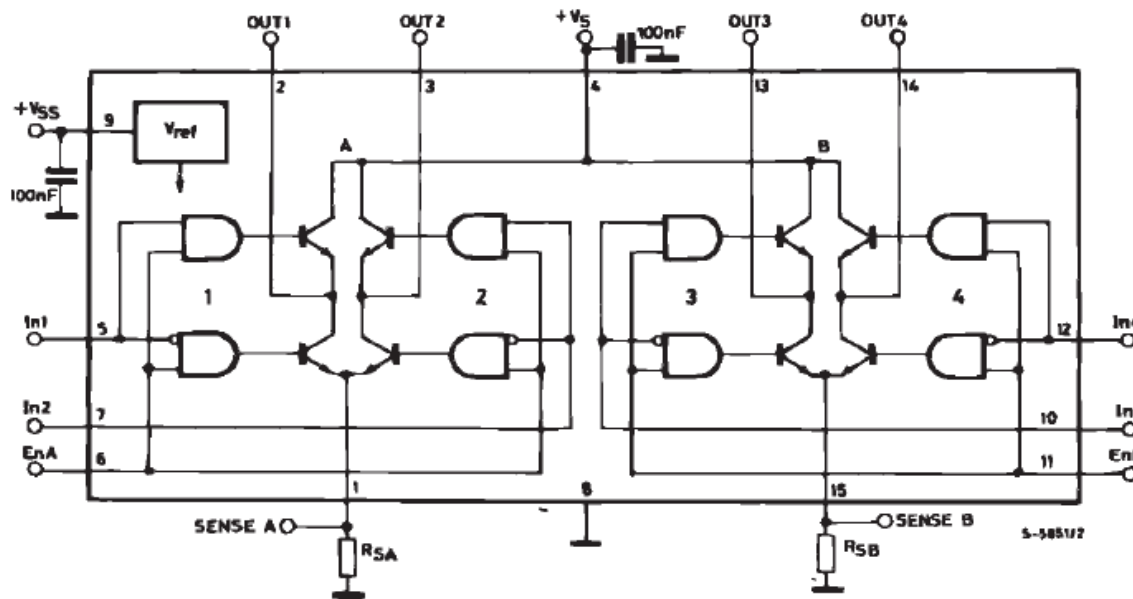
FUNCTION TABLE  
(each driver)

INPUTS†		OUTPUT
A	EN	Y
H	H	H
L	H	L
X	L	Z

H = high level, L = low level, X = irrelevant, Z = high impedance (off)  
 † In the thermal shutdown mode, the output is in the high-impedance state, regardless of the input levels.

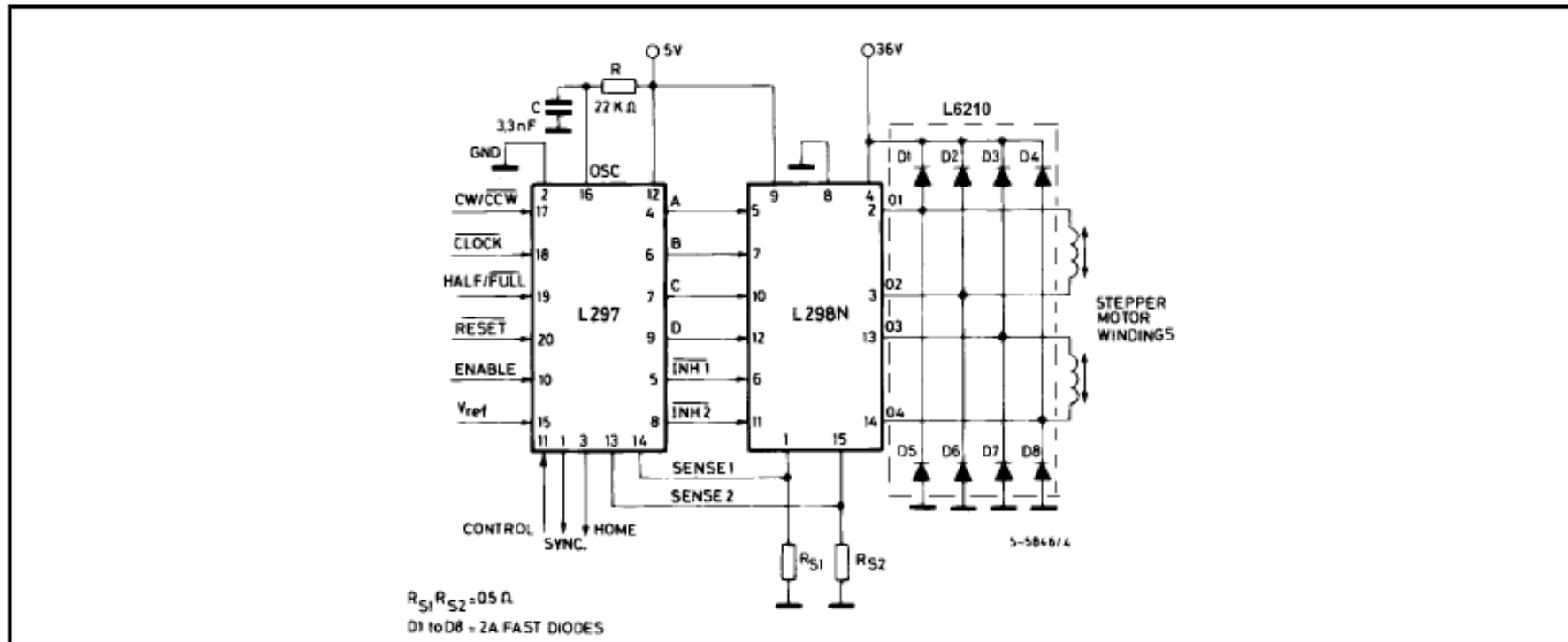


L298 40V 4A EM PONTE

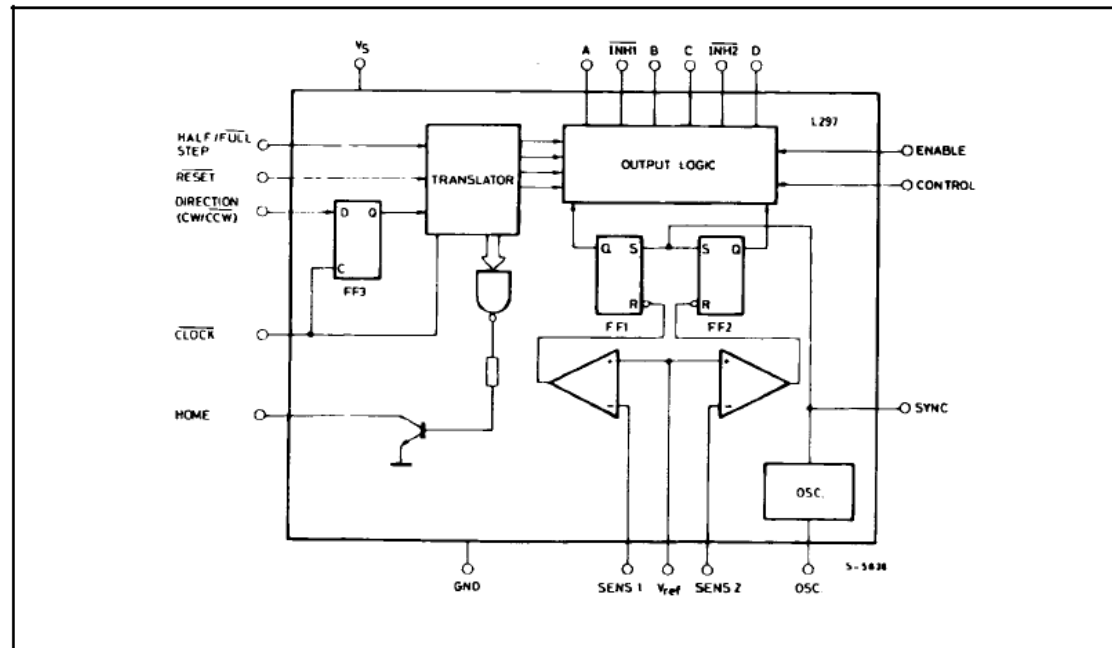
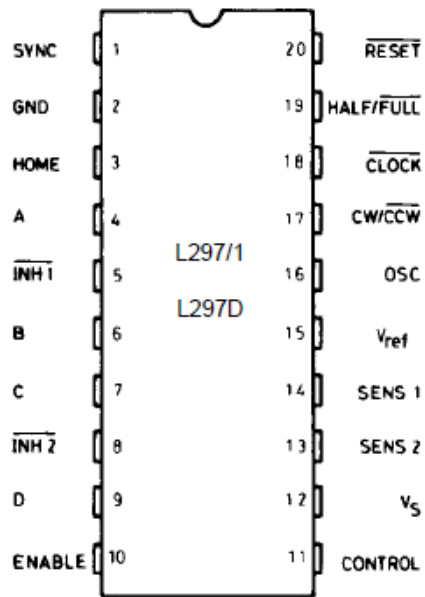


297 STEP MOTOR CONTROLLER

TWO PHASE BIPOLAR STEPPER MOTOR CONTROL CIRCUIT



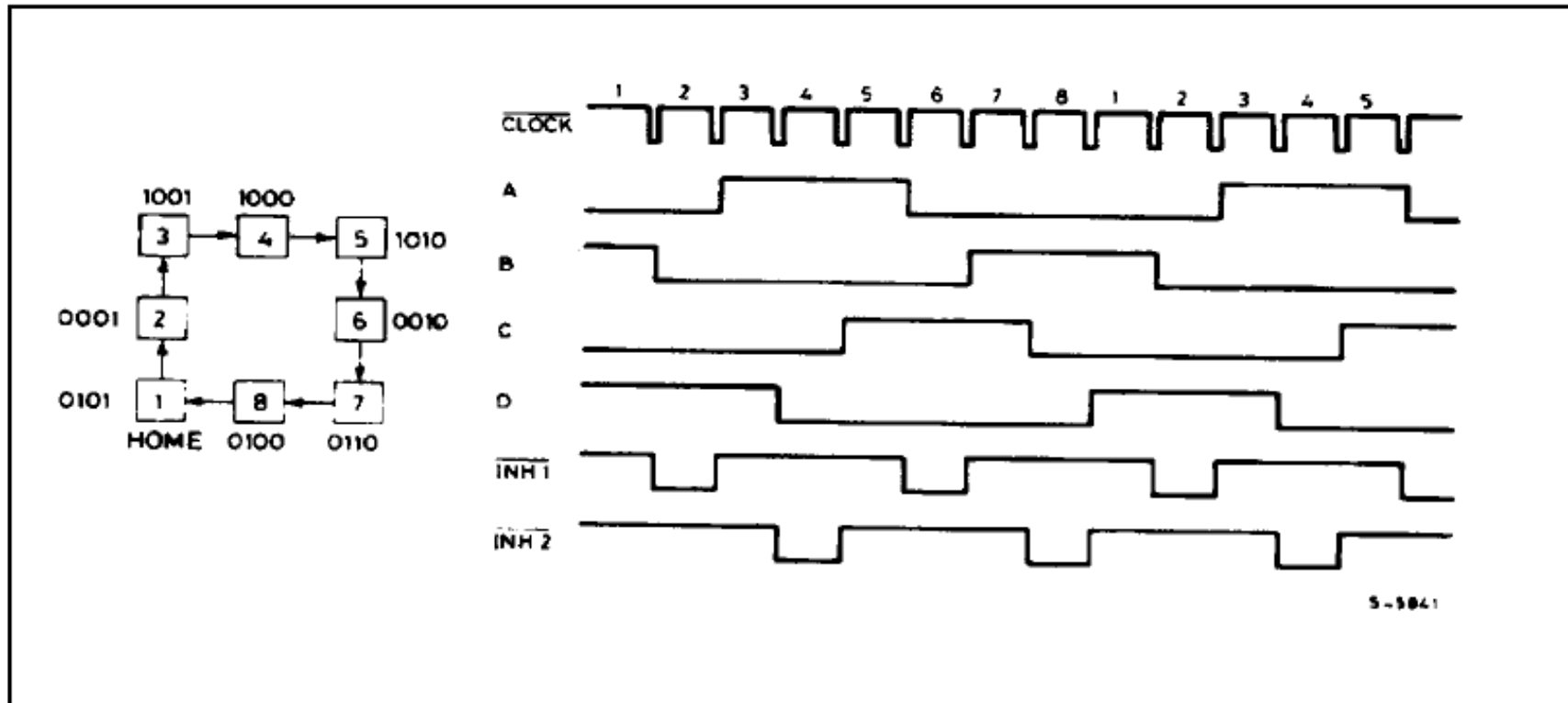
BLOCK DIAGRAM (L297/1 - L297D)





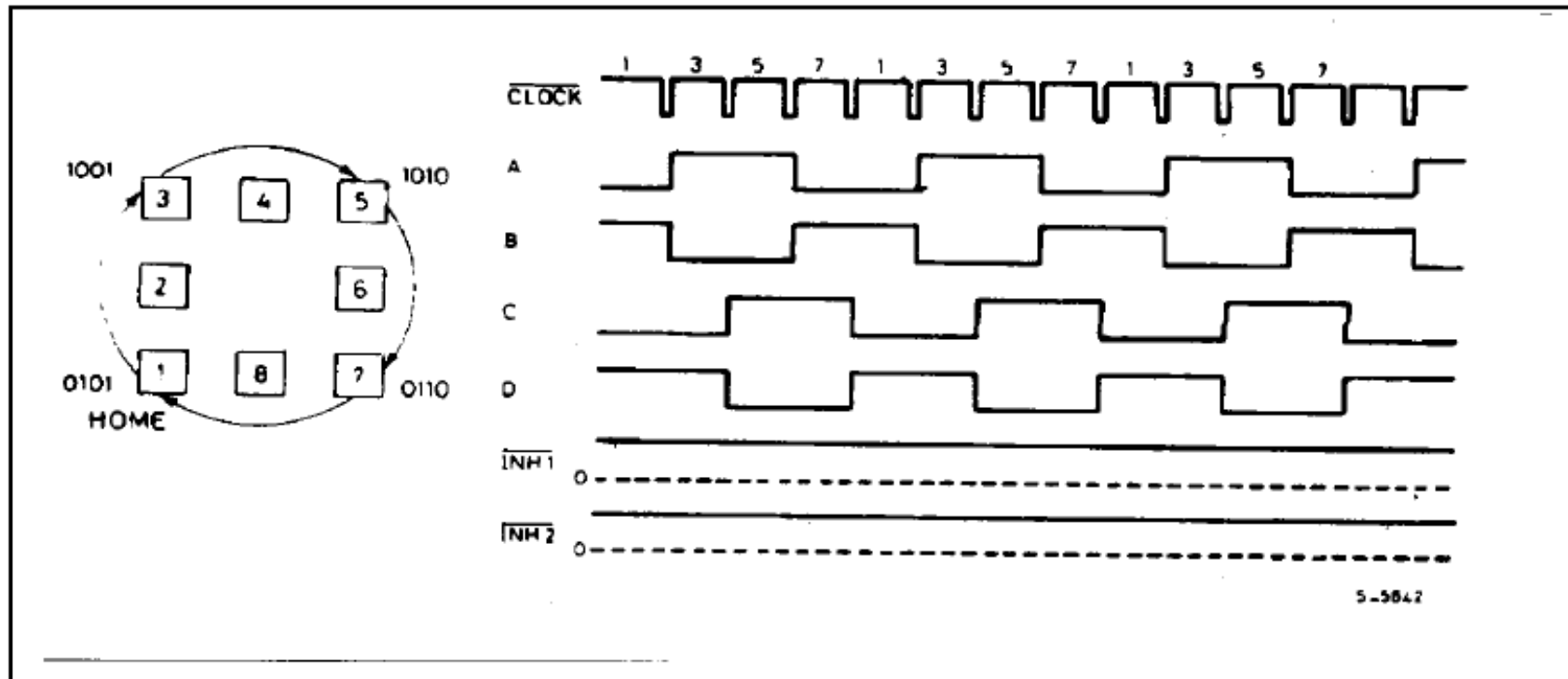
### HALF STEP MODE

Half step mode is selected by a high level on the HALF/FULL input.

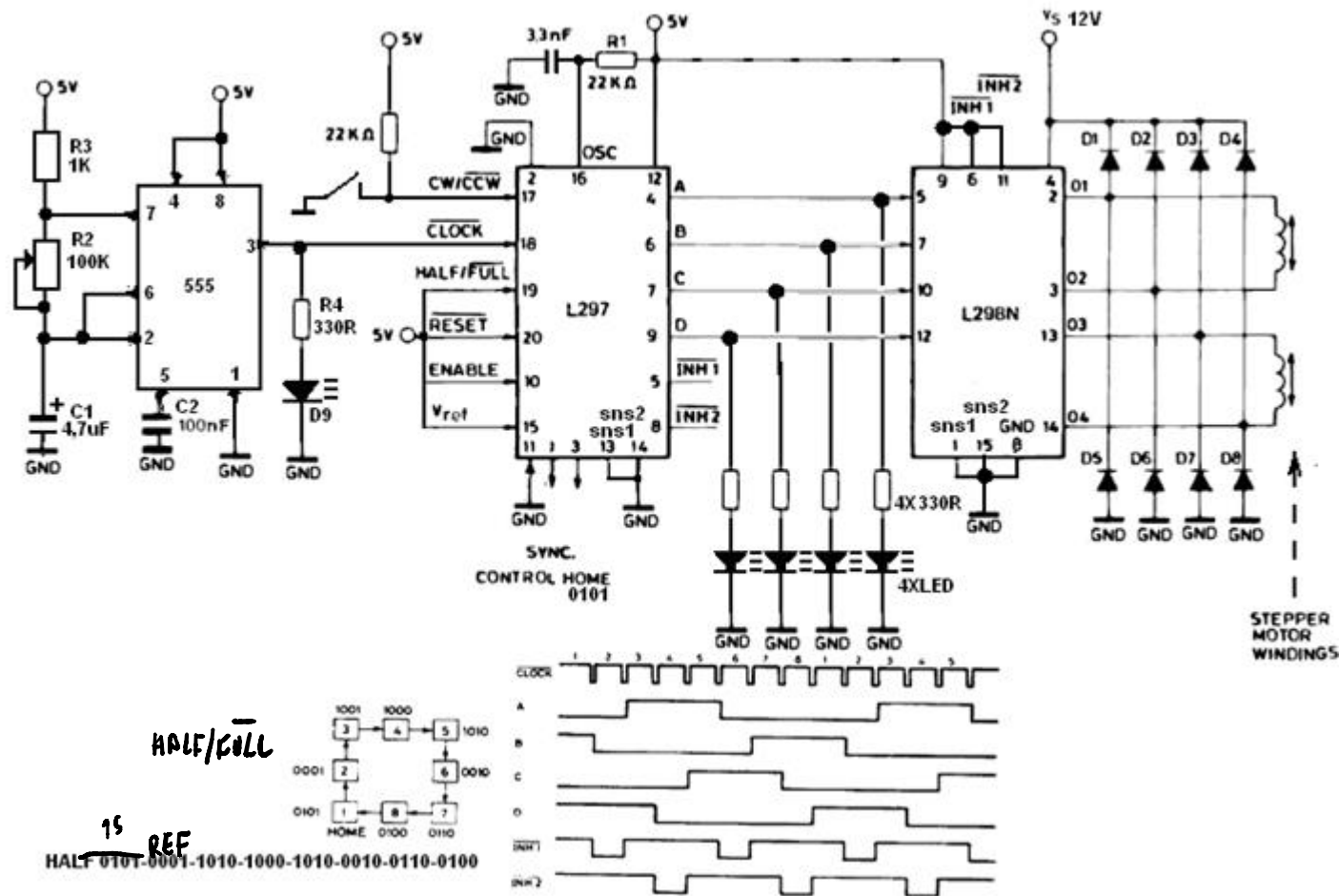


### NORMAL DRIVE MODE

Normal drive mode (also called "two-phase-on" drive) is selected by a low level on the  $\overline{\text{HALF/FULL}}$  input when the translator is at an odd numbered state (1, 3, 5 or 7). In this mode the  $\overline{\text{INH1}}$  and  $\overline{\text{INH2}}$  outputs remain high throughout.



Exemplo de motor de passo:



## JFET

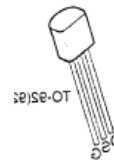
BF245      VP=2,4V      IDSS=24 mA

2N5457/58/59 tá na lista

MPF102      VHFAMPLIFIER VGSOFF=-8V      |Yfs|=2000 a 7500 umhos ciss 7pf

LISTA CANAL N

Device	BV <sub>GSS</sub> (V) Min	V <sub>p</sub> @ V <sub>DS</sub> f <sub>o</sub>				Re(Y <sub>fs</sub> ) (mmho) @ f		Re(Y <sub>os</sub> ) (μmho) @ f		C <sub>iss</sub> (pF) Max	C <sub>rss</sub> (pF) Max	NF (dB) @ R <sub>g</sub> = 1k		Package
		Min	Max	(V)	(nA)	Min	(MHz)	Max	(MHz)			Max	(MHz)	
2N3819	25		8	15	2	1.6	100			8	4			TO-92(94)
2N4416	30	2.5	6	15	1	4	400	100	400	4	0.9	4	400	TO-72
PN4416	30	2.5	6	15	1	4	400	100	400	4	0.9	4	400	TO-92(92)
MMBF4416	30	2.5	6	15	1	4	400	100	400	4	0.9	4	400	TO-236*
2N5245	30	1	6	15	10	4	400	100	400	4.5	1.0	4	400	TO-92(97)
2N5246	30	0.5	4	15	10	2.5	400	100	400	4.5	1.0	4	400	TO-92(97)
2N5247	30	1.5	8	15	10	4	400	150	400	4.5	1.0	4	400	TO-92(97)
2N5397	25	1	6	10	1	5.5	450	400	450	5	1.2	3.5	450	TO-72
2N5484	25	0.3	3	15	10	2.5	100	75	100	5	1	3	100	TO-92(92)
MMBF5484	25	0.3	3	15	10	2.5	100	75	100	5	1	3	100	TO-236*
2N5485	25	0.5	4	15	10	3	400	100	400	5	1	4	400	TO-92(92)
MMBF5485	25	0.5	4	15	10	3	400	100	400	5	1	4	400	TO-236*
2N5486	25	2	6	15	10	3.5	400	100	400	5	1	4	400	TO-92(92)
MMBF5486	25	2	6	15	10	3.5	400	100	400	5	1	4	400	TO-236*
2N5949	30	3	7	15	100	3	100	150	100	6	2	5	100	TO-92(97)
2N5950	30	2.5	6	15	100	3	100	125	100	6	2	5	100	TO-92(97)
2N5951	30	2	5	15	100	3	100	100	100	6	2	5	100	TO-92(97)
2N5952	30	1.3	4.5	15	100	1	100	75	100	6	2	5	100	TO-92(97)
2N5953	30	0.8	3	15	100	1	100	50	100	6	2	5	100	TO-92(97)
J300	25	1	6	10	1	4.5	0.001	200	0.001	5.5	1.7			TO-92(92)
MMBFJ304	30	2	6	15	1	4.2	400	180	100					TO-236*
J304	30	2	6	15	1	4.2	400	180	100					TO-92(92)
J305	30	0.5	3	15	1	3.0	400	180	100					TO-92(92)
MMBFJ305	30	0.5	3	15	1	3.0	400	180	100					TO-236*
MMBFJ309	25	1	4	10	1	10	0.001	150	0.001	7.5	2.5			TO-236*
J309	25	1	4	10	1	10	0.001	150	0.001	7.5	2.5			TO-92(92)
MMBFJ310	25	2	6.5	10	1	8	0.001	150	0.001	7.5	2.5			TO-236*
J310	25	2	6.5	10	1	8	0.001	150	0.001	7.5	2.5			TO-92(92)
U303	25	1	4	10	1	10	0.001	150	100	5	2.5			TO-52
U310	25	2.5	6	10	1	10	0.001	150	100	5	2.5			TO-52



BF245



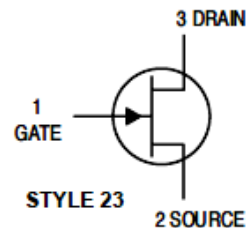
**ON Semiconductor™**

**JFET VHF/UHF Amplifiers**  
N-Channel — Depletion

**BF245A**  
**BF245B**

**MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Drain-Source Voltage	$V_{DS}$	$\pm 30$	Vdc
Drain-Gate Voltage	$V_{DG}$	30	Vdc
Gate-Source Voltage	$V_{GS}$	30	Vdc
Drain Current	$I_D$	100	mAdc
Forward Gate Current	$I_{G(f)}$	10	mAdc
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	350 2.8	mW mW/ $^\circ\text{C}$
Storage Channel Temperature Range	$T_{stg}$	-65 to +150	$^\circ\text{C}$



**ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Gate–Source Breakdown Voltage ( $I_G = 1.0 \mu\text{A}_{dc}$ , $V_{DS} = 0$ )	$V_{(BR)GSS}$	30	—	—	Vdc
Gate–Source ( $V_{DS} = 15 \text{ Vdc}$ , $I_D = 200 \mu\text{A}_{dc}$ )	$V_{GS}$	0.4	—	7.5	Vdc
BF245(1)		0.4	—	2.2	
BF245A, BF244A(2)		1.6	—	3.8	
BF245B, BF244B		3.2	—	7.5	
BF245C					
Gate–Source Cutoff Voltage ( $V_{DS} = 15 \text{ Vdc}$ , $I_D = 10 \text{ nA}_{dc}$ )	$V_{GS(off)}$	–0.5	—	–8.0	Vdc
Gate Reverse Current ( $V_{GS} = 20 \text{ Vdc}$ , $V_{DS} = 0$ )	$I_{GSS}$	—	—	5.0	nA <sub>dc</sub>
<b>ON CHARACTERISTICS</b>					
Zero–Gate–Voltage Drain Current ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ )	$I_{DSS}$	2.0	—	25	mA <sub>dc</sub>
BF245(1)		2.0	—	6.5	
BF245A, BF244A(2)		6.0	—	15	
BF245B, BF244B		12	—	25	
BF245C					

1. On orders against the BF245, any or all subgroups might be shipped.
2. On orders against the BF244A, any or all subgroups might be shipped.

**BF245A BF245B****ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted) (Continued)

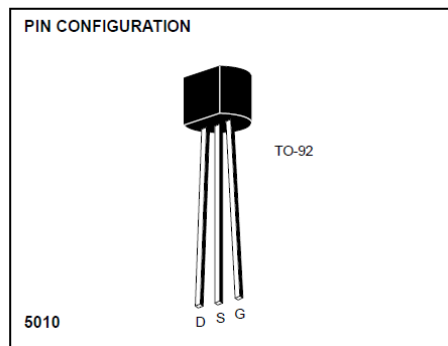
Characteristic	Symbol	Min	Typ	Max	Unit
<b>SMALL-SIGNAL CHARACTERISTICS</b>					
Forward Transfer Admittance ( $V_{DS} = 15\text{ Vdc}, V_{GS} = 0, f = 1.0\text{ kHz}$ )	$ Y_{fs} $	3.0	—	6.5	mmhos
Output Admittance ( $V_{DS} = 15\text{ Vdc}, V_{GS} = 0, f = 1.0\text{ kHz}$ )	$ Y_{os} $	—	40	—	$\mu\text{mhos}$
Forward Transfer Admittance ( $V_{DS} = 15\text{ Vdc}, V_{GS} = 0, f = 200\text{ MHz}$ )	$ Y_{fs} $	—	5.6	—	mmhos
Reverse Transfer Admittance ( $V_{DS} = 15\text{ Vdc}, V_{GS} = 0, f = 200\text{ MHz}$ )	$ Y_{rs} $	—	1.0	—	mmhos
Input Capacitance ( $V_{DS} = 20\text{ Vdc}, -V_{GS} = 1.0\text{ Vdc}$ )	$C_{iss}$	—	3.0	—	pF
Reverse Transfer Capacitance ( $V_{DS} = 20\text{ Vdc}, -V_{GS} = 1.0\text{ Vdc}, f = 1.0\text{ MHz}$ )	$C_{rss}$	—	0.7	—	pF
Output Capacitance ( $V_{DS} = 20\text{ Vdc}, -V_{GS} = 1.0\text{ Vdc}, f = 1.0\text{ MHz}$ )	$C_{oss}$	—	0.9	—	pF
Cut-off Frequency <sup>(3)</sup> ( $V_{DS} = 15\text{ Vdc}, V_{GS} = 0$ )	$F(Y_{fs})$	—	700	—	MHz

3. The frequency at which  $g_{fs}$  is 0.7 of its value at 1 kHz.



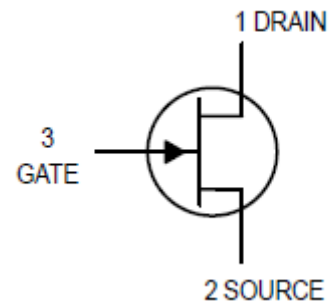
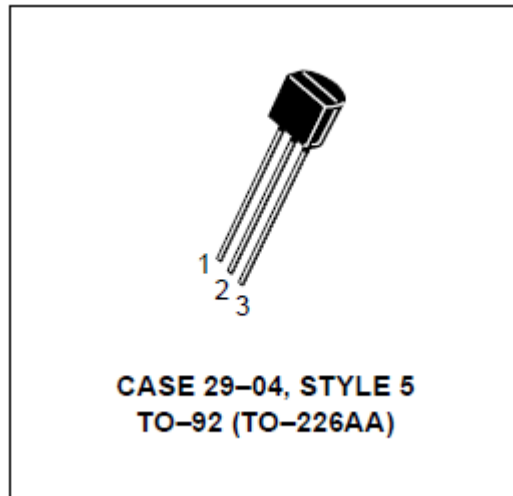
2N5457/2N5459 JFET MUITO USADO PELOS AMERICANOS

*2N5457 – 2N5459*



Uso geral muito usado pelos americanos

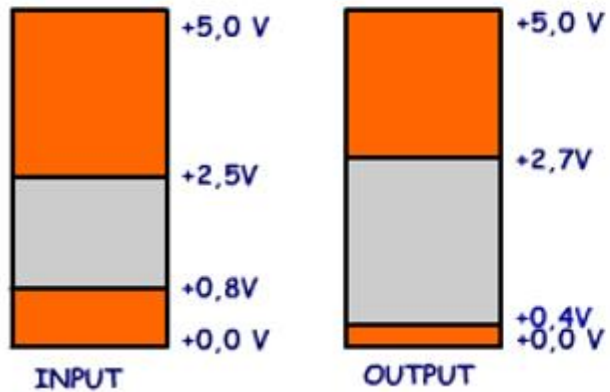
MPF-102 VHF AMPLIFIER



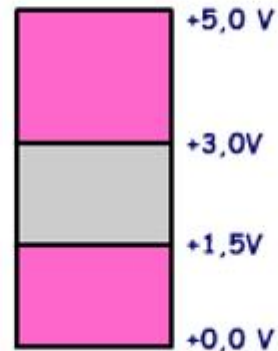
DIGITAL

NÍVEIS DE TENSÃO TTL CMOS

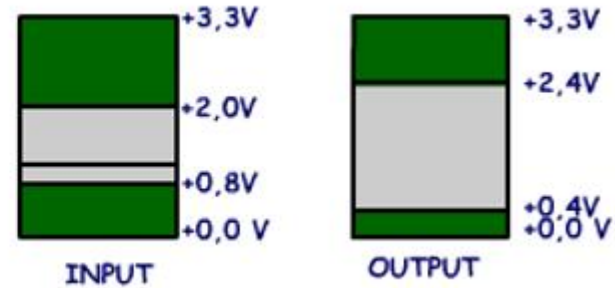
5-V TTL  
Standard TTL: ABT,  
AHCT, HCT, ACT,  
bipolar, LV1T, LV4T



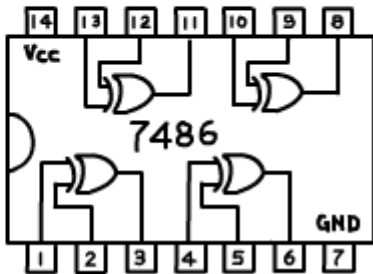
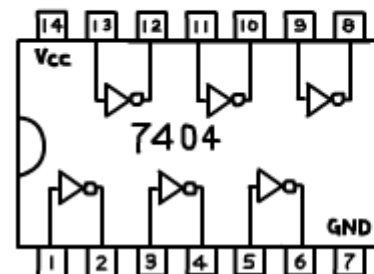
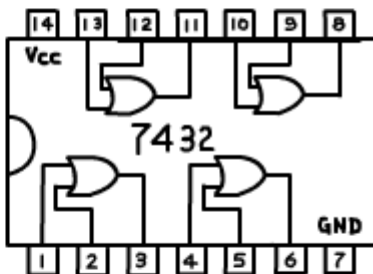
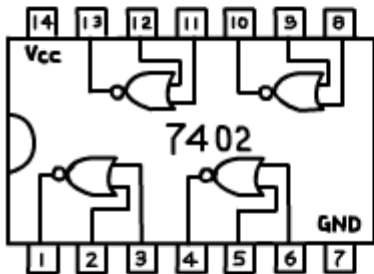
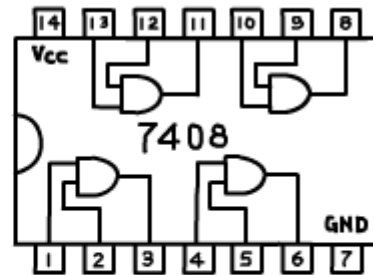
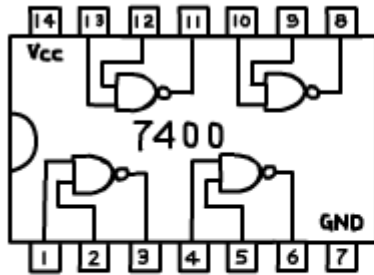
5-V CMOS  
Rail-to-Rail 5 V  
HC, AHC, AC, LV-A,  
LV1T, LV4T



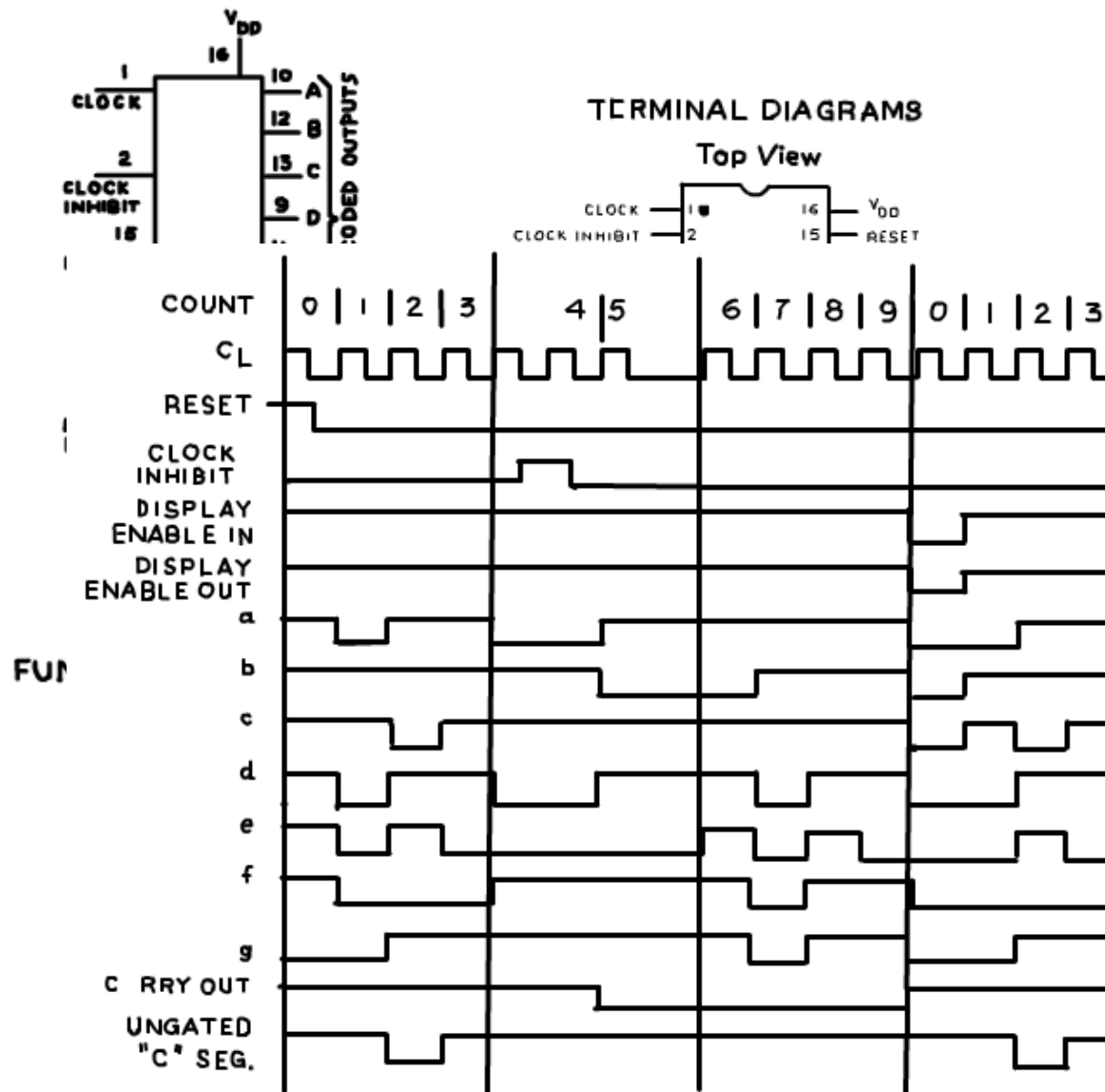
3.3-V LVTTTL  
LVT, LV1T, LV4T,  
LVC, ALVC, AUP,  
LV-A, ALVT



LISTA DAS PORTAS LÓGICAS TTL MAIS USADAS

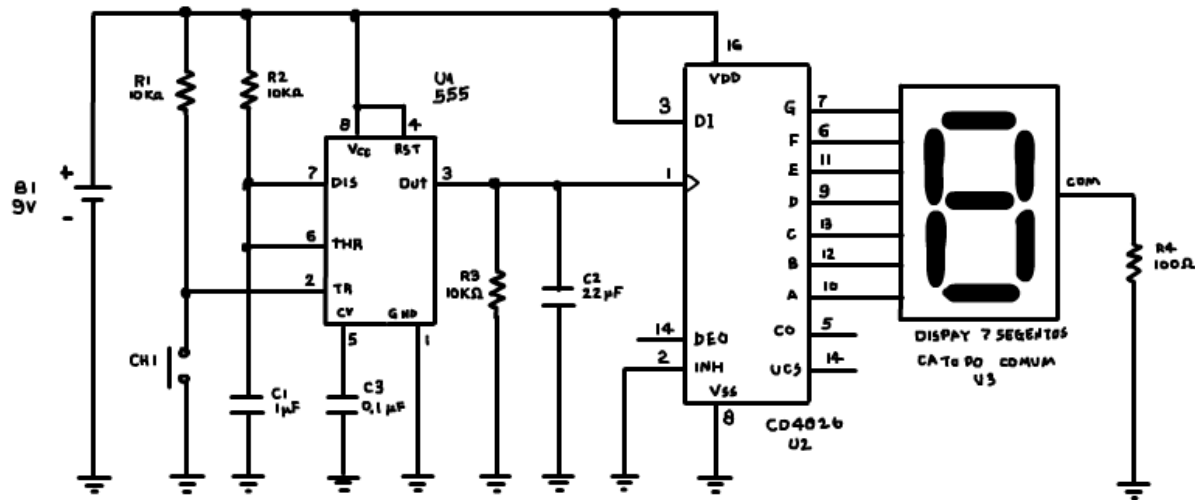


CONTADOR 7 SEGMENTOS 4026



*Fig. 3 – CD4026B timing diagram.*

Exemplo de aplicação.

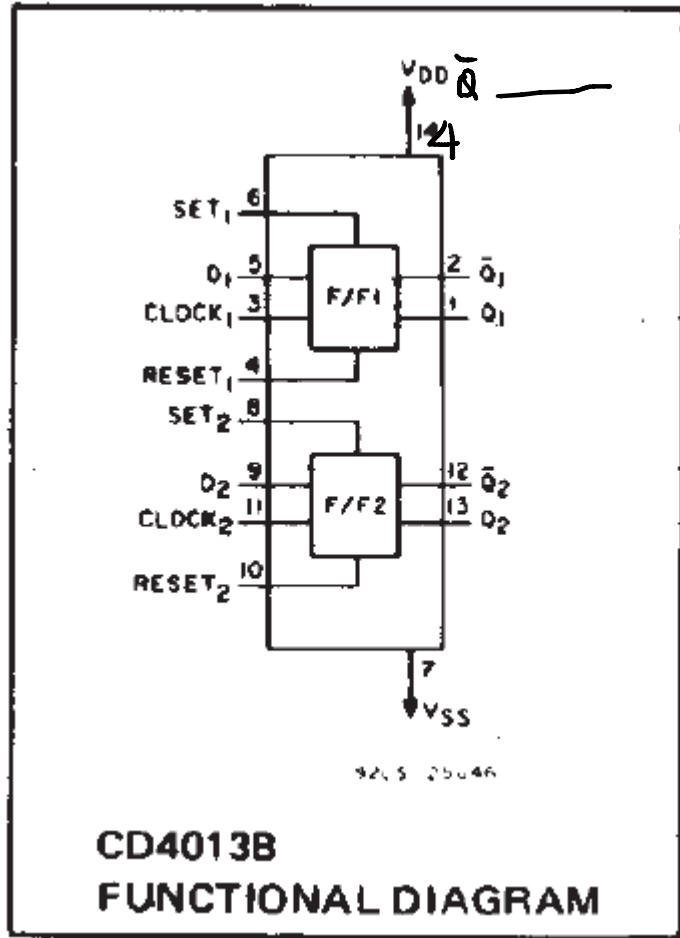


contador digital 7segmentos com 4026

YOUTUBE: <https://youtu.be/u0xZZCLX8hg>

CD4013 DUAL D F-F CMOS



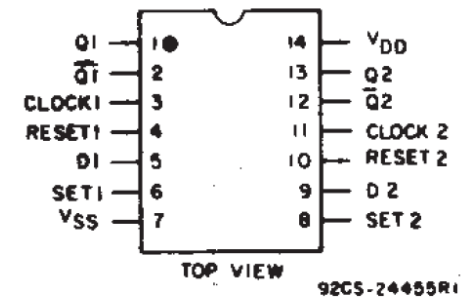


TRUTH TABLE

CL <sup>+</sup>	D	R	S	Q	Q̄
	0	0	0	0	1
	1	0	0	1	0
	X	0	0	Q	Q̄
X	X	1	0	0	1
X	X	0	1	1	0
X	X	1	1	1	1

NO CHANGE

LOGIC 0 = LOW  
LOGIC 1 = HIGH  
▲ = LEVEL CHANGE  
X = DON'T CARE  
N(N) = FF1/FF2 TERMINAL ASSIGNMENTS



**TERMINAL ASSIGNMENT**

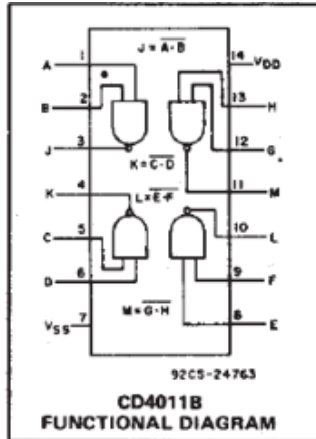
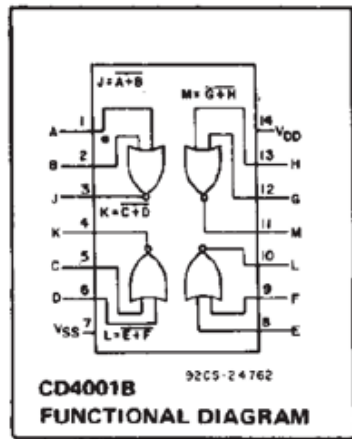
PORTAS LÓGICAS DA FAMÍLIA CMOS MAIS COMUNS

CD4001

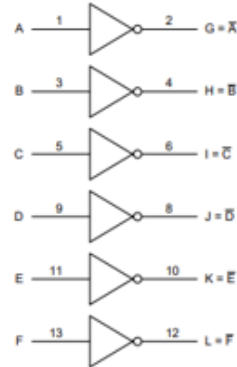
CD4001

CD4069

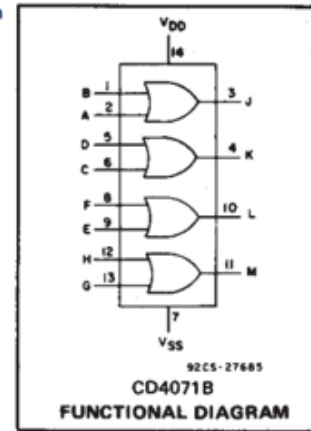
CD4071



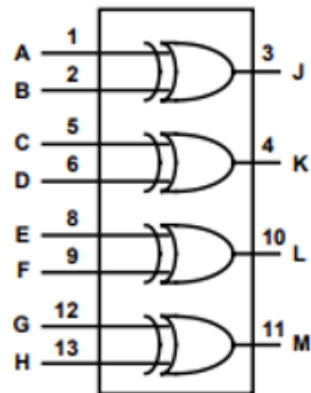
CD4069UB Functional Diagram



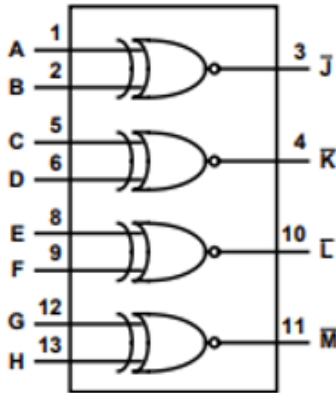
VDD = Pin 14  
VSS = Pin 7



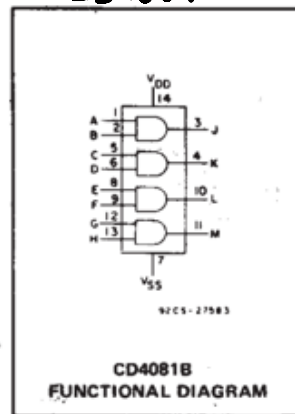
CD4070B



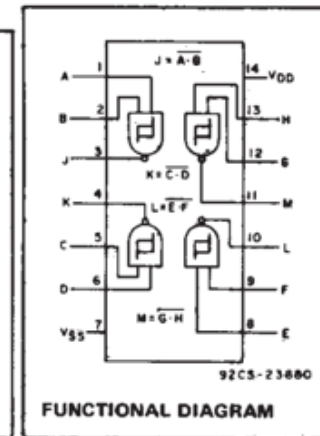
CD4077B



CD4081



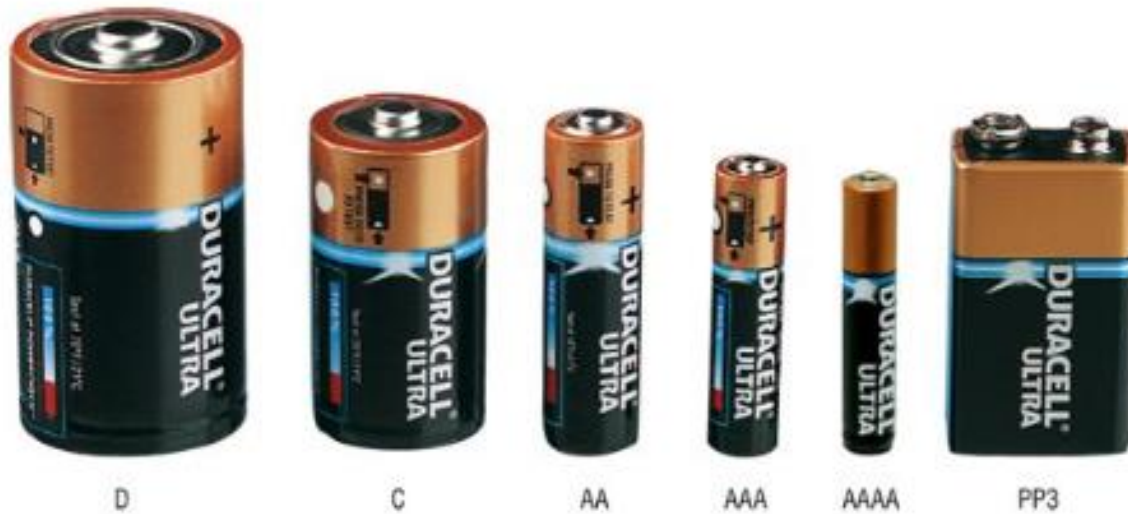
CD4093



BATERIAS PILHAS

## TIPOS DE PILHAS

Fun fact: AAAA ("quadruple-A") cells exist, too, and they're tiny. As my daughter says, "the smaller the battery, the longer it screams."



2N2646 2N2647 UNIJUNÇÃO

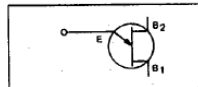
**Boca Semiconductor Corp. (BSC)**  
**PN Unijunction Transistors**  
**Silicon PN Unijunction Transistors**

... designed for use in pulse and timing circuits, sensing circuits and thyristor trigger circuits. These devices feature:

- Low Peak Point Current — 2  $\mu$ A (Max)
- Low Emitter Reverse Current — 200 nA (Max)
- Passivated Surface for Reliability and Uniformity

**2N2646**  
**2N2647**

PN UJTs



**\*ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted.)

Characteristic	Symbol	Min	Typ	Max	Unit
Intrinsic Standoff Ratio ( $V_{B2B1} = 10\text{ V}$ ), Note 1	$\eta$	0.56 0.68	—	0.75 0.82	—
Interbase Resistance ( $V_{B2B1} = 3\text{ V}$ , $I_E = 0$ )	$r_{BB}$	4.7	7	9.1	k ohms
Interbase Resistance Temperature Coefficient ( $V_{B2B1} = 3\text{ V}$ , $I_E = 0$ , $T_A = -55^\circ\text{C}$ to $+125^\circ\text{C}$ )	$\alpha r_{BB}$	0.1	—	0.9	%/°C
Emitter Saturation Voltage ( $V_{B2B1} = 10\text{ V}$ , $I_E = 50\text{ mA}$ ), Note 2	$V_{EB1}(\text{sat})$	—	3.5	—	Volts
Modulated Interbase Current ( $V_{B2B1} = 10\text{ V}$ , $I_E = 50\text{ mA}$ )	$I_{B2}(\text{mod})$	—	15	—	mA
Emitter Reverse Current ( $V_{B2E} = 30\text{ V}$ , $I_{B1} = 0$ )	$I_{EB2O}$	— —	0.005 0.005	12 0.2	$\mu$ A
Peak Point Emitter Current ( $V_{B2B1} = 25\text{ V}$ )	$I_p$	— —	1 1	5 2	$\mu$ A
Valley Point Current ( $V_{B2B1} = 20\text{ V}$ , $R_{B2} = 100\text{ ohms}$ ), Note 2	$I_v$	4 8	6 10	— 18	mA
Base-One Peak Pulse Voltage (Note 3, Figure 3)	$V_{OB1}$	3 6	5 7	— —	Volts

**\*MAXIMUM RATINGS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted.)

Rating	Symbol	Value	Unit
Power Dissipation, Note 1	$P_D$	300	mW
RMS Emitter Current	$I_E(\text{RMS})$	50	mA
Peak Pulse Emitter Current, Note 2	$i_E$	2	Amps
Emitter Reverse Voltage	$V_{B2E}$	30	Volts
Interbase Voltage	$V_{B2B1}$	35	Volts
Operating Junction Temperature Range	$T_J$	-65 to +125	°C
Storage Temperature Range	$T_{stg}$	-65 to +150	°C

RELÉ DE ESTADO SÓLIDO TRIFÁSICO (SSR)

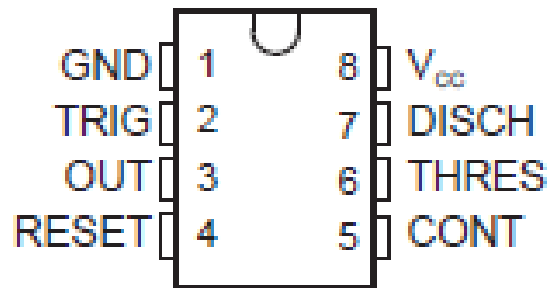


Relé Estado Sólido Trifásico Ssr 60a + Dissipador + Cooler

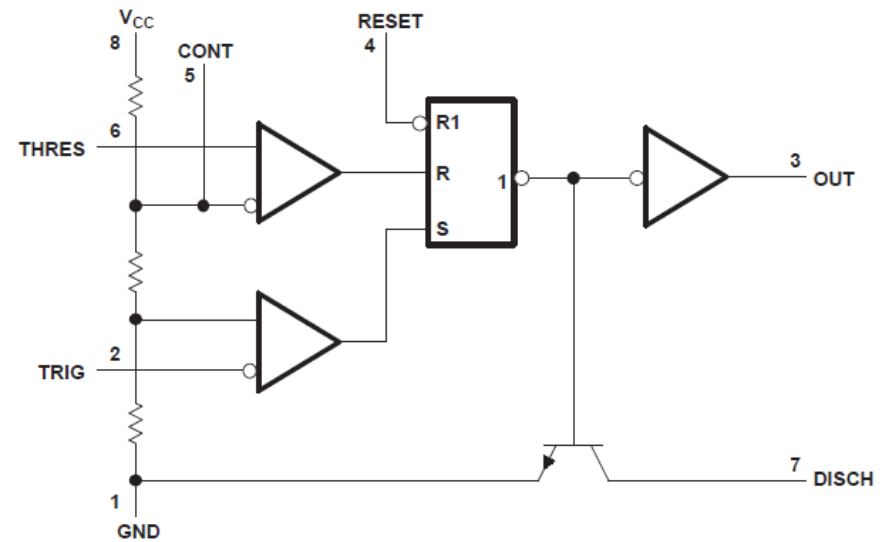
TEMPORIZADORES

TEMPORIZADOR 555

**NA555...D OR P PACKAGE**  
**NE555...D, P, PS, OR PW PACKAGE**  
**SA555...D OR P PACKAGE**  
**SE555...D, JG, OR P PACKAGE**  
**(TOP VIEW)**



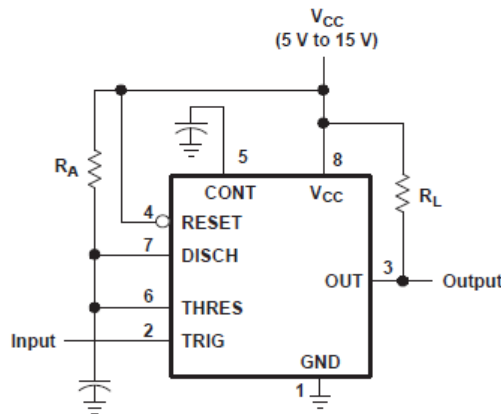
FUNCTIONAL BLOCK DIAGRAM



- A. Pin numbers shown are for the D, JG, P, PS, and PW packages.
- B. RESET can override TRIG, which can override THRES.

**Monostable Operation**

For monostable operation, any of these timers can be connected as shown in Figure 9. If the output is low, application of a negative-going pulse to the trigger (TRIG) sets the flip-flop ( $\bar{Q}$  goes low), drives the output high, and turns off Q1. Capacitor C then is charged through  $R_A$  until the voltage across the capacitor reaches the threshold voltage of the threshold (THRES) input. If TRIG has returned to a high level, the output of the threshold comparator resets the flip-flop ( $\bar{Q}$  goes high), drives the output low, and discharges C through Q1.

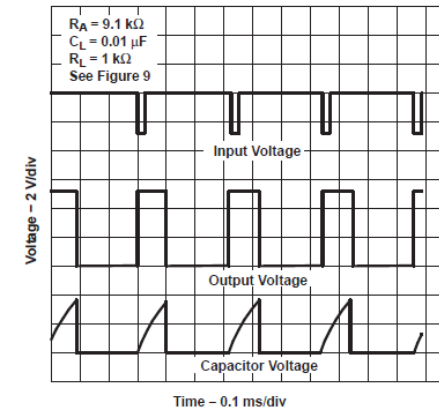


Pin numbers shown are for the D, JG, P, PS, and PW packages.

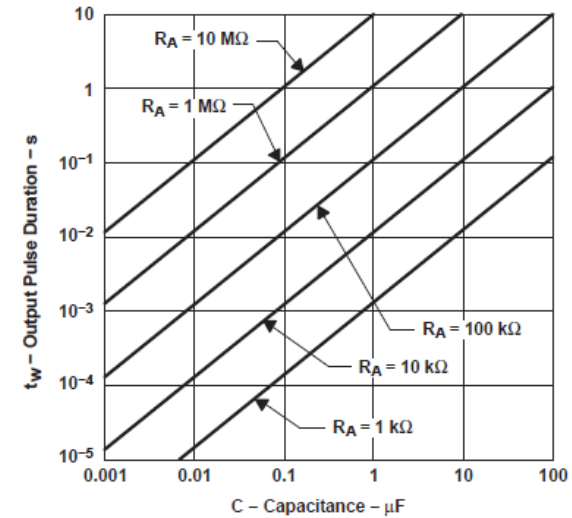
**Figure 9. Circuit for Monostable Operation**

Monostable operation is initiated when TRIG voltage falls below the trigger threshold. Once initiated, the sequence ends only if TRIG is high for at least 10  $\mu$ s before the end of the timing interval. When the trigger is grounded, the comparator storage time can be as long as 10  $\mu$ s, which limits the minimum monostable pulse width to 10  $\mu$ s. Because of the threshold level and saturation voltage of Q1, the output pulse duration is approximately  $t_w = 1.1R_A C$ . Figure 11 is a plot of the time constant for various values of  $R_A$  and C. The threshold levels and charge rates both are directly proportional to the supply voltage,  $V_{CC}$ . The timing interval is, therefore, independent of the supply voltage, so long as the supply voltage is constant during the time interval.

Applying a negative-going trigger pulse simultaneously to RESET and TRIG during the timing interval discharges C and reinitiates the cycle, commencing on the positive edge of the reset pulse. The output is held low as long as the reset pulse is low. To prevent false triggering, when RESET is not used, it should be connected to  $V_{CC}$ .



**Figure 10. Typical Monostable Waveforms**



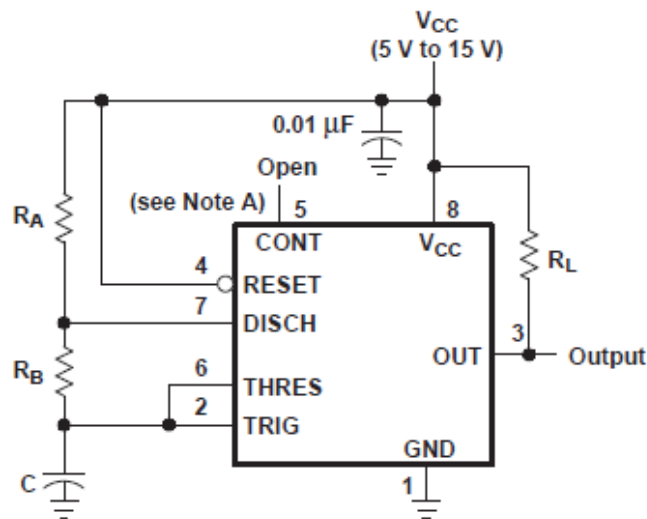
**Figure 11. Output Pulse Duration vs Capacitance**



**Astable Operation**

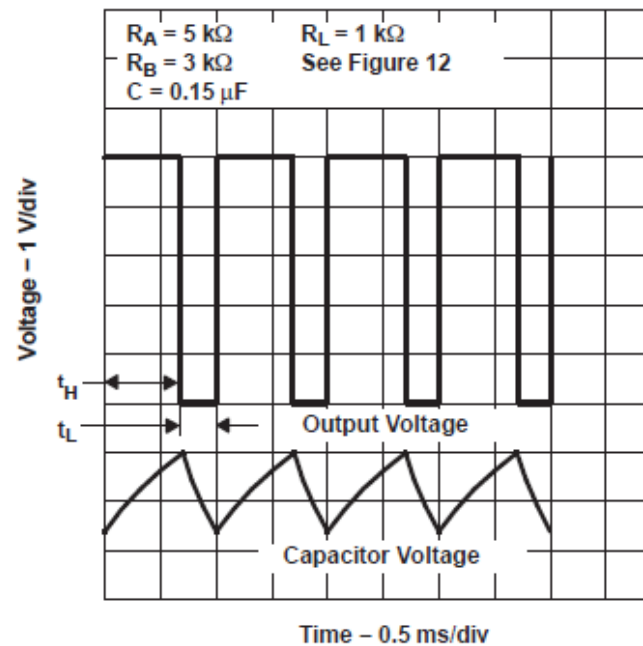
As shown in Figure 12, adding a second resistor,  $R_B$ , to the circuit of Figure 9 and connecting the trigger input to the threshold input causes the timer to self-trigger and run as a multivibrator. The capacitor  $C$  charges through  $R_A$  and  $R_B$  and then discharges through  $R_B$  only. Therefore, the duty cycle is controlled by the values of  $R_A$  and  $R_B$ .

This astable connection results in capacitor  $C$  charging and discharging between the threshold-voltage level ( $\approx 0.67 \times V_{CC}$ ) and the trigger-voltage level ( $\approx 0.33 \times V_{CC}$ ). As in the monostable circuit, charge and discharge times (and, therefore, the frequency and duty cycle) are independent of the supply voltage.



Pin numbers shown are for the D, JG, P, PS, and PW packages.  
 NOTE A: Decoupling CONT voltage to ground with a capacitor can improve operation. This should be evaluated for individual applications.

**Figure 12. Circuit for Astable Operation**



**Figure 13. Typical Astable Waveforms**

Figure 12 shows typical waveforms generated during astable operation. The output high-level duration  $t_H$  and low-level duration  $t_L$  can be calculated as follows:

$$t_H = 0.693 (R_A + R_B) C$$

$$t_L = 0.693 (R_B) C$$

Other useful relationships are shown below.

$$\text{period} = t_H + t_L = 0.693 (R_A + 2R_B) C$$

$$\text{frequency} \approx \frac{1.44}{(R_A + 2R_B) C}$$

$$\text{Output driver duty cycle} = \frac{t_L}{t_H + t_L} = \frac{R_B}{R_A + 2R_B}$$

Output waveform duty cycle

$$= \frac{t_H}{t_H + t_L} = 1 - \frac{R_B}{R_A + 2R_B}$$

$$\text{Low-to-high ratio} = \frac{t_L}{t_H} = \frac{R_B}{R_A + R_B}$$

Figure .

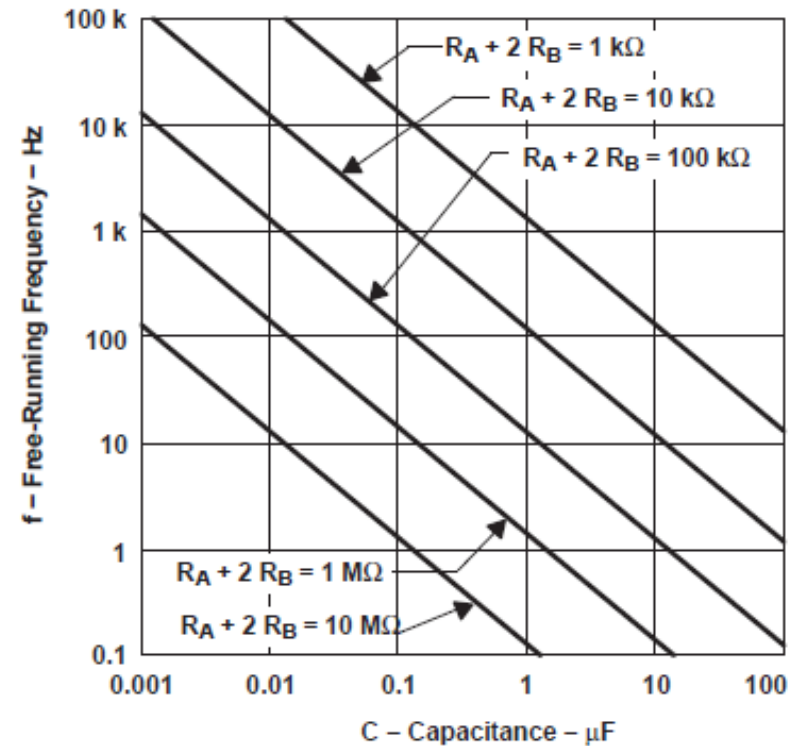
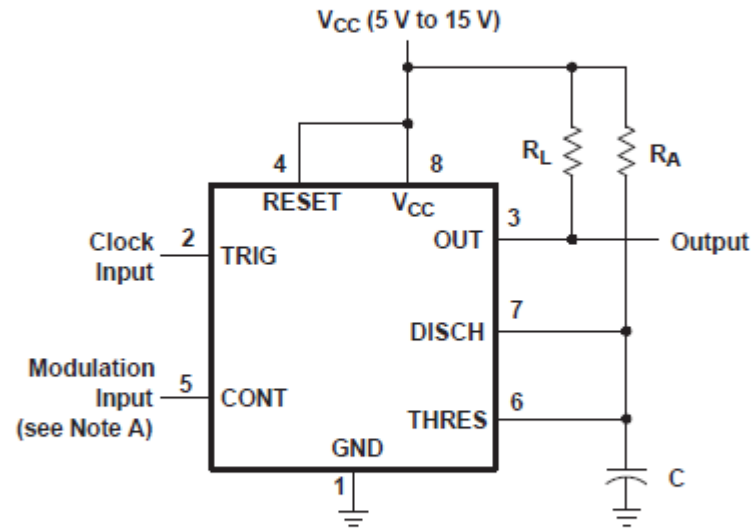


Figure 14. Free-Running Frequency

### Pulse-Width Modulation

The operation of the timer can be modified by modulating the internal threshold and trigger voltages, which is accomplished by applying an external voltage (or current) to CONT. Figure 18 shows a circuit for pulse-width modulation. A continuous input pulse train triggers the monostable circuit, and a control signal modulates the threshold voltage. Figure 19 shows the resulting output pulse-width modulation. While a sine-wave modulation signal is shown, any wave shape could be used.



Pin numbers shown are for the D, JG, P, PS, and PW packages.  
 NOTE A: The modulating signal can be direct or capacitively coupled to CONT. For direct coupling, the effects of modulation source voltage and impedance on the bias of the timer should be considered.

Figure 18. Circuit for Pulse-Width Modulation

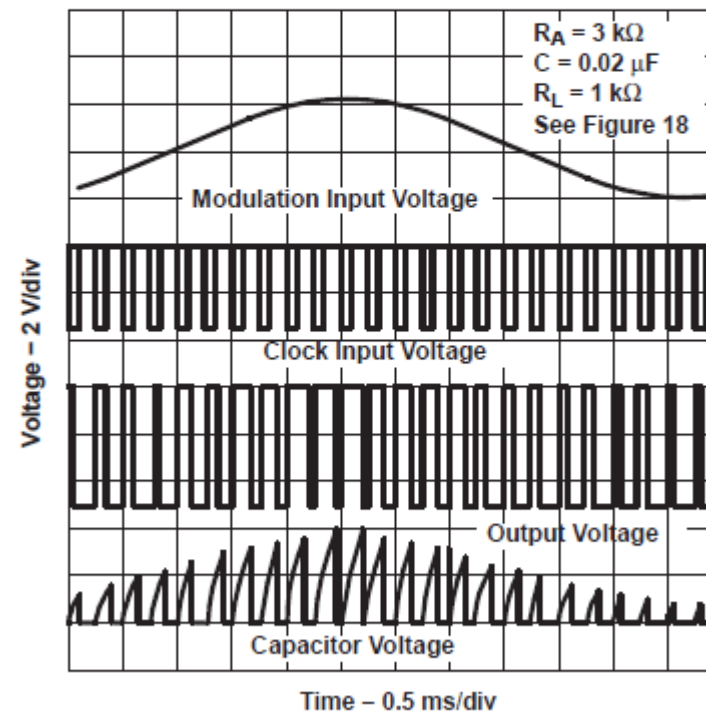
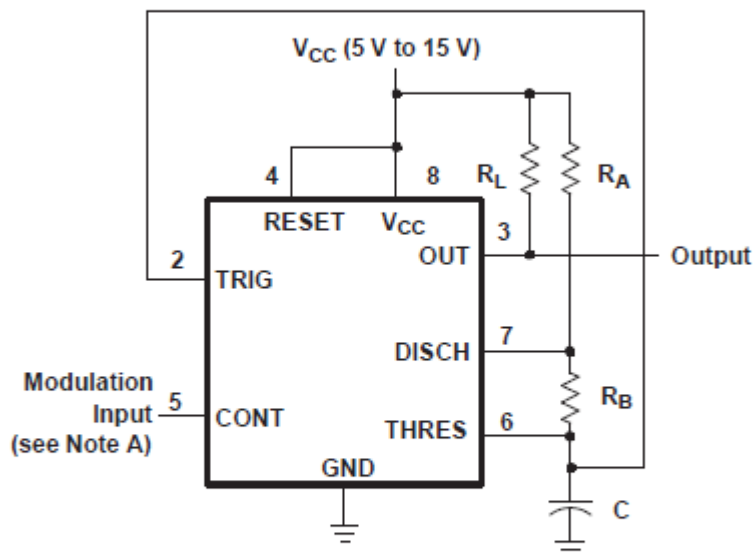


Figure 19. Pulse-Width-Modulation Waveforms

### Pulse-Position Modulation

As shown in Figure 20, any of these timers can be used as a pulse-position modulator. This application modulates the threshold voltage and, thereby, the time delay, of a free-running oscillator. Figure 21 shows a triangular-wave modulation signal for such a circuit; however, any wave shape could be used.



Pin numbers shown are for the D, JG, P, PS, and PW packages.  
 NOTE A: The modulating signal can be direct or capacitively coupled to CONT. For direct coupling, the effects of modulation source voltage and impedance on the bias of the timer should be considered.

Figure 20. Circuit for Pulse-Position Modulation

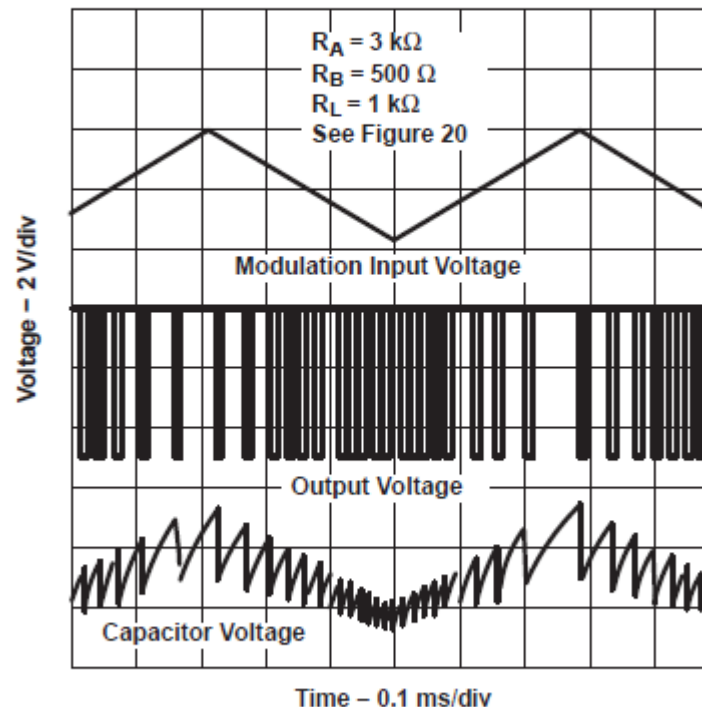
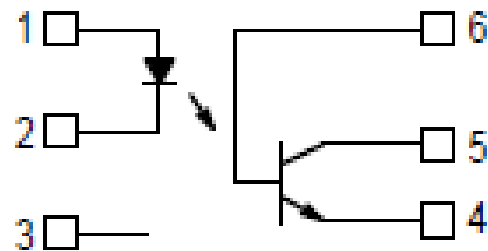


Figure 21. Pulse-Position-Modulation Waveforms

OPTO

OPTO-ACOPLADOR 4N25/4N25A/4N26/4N27/4N28

**SCHEMATIC**



- PIN 1. LED ANODE
- 2. LED CATHODE
- 3. N.C.
- 4. EMITTER
- 5. COLLECTOR
- 6. BASE

**MAXIMUM RATINGS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)

Rating	Symbol	Value	Unit
--------	--------	-------	------

**INPUT LED**

Reverse Voltage	$V_R$	3	Volts
Forward Current — Continuous	$I_F$	60	mA
LED Power Dissipation @ $T_A = 25^\circ\text{C}$ with Negligible Power in Output Detector Derate above $25^\circ\text{C}$	$P_D$	120	mW
		1.41	mW/°C

**OUTPUT TRANSISTOR**

Collector–Emitter Voltage	$V_{CEO}$	30	Volts
Emitter–Collector Voltage	$V_{ECO}$	7	Volts
Collector–Base Voltage	$V_{CBO}$	70	Volts
Collector Current — Continuous	$I_C$	150	mA
Detector Power Dissipation @ $T_A = 25^\circ\text{C}$ with Negligible Power in Input LED Derate above $25^\circ\text{C}$	$P_D$	150	mW
		1.76	mW/°C

**TOTAL DEVICE**

Isolation Surge Voltage <sup>(1)</sup> (Peak ac Voltage, 60 Hz, 1 sec Duration)	$V_{ISO}$	7500	Vac(pk)
Total Device Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	250	mW
		2.94	mW/°C
Ambient Operating Temperature Range <sup>(2)</sup>	$T_A$	-55 to +100	°C
Storage Temperature Range <sup>(2)</sup>	$T_{stg}$	-55 to +150	°C
Soldering Temperature (10 sec, 1/16" from case)	$T_L$	260	°C

**4N25 4N25A 4N26 4N27 4N28****ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)<sup>(1)</sup>

Characteristic	Symbol	Min	Typ <sup>(1)</sup>	Max	Unit	
<b>INPUT LED</b>						
Forward Voltage ( $I_F = 10 \text{ mA}$ )	$V_F$	—	$T_A = 25^\circ\text{C}$	1.15	1.5	Volts
			$T_A = -55^\circ\text{C}$	1.3	—	
			$T_A = 100^\circ\text{C}$	1.05	—	
Reverse Leakage Current ( $V_R = 3 \text{ V}$ )	$I_R$	—	—	100	$\mu\text{A}$	
Capacitance ( $V = 0 \text{ V}$ , $f = 1 \text{ MHz}$ )	$C_J$	—	18	—	pF	
<b>OUTPUT TRANSISTOR</b>						
Collector–Emitter Dark Current ( $V_{CE} = 10 \text{ V}$ , $T_A = 25^\circ\text{C}$ )	$I_{CEO}$	—	4N25,25A,26,27	1	50	nA
			4N28	1	100	
( $V_{CE} = 10 \text{ V}$ , $T_A = 100^\circ\text{C}$ )	$I_{CEO}$	—	All Devices	1	—	$\mu\text{A}$
Collector–Base Dark Current ( $V_{CB} = 10 \text{ V}$ )	$I_{CBO}$	—	—	0.2	—	nA
Collector–Emitter Breakdown Voltage ( $I_C = 1 \text{ mA}$ )	$V_{(BR)CEO}$	30	45	—	Volts	
Collector–Base Breakdown Voltage ( $I_C = 100 \mu\text{A}$ )	$V_{(BR)CBO}$	70	100	—	Volts	
Emitter–Collector Breakdown Voltage ( $I_E = 100 \mu\text{A}$ )	$V_{(BR)ECO}$	7	7.8	—	Volts	
DC Current Gain ( $I_C = 2 \text{ mA}$ , $V_{CE} = 5 \text{ V}$ )	$h_{FE}$	—	500	—	—	
Collector–Emitter Capacitance ( $f = 1 \text{ MHz}$ , $V_{CE} = 0$ )	$C_{CE}$	—	7	—	pF	
Collector–Base Capacitance ( $f = 1 \text{ MHz}$ , $V_{CB} = 0$ )	$C_{CB}$	—	19	—	pF	
Emitter–Base Capacitance ( $f = 1 \text{ MHz}$ , $V_{EB} = 0$ )	$C_{EB}$	—	9	—	pF	

## COUPLED

Output Collector Current ( $I_F = 10 \text{ mA}$ , $V_{CE} = 10 \text{ V}$ ) 4N25,25A,26 4N27,28	$I_C$ (CTR) <sup>(2)</sup>	2 (20) 1 (10)	7 (70) 5 (50)	— —	mA (%)
Collector–Emitter Saturation Voltage ( $I_C = 2 \text{ mA}$ , $I_F = 50 \text{ mA}$ )	$V_{CE(sat)}$	—	0.15	0.5	Volts
Turn–On Time ( $I_F = 10 \text{ mA}$ , $V_{CC} = 10 \text{ V}$ , $R_L = 100 \Omega$ ) <sup>(3)</sup>	$t_{on}$	—	2.8	—	$\mu\text{s}$
Turn–Off Time ( $I_F = 10 \text{ mA}$ , $V_{CC} = 10 \text{ V}$ , $R_L = 100 \Omega$ ) <sup>(3)</sup>	$t_{off}$	—	4.5	—	$\mu\text{s}$
Rise Time ( $I_F = 10 \text{ mA}$ , $V_{CC} = 10 \text{ V}$ , $R_L = 100 \Omega$ ) <sup>(3)</sup>	$t_r$	—	1.2	—	$\mu\text{s}$
Fall Time ( $I_F = 10 \text{ mA}$ , $V_{CC} = 10 \text{ V}$ , $R_L = 100 \Omega$ ) <sup>(3)</sup>	$t_f$	—	1.3	—	$\mu\text{s}$
Isolation Voltage ( $f = 60 \text{ Hz}$ , $t = 1 \text{ sec}$ ) <sup>(4)</sup>	$V_{ISO}$	7500	—	—	Vac(pk)
Isolation Resistance ( $V = 500 \text{ V}$ ) <sup>(4)</sup>	$R_{ISO}$	$10^{11}$	—	—	$\Omega$
Isolation Capacitance ( $V = 0 \text{ V}$ , $f = 1 \text{ MHz}$ ) <sup>(4)</sup>	$C_{ISO}$	—	0.2	—	pF

1. Always design to the specified minimum/maximum electrical limits (where applicable).
2. Current Transfer Ratio (CTR) =  $I_C/I_F \times 100\%$ .
3. For test circuit setup and waveforms, refer to Figure 11.
4. For this test, Pins 1 and 2 are common, and Pins 4, 5 and 6 are common.

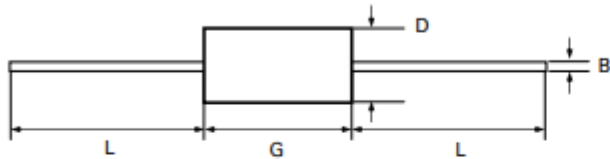
Se alimentar o LED com 10mA, no 4N25 a saída será de 20Ma





## DIODO SIDAC SÉRIE K000 LITTELFUSE

## Dimensions — DO-15 (G Package)



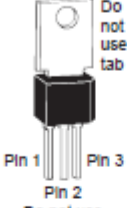



Dimension	Inches		Millimeters	
	Max	Max	Min	Max
B	0.028	0.034	0.711	0.864
D	0.120	0.140	3.048	3.556
G	0.235	0.270	5.969	6.858
L	1.000		25.400	

## Product Selector

Part Number	Switching Voltage Range		Blocking Voltage	Packages
	$V_{BO}$ Minimum	$V_{BO}$ Maximum	$V_{DRM}$	
K2201G	200V	230V	180V	DO-15
K2401G	220V	250V	190V	DO-15
K2501G	240V	280V	200V	DO-15
K3601G	340V	380V	270V	DO-15

Onde comprar: [https://produto.mercadolivre.com.br/MLB-3065555182-diodo-sidac-k220g-littelfuse-JM#position=7&search\\_layout=stack&type=item&tracking\\_id=d20dc5e3-9686-4e31-85f7-32db44dd2f9e](https://produto.mercadolivre.com.br/MLB-3065555182-diodo-sidac-k220g-littelfuse-JM#position=7&search_layout=stack&type=item&tracking_id=d20dc5e3-9686-4e31-85f7-32db44dd2f9e)

Type	Part No.				$I_{T(RMS)}$	$V_{DRM}$	$V_{BO}$		$I_{DRM}$	$I_{BO}$	$I_H$	
	 TO-92	 DO-15X	(10)  Do not use tab Pin 1 Pin 3 Pin 2 Do not use TO-202	 DO-214	(6) (7) (8)  Amps	  Volts	(1)  Volts		  $\mu$ Amps	(2)  $\mu$ Amps	(3) (4)  mAmps	
See "Package Dimensions" section for variations. (9)					MAX	MIN	MIN	MAX	MAX	MAX	TYP	MAX
	K0900E70	K0900G		K0900S	1	$\pm 70$	79	97	5	10	60	150
	K1050E70	K1050G		K1050S	1	$\pm 90$	95	113	5	10	60	150
	K1100E70	K1100G		K1100S	1	$\pm 90$	104	118	5	10	60	150
	K1200E70	K1200G		K1200S	1	$\pm 90$	110	125	5	10	60	150
	K1300E70	K1300G		K1300S	1	$\pm 90$	120	138	5	10	60	150
	K1400E70	K1400G		K1400S	1	$\pm 90$	130	146	5	10	60	150
	K1500E70	K1500G		K1500S	1	$\pm 90$	140	170	5	10	60	150
	K2000E70	K2000G	K2000F1	K2000S	1	$\pm 180$	190	215	5	10	60	150
	K2200E70	K2200G	K2200F1	K2200S	1	$\pm 180$	205	230	5	10	60	150
	K2400E70	K2400G	K2400F1	K2400S	1	$\pm 190$	220	250	5	10	60	150
			K2401F1		(11)	$\pm 190$	220	250	5	10	60	150
	K2500E70	K2500G	K2500F1	K2500S	1	$\pm 200$	240	280	5	10	60	150
			K3000F1		1	$\pm 200$	270	330	5	10	60	150

SIDACTOR SIDAC LITTELFUSE SÉRIE P0000

**SIDACTor® Device**



TO-92 *SIDACTor* solid state protection devices protect telecommunications equipment such as modems, line cards, and CPE (telephones, answering machines, and fax machines).

*SIDACTor* devices enable equipment to comply with various regulatory requirements including GR 1089, ITU K.20, K.21 and K.45, IEC 60950, UL 60950, and TIA-968-A (formerly known as FCC Part 68).

SIDACTor Devices

Electrical Parameters

Part Number *	V <sub>DRM</sub> Volts	V <sub>s</sub> Volts	V <sub>T</sub> Volts	I <sub>DRM</sub> μAmps	I <sub>s</sub> mAmps	I <sub>T</sub> Amps	I <sub>H</sub> mAmps
P0080E_L	6	25	4	5	800	2.2	50
P0300E_L	25	40	4	5	800	2.2	50
P0640E_L	58	77	4	5	800	2.2	150
P0720E_L	65	88	4	5	800	2.2	150
P0900E_L	75	98	4	5	800	2.2	150
P1100E_L	90	130	4	5	800	2.2	150
P1300E_L	120	160	4	5	800	2.2	150
P1500E_L	140	180	4	5	800	2.2	150
P1800E_L	170	220	4	5	800	2.2	150
P2300E_L	190	260	4	5	800	2.2	150
P2600E_L	220	300	4	5	800	2.2	150
P3100E_L	275	350	4	5	800	2.2	150
P3500E_L	320	400	4	5	800	2.2	150

Onde comprar: [https://produto.mercadolivre.com.br/MLB-2044642855-tiristor-sidac-p3100ealrp2-to-92-2-JM#position=4&search\\_layout=stack&type=item&tracking\\_id=6e999ec4-526a-4ae2-9c5c-8eb27589f7d9](https://produto.mercadolivre.com.br/MLB-2044642855-tiristor-sidac-p3100ealrp2-to-92-2-JM#position=4&search_layout=stack&type=item&tracking_id=6e999ec4-526a-4ae2-9c5c-8eb27589f7d9)

T10A SIDACTor® Device



The bi-directional T10A devices are a through-hole technology SIDACTor protector. It is intended for cost-sensitive telecommunication applications. This T10 SIDACTor series enables equipment to comply with various regulatory requirements including GR 1089, ITU K.20, K.21, and K.45, IEC 60950, UL 60950, and TIA-968-A (formerly known as FCC Part 68).

SIDACTor Devices

Electrical Parameters

Part Number *	V <sub>DRM</sub> @ 5 μA Volts	V <sub>S</sub> Volts	V <sub>T</sub> Volts	I <sub>S</sub> mAmps	I <sub>H</sub> mAmps	pF TYP
T10A060B	58	80	4	800	120	50
T10A060E	58	80	4	800	180	50
T10A062	60	82	4	800	150	50
T10A068	65	90	4	800	150	50
T10A080B	75	120	4	800	120	43
T10A080E	75	120	4	800	180	43
T10A100	100	133	4	800	150	43
T10A110B	110	135	4	800	120	38
T10A110E	110	135	4	800	180	38
T10A120	120	160	4	800	150	38
T10A130	130	173	4	800	150	38
T10A140B	140	170	4	800	120	34
T10A140E	140	170	4	800	180	34
T10A180	180	240	4	800	150	34
T10A180B	175	210	4	800	120	32
T10A180E	175	210	4	800	180	32
T10A200	200	267	4	800	150	30
T10A220	220	293	4	800	150	30
T10A220B	215	265	4	800	120	30
T10A220E	215	265	4	800	180	30
T10A240	240	320	4	800	150	30
T10A270	270	360	4	800	150	30
T10A270B	270	360	4	800	120	30
T10A270E	270	360	4	800	180	30

[https://produto.mercadolivre.com.br/MLB-](https://produto.mercadolivre.com.br/MLB-3204374806-diodo-sidactor-t10a180e-_JM#position=7&search_layout=stack&type=item&tracking_id=77186c4d-872c-4609-85d0-9246f9ef7d5a)

3204374806-diodo-sidactor-t10a180e-\_JM#position=7&search\_layout=stack&type=item&tracking\_id=77186c4d-872c-4609-85d0-9246f9ef7d5a

## DISJUNTORES, TIPO E CORRENTE CURVA CARACTERÍSTICA

<https://www.slideshare.net/akerman22/disjuntores-siemens>

In (A)	Tipo		
	Curva C (disparo em curto-circuito 5 a 10 x In) monopolar	bipolar	tripolar
0.5	5SX1 105-7	5SX1 205-7	5SX1 305-7
1	5SX1 101-7	5SX1 201-7	5SX1 301-7
2	5SX1 102-7	5SX1 202-7	5SX1 302-7
4	5SX1 104-7	5SX1 204-7	5SX1 304-7
6	5SX1 106-7	5SX1 206-7	5SX1 306-7
10	5SX1 110-7	5SX1 210-7	5SX1 310-7
13	5SX1 113-7	5SX1 213-7	5SX1 313-7
16	5SX1 116-7	5SX1 216-7	5SX1 316-7
20	5SX1 120-7	5SX1 220-7	5SX1 320-7
25	5SX1 125-7	5SX1 225-7	5SX1 325-7
32	5SX1 132-7	5SX1 232-7	5SX1 332-7
40	5SX1 140-7	5SX1 240-7	5SX1 340-7
50	5SX1 150-7	5SX1 250-7	5SX1 350-7
63	5SX1 163-7	5SX1 263-7	5SX1 363-7
70	5SX1 170-7	5SX1 270-7	5SX1 370-7
80	5SX1 180-1	5SX1 280-1	5SX1 380-1

*Nota: O disjuntor 5SX1 de 80A possui somente a proteção contra curto-circuito. Para proteção contra sobrecarga faz-se necessário a utilização de um outro dispositivo complementar.*

## A. Disjuntores 5SX1

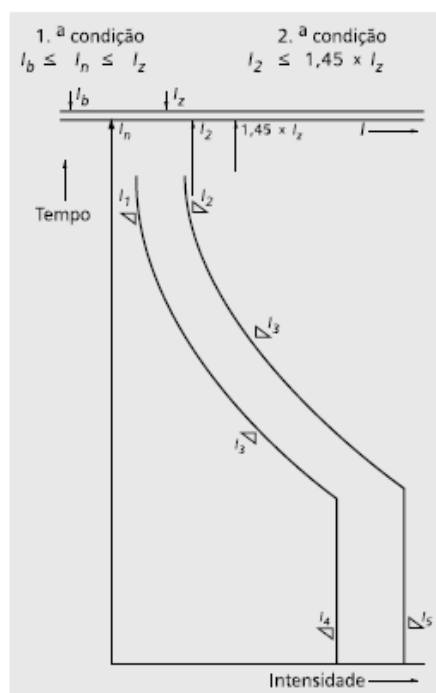
In (A)	Tipo	
	Curva B (disparo em curto-circuito 3 a 5 x In) Monopolar	Bipolar
6	5SX1 106-6	5SX1 206-6
10	5SX1 110-6	5SX1 210-6
13	5SX1 113-6	5SX1 213-6
16	5SX1 116-6	5SX1 216-6
20	5SX1 120-6	5SX1 220-6
25	5SX1 125-6	5SX1 225-6
32	5SX1 132-6	5SX1 232-6
In	Tipo	

Material retirado do catálogo de mini disjuntores da Siemens 2022

# Curvas características de disparo

## Curvas características

A função dos disjuntores termomagnéticos é a proteção dos condutores contra sobrecargas térmicas ou curto-circuitos. É por isso que as curvas de disparo dos disjuntores se adaptam às curvas dos condutores.



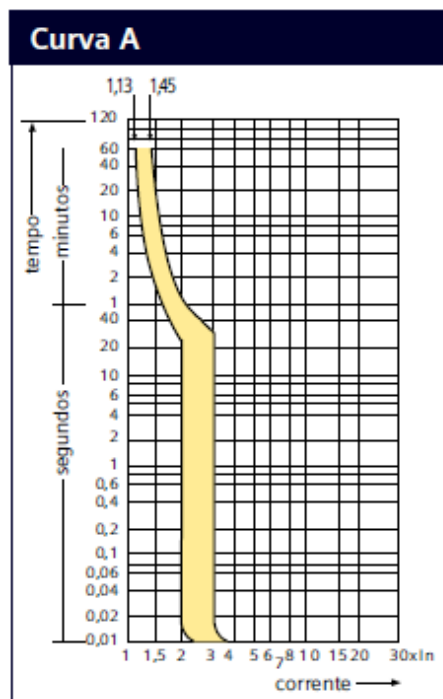
Na representação da figura ao lado, são coordenados os valores de referência dos condutores com os disjuntores termomagnéticos. Na Norma NBR NM 60898, são definidas as características, curvas B, C e D.

Deve-se cumprir para uma boa seleção, a seguinte fórmula:

$I_b < I_n < I_z$  e além disso, que  $I_2 < 1,45 \times I_z$

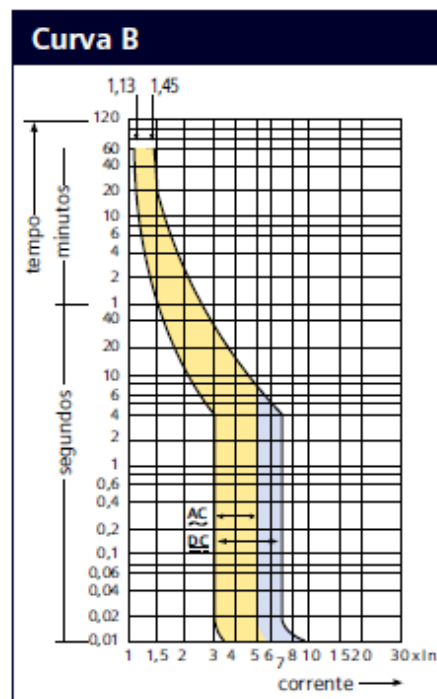
Onde:

- $I_b$  = Corrente de projeto do circuito.
- $I_n$  = Corrente nominal do disjuntor termomagnético, nas condições previstas na instalação.
- $I_z$  = Capacidade de condução de corrente dos condutores, nas condições previstas para sua instalação.
- $1,45 \times I_z$  = Corrente de sobrecarga máxima permitida, para uma condição de temperatura excedida, sem que haja o comprometimento do isolante dos condutores.
- $I_1$  = Corrente convencional de não atuação na sobrecarga.
- $I_2$  = Corrente convencional de atuação na sobrecarga
- $I_3$  = Limite de tolerância do disparador
- $I_4$  = Corrente convencional de não atuação no curto-circuito.
- $I_5$  = Corrente convencional de atuação no curto-circuito.

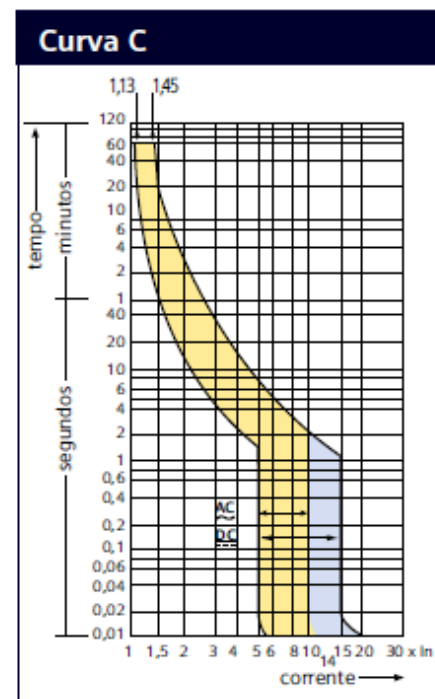


**Curva A:** Para proteção de circuitos com semicondutores e circuitos de medição.

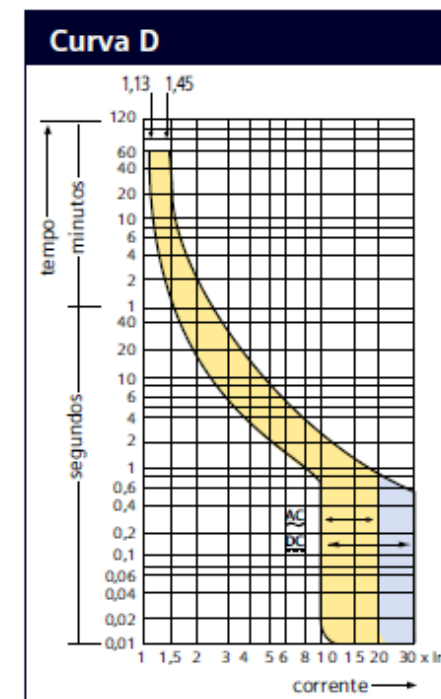
\* Não designada na NBR NM 60898



**Curva B:** Para proteção de circuitos que alimentam cargas com características predominantemente resistivas, como lâmpadas incandescentes, chuveiros, torneiras e aquecedores elétricos, além dos circuitos de tomadas de uso geral.



**Curva C:** Para proteção de circuitos que alimentam especificamente cargas de natureza indutiva que apresentam picos de corrente no momento de ligação, como microondas, ar condicionado, motores para bombas, além de circuitos com cargas de características semelhantes a essas.



**Curva D:** Para proteção de circuitos que alimentam cargas altamente indutivas que apresentam elevados picos de corrente no momento de ligação, como grandes motores, transformadores, além de circuitos com cargas de características semelhantes a essas.



## TABELA SELEÇÃO DE INDUTORES

Table 2. Inductor Selection by Manufacturer's Part Number

Inductor Code	Inductor Value	Tech 39	Schott Corp.	Pulse Eng.	Renco
L47	47 $\mu$ H	77 212	671 26980	PE-53112	RL2442
L68	68 $\mu$ H	77 262	671 26990	PE-92114	RL2443
L100	100 $\mu$ H	77 312	671 27000	PE-92108	RL2444
L150	150 $\mu$ H	77 360	671 27010	PE-53113	RL1954
L220	220 $\mu$ H	77 408	671 27020	PE-52626	RL1953
L330	330 $\mu$ H	77 456	671 27030	PE-52627	RL1952
L470	470 $\mu$ H	*	671 27040	PE-53114	RL1951
L680	680 $\mu$ H	77 506	671 27050	PE-52629	RL1950
H150	150 $\mu$ H	77 362	671 27060	PE-53115	RL2445
H220	220 $\mu$ H	77 412	671 27070	PE-53116	RL2446
H330	330 $\mu$ H	77 462	671 27080	PE-53117	RL2447
H470	470 $\mu$ H	*	671 27090	PE-53118	RL1961
H680	680 $\mu$ H	77 508	671 27100	PE-53119	RL1960
H1000	1000 $\mu$ H	77 556	671 27110	PE-53120	RL1959
H1500	1500 $\mu$ H	*	671 27120	PE-53121	RL1958
H2200	2200 $\mu$ H	*	671 27130	PE-53122	RL2448

NOTE: \*Contact Manufacturer

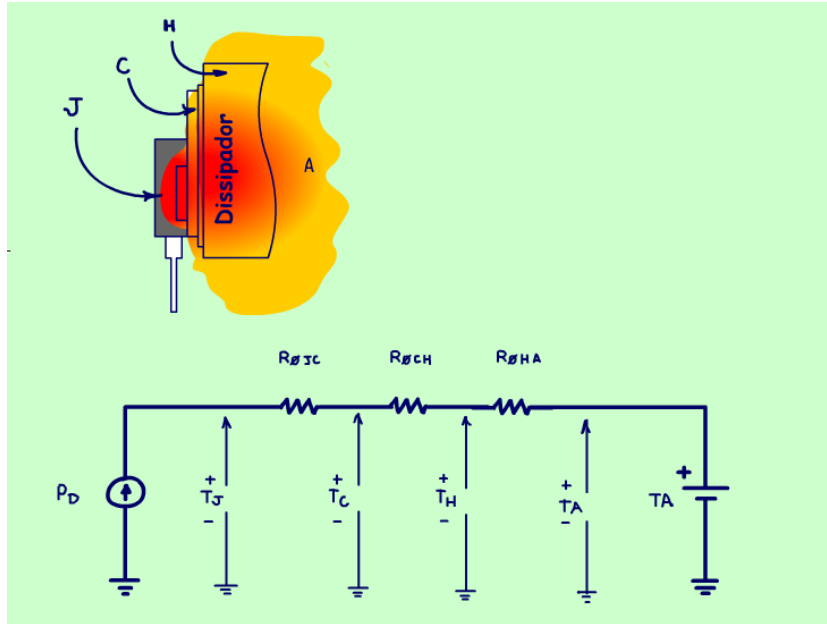
FONTE: Manual do LM2576

## TABELA SELEÇÃO INDUTORES E CAPACITORES PARA CIRCUITO DE SINTONIA AM.

**Capacitor for Ferrite Rod Coil for AM radio**

uH	pF
165	540 (270 x 2)
250	330
260	330
330	270
470	180
570	155
680	140

## DISSIPADORES



Esse é o circuito básico para o uso do dissipador de calor.

$P_D$  é a potência no transistor, essa será a corrente térmica nesse circuito.

$R_{\theta}$  é a resistência térmica.

A temperatura em cada resistência térmica é a tensão, o produto da corrente térmica e da resistência térmica.

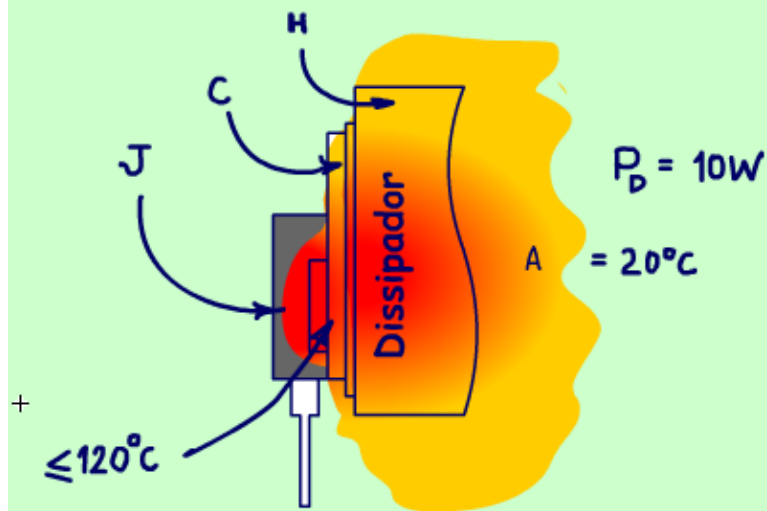
$R_{\theta JC}$  é a resistência térmica entre a junção e o case (carcaça metálica do transistor)

$R_{\theta CH}$  é a resistência térmica entre a carcaça e o dissipador, é a resistência térmica do isolante se tiver.

$R_{\theta HA}$  é a resistência térmica entre o dissipador e o meio ambiente.

$T_A$  é a temperatura ambiente.

Exemplo

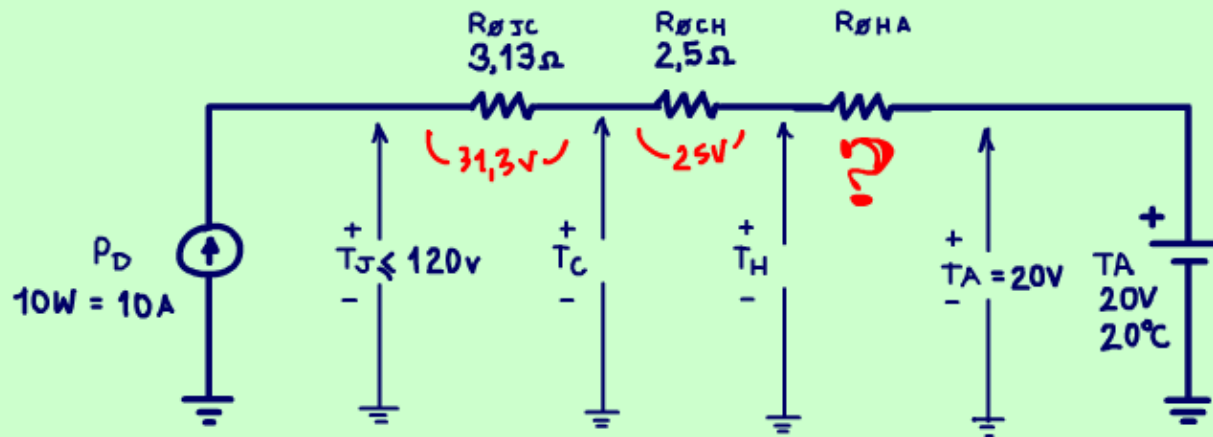


$$V_{JC} = R_{\theta JC} \cdot I = 3,13 \cdot 10 = 31,3V$$

$$V_{CH} = R_{\theta CH} \cdot I = 2,5 \cdot 10 = 25V$$

$$V_{HA} \leq 120V - 31,3V - 25V - Z_0$$

$$R_{\theta HA} \leq \frac{43,7}{10A} \leq 4,37 \Omega$$



## RESISTÊNCIA TÉRMICA TÍPICA DOS ISOLADORES

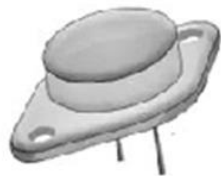
Fonte Pomilo.

Tabela 11.4 Valores típicos de resistência térmica entre cápsula e dissipador

Tipo de cápsula	Tipo de isolador	R <sub>ted</sub> (°C/W) (R <sub>ecw</sub> )	
		c/ pasta	s/ pasta
TO - 3	s/ isolador	0,1	0,3
	teflon	0,7 a 0,8	1,25 a 1,45
	mica	0,5 a 0,7	1,2 a 1,5
TO - 66	s/ isolador	0,15 a 0,2	0,4 a 0,5
	mica	0,6 a 0,8	1,5 a 2,0
	mylar	0,6 a 0,8	1,2 a 1,4
TO - 220AB	s/ isolador	0,3 a 0,5	1,5 a 2,0
	mica	2,0 a 2,5	4,0 a 6,0

Obs.: mica e mylar com espessura de 50 µm a 100 µm.

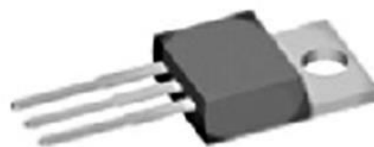
TO-3



TO-56



TO-220AB



CORREÇÃO DO COMPRIMENTO DISSIPADORES DA HS

# ÍNDICE REMISSIVO

(clique sobre o nome para navegar até a página)

- nas páginas 8 a 14 encontra-se um índice com desenhos -

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HS 1508	19,8 °C/W/4"	.....	página 22
HS 1509	19,8 °C/W/4"	.....	página 23
HS 1511	15,5 °C/W/4"	.....	página 24
HS 1515	20,6 °C/W/4"	.....	página 25
HS 1616	9,0 °C/W/4"	.....	página 26
HS 1616L	10,6 °C/W/4"	.....	página 27
HS 1710	14,6 °C/W/4"	.....	página 28
HS 1807	19,8 °C/W/4"	.....	página 29
HS 1818	13,4 °C/W/4"	.....	página 30
HS 1920	11,5 °C/W/4"	.....	página 31
HS 2053	4,2 °C/W/4"	.....	página 32
HS 2053E	4,2 °C/W/4"	.....	página 33
HS 2315	10,2 °C/W/4"	.....	página 34
HS 2315M	10,2 °C/W/4"	.....	página 35
HS 2811	10,0 °C/W/4"	.....	página 36
HS 2816	7,9 °C/W/4"	.....	página 37
HS 3030	5,7 °C/W/4"	.....	página 38
HS 3125	6,2 °C/W/4"	.....	página 39
HS 3232	6,3 °C/W/4"	.....	página 40
HS 3512	8,4 °C/W/4"	.....	página 41
HS 3520	4,9 °C/W/4"	.....	página 42
HS 3542	3,2 °C/W/4"	.....	página 43
HS 3542L	3,9 °C/W/4"	.....	página 44
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HS 4017	7,8 °C/W/4"	.....	página 46
HS 4225	4,4 °C/W/4"	.....	página 47
HS 4262	3,7 °C/W/4"	.....	página 48
HS 4313	8,9 °C/W/4"	.....	página 49
HS 4320	4,1 °C/W/4"	.....	página 50
HS 4328	3,1 °C/W/4"	.....	página 51
HS 4425	4,4 °C/W/4"	.....	página 52
HS 4525	4,4 °C/W/4"	.....	página 53
HS 5073	2,28 °C/W/4"	.....	página 54
HS 5620	2,9 °C/W/4"	.....	página 55
HS 6524	3,7 °C/W/4"	.....	página 56
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HS 7021	4,5 °C/W/4"	.....	página 60
HS 7028	2,7 °C/W/4"	.....	página 61
HS 7032	3,6 °C/W/4"	.....	página 62
HS 7223	3,2 °C/W/4"	.....	página 63
HS 7245	2,7 °C/W/4"	.....	página 64

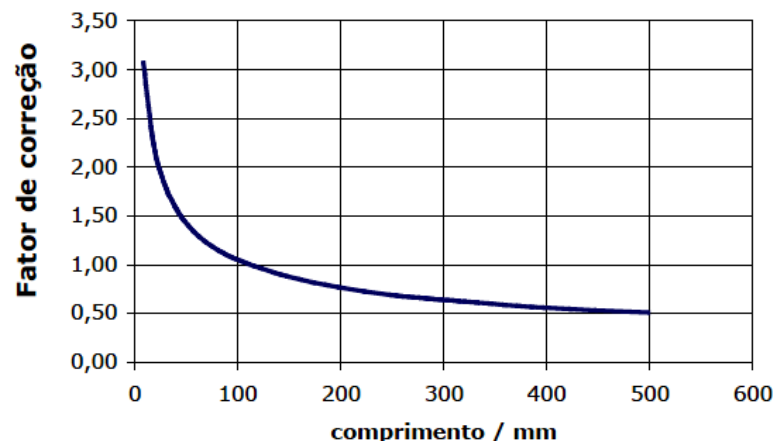
clique aqui para ir ao índice

PRETO = perfis normalmente disponíveis  
 VERDE = esperamos disponibilizar em breve  
 VERMELHO = perfis indisponíveis (\*)

(\*) podemos enviar alguns dados perfis indisponíveis mediante

HS 8550	2,1 °C/W/4"	.....	página 69
HS 8585	1,0 °C/W/4"	.....	página 70
HS 8585T	1,0 °C/W/4"	.....	página 71
HS 8620	2,92 °C/W/4"	.....	página 72
HS 8620L	3,20 °C/W/4"	.....	página 73
HS 8858	1,74 °C/W/4"	.....	página 74
HS 9438	1,4 °C/W/4"	.....	página 75
HS 9555	1,29 °C/W/4"	.....	página 76
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HS 10334	1,5 °C/W/4"	.....	página 78
HS 10334L	1,9 °C/W/4"	.....	página 79
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HS 10526	2,5 °C/W/4"	.....	página 82
HS 11330	1,5 °C/W/4"	.....	página 83
HS 11432	1,7 °C/W/4"	.....	página 84
HS 11450	1,4 °C/W/4"	.....	página 85
HS 11550	1,3 °C/W/4"	.....	página 86
HS 11555	1,17 °C/W/4"	.....	página 87
HS 11960	1,41 °C/W/4"	.....	página 88
HS 12060	1,17 °C/W/4"	.....	página 89
HS 12135	1,92 °C/W/4"	.....	página 90
HS 12135N	1,88 °C/W/4"	.....	página 91
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HS 12764L	1,22 °C/W/4"	.....	página 102
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HS 14376	1,11 °C/W/4"	.....	página 107
HS 14569	1,17 °C/W/4"	.....	página 108
HS 14676	1,02 °C/W/4"	.....	página 109
HS 15073	1,24 °C/W/4"	.....	página 110
HS 15450	1,22 °C/W/4"	.....	página 111
HS 15559	0,73 °C/W/4"	.....	página 112
HS 17050	2,14 °C/W/4"	.....	página 115
HS 17232L	1,07 °C/W/4"	.....	página 116
HS 17909	2,06 °C/W/4"	.....	página 117
HS 18034	2,9 °C/W/4"	.....	página 118
HS 19013	1,79 °C/W/4"	.....	página 119
HS 19032	0,98 °C/W/4"	.....	página 120
HS 19334	1,07 °C/W/4"	.....	página 121
HS 20855	1,03 °C/W/4"	.....	página 122
HS 21019	1,64 °C/W/4"	.....	página 123
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HS 21060	0,83 °C/W/4"	.....	página 125
HS 21073	0,66 °C/W/4"	.....	página 126
HS 21526	1,07 °C/W/4"	.....	página 127
HS 21575	0,56 °C/W/4"	.....	página 128
HS 21577E	0,52 °C/W/4"	.....	página 129
HS 23742	0,57 °C/W/4"	.....	página 130
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HS 125135L	0,73 °C/W/4"	.....	página 140
HS 125136	0,74 °C/W/4"	.....	página 142
HS 125137	0,66 °C/W/4"	.....	página 144

## correção da resistência térmica para outros comprimentos de dissipadores



Exemplo: O dissipador HS 6524 possui resistência térmica igual a 3,72 °C/W/4". Isto quer dizer que um pedaço medindo 4 polegadas de comprimento possui a resistência térmica de 3,72 °C/W. O mesmo dissipador com 40mm de comprimento possui resistência térmica igual a

$$3,72 \text{ °C/W} \times 1,59 = 5,91 \text{ °C/W}$$

Obs: 4''=10,16cm

## ÍNDICE REMISSIVO DISSIPADORES DA HS

HS 7624	3,6 °C/W/4"	.....	página 66
HS 8044	2,0 °C/W/4"	.....	página 67
HS 8134	2,4 °C/W/4"	.....	página 68
HS 15560L	1,01 °C/W/4"	.....	página 114

comprimento	
10 mm	
20 mm	
30 mm	
40 mm	
50 mm	
70 mm	
100 mm	
150 mm	
200 mm	
250 mm	
300 mm	
400 mm	
500 mm	

DISSIPADOR HS2315 10,2°C/W/4"

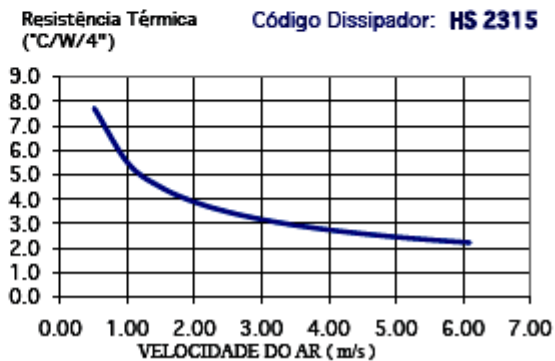
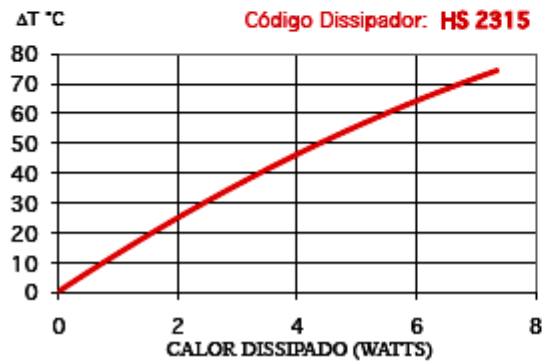
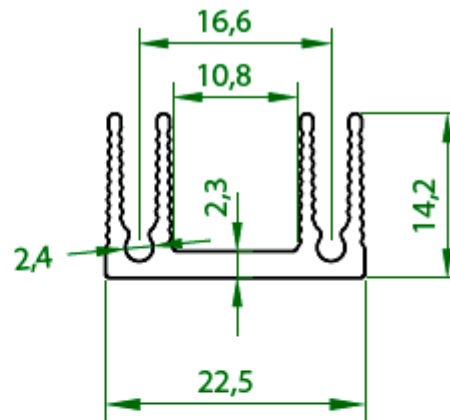
# Código: HS 2315

Perímetro: 151 mm

Resistência Térmica: 10,2 °C / W / 4"

Peso Linear: 0,28 kg/m

Capacidade Térmica: 921 J/kg K



DISSIPADOR HS3520 4,89°C/W/4"



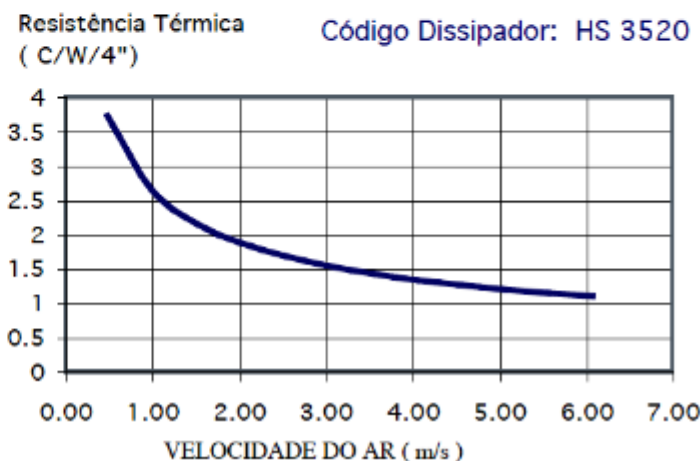
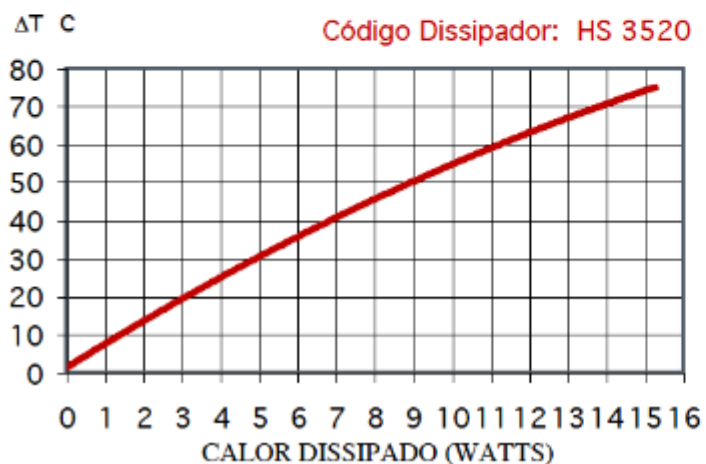
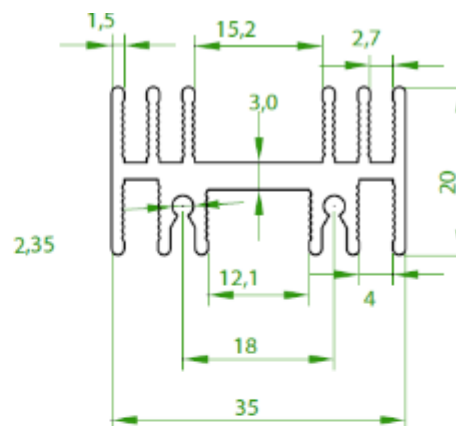
# Código: HS 3520

Perímetro: 316 mm

Resistência Térmica: 4,89 °C / W / 4"

Peso Linear: 0,71 kg/m

Capacidade Térmica: 921 J/kg K



DISSIPADOR HS8620 L 3,20 °C/W/4"

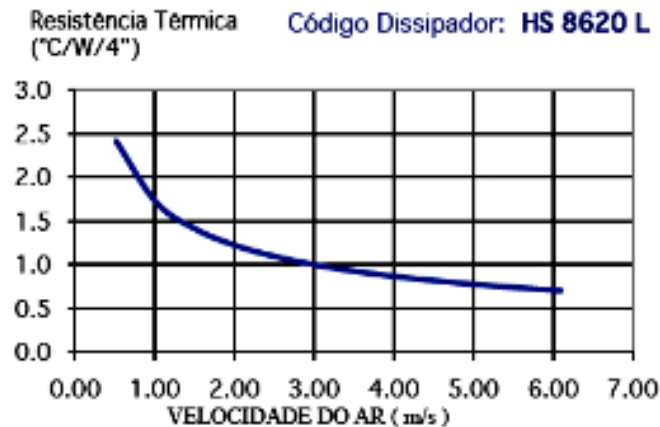
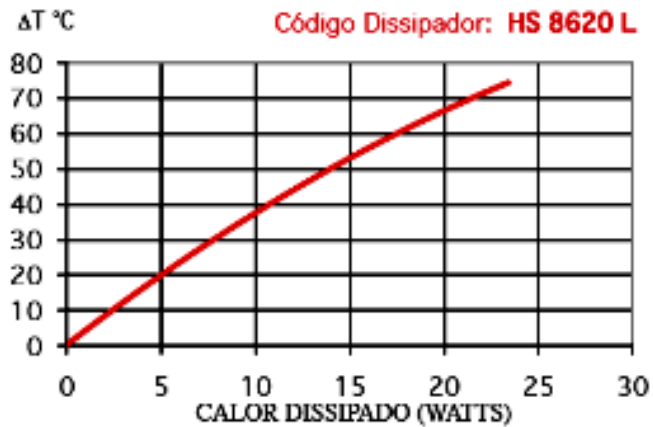
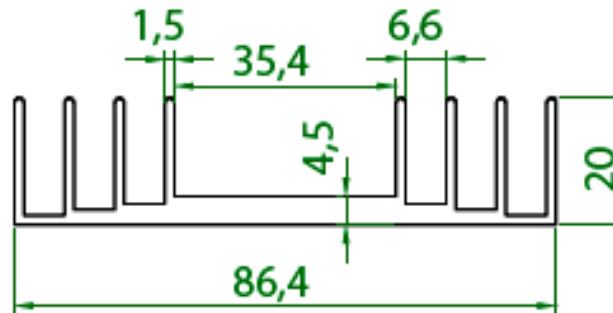
# Código: HS 8620 L

Perímetro: 484 mm

Resistência Térmica: 3,20 °C / W / 4"

Peso Linear: 1,6 kg/m

Capacidade Térmica: 921 J/kg K



DISSIPADOR HS12135 1,92 °C/W/4"

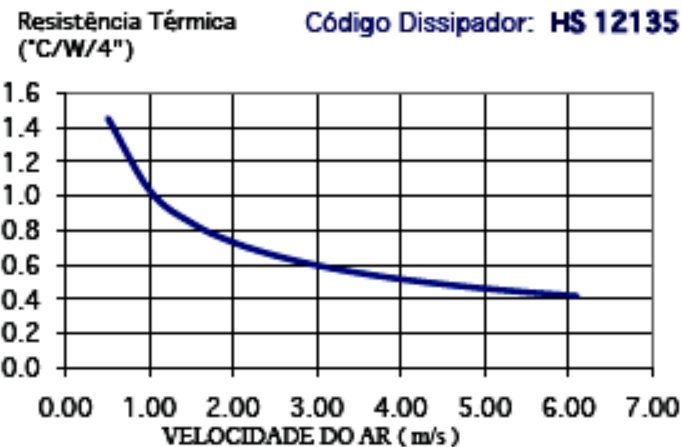
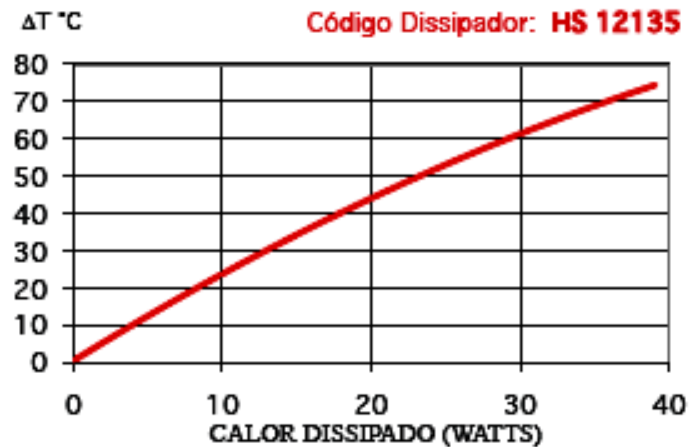
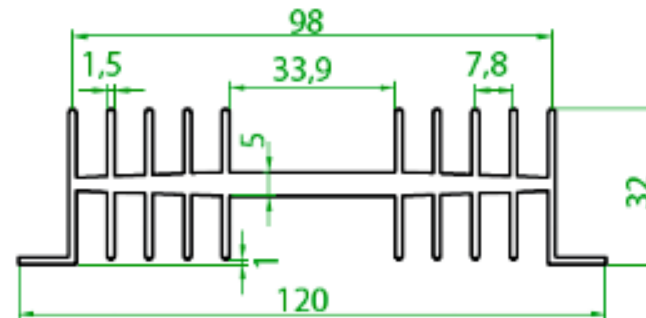
# Código: HS 12135

Perímetro: 804 mm

Resistência Térmica: 1,92 °C / W / 4"

Peso Linear: 2,6 kg/m

Capacidade Térmica: 921 J/kg K



DISSIPADOR HS10425 1,75°C/W/4"

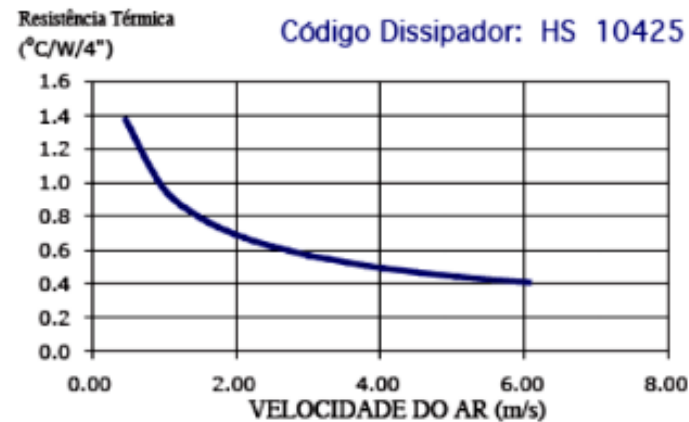
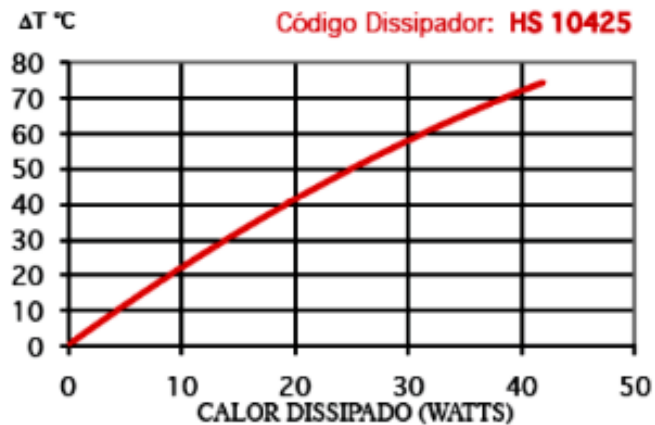
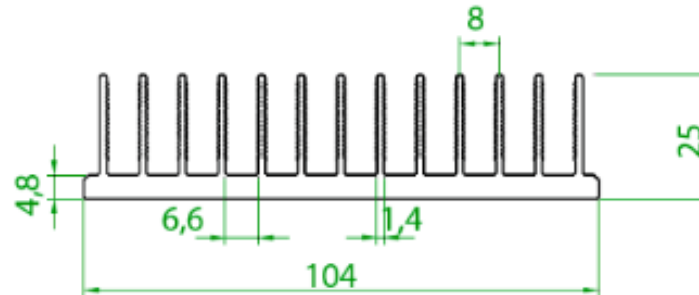
# Código: HS 10425

Perímetro: 862 mm

Resistência Térmica: 1,79 °C / W / 4"

Peso Linear: 2,3 kg/m

Capacidade Térmica: 921 J/kg K



DISSIPADOR HS1643 1,72°C/W/4"

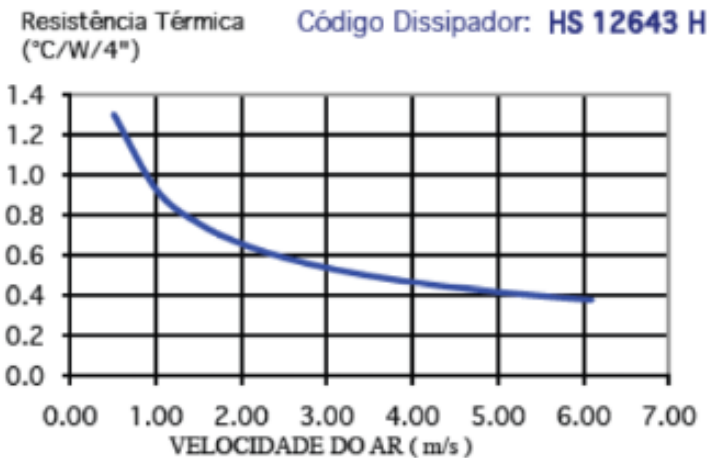
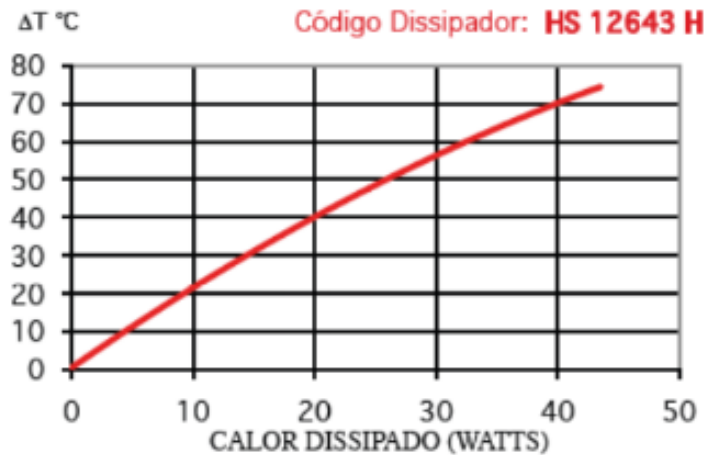
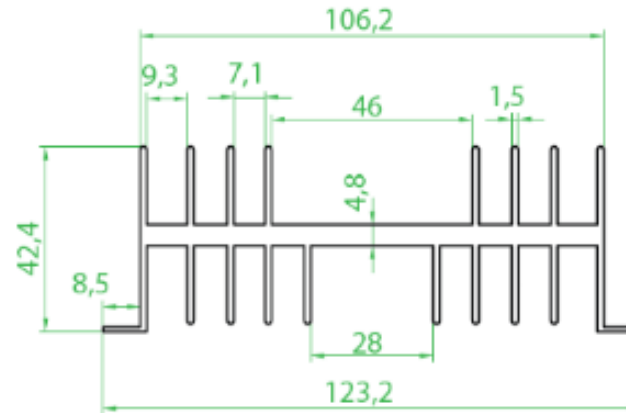
# Código: HS 12643 H

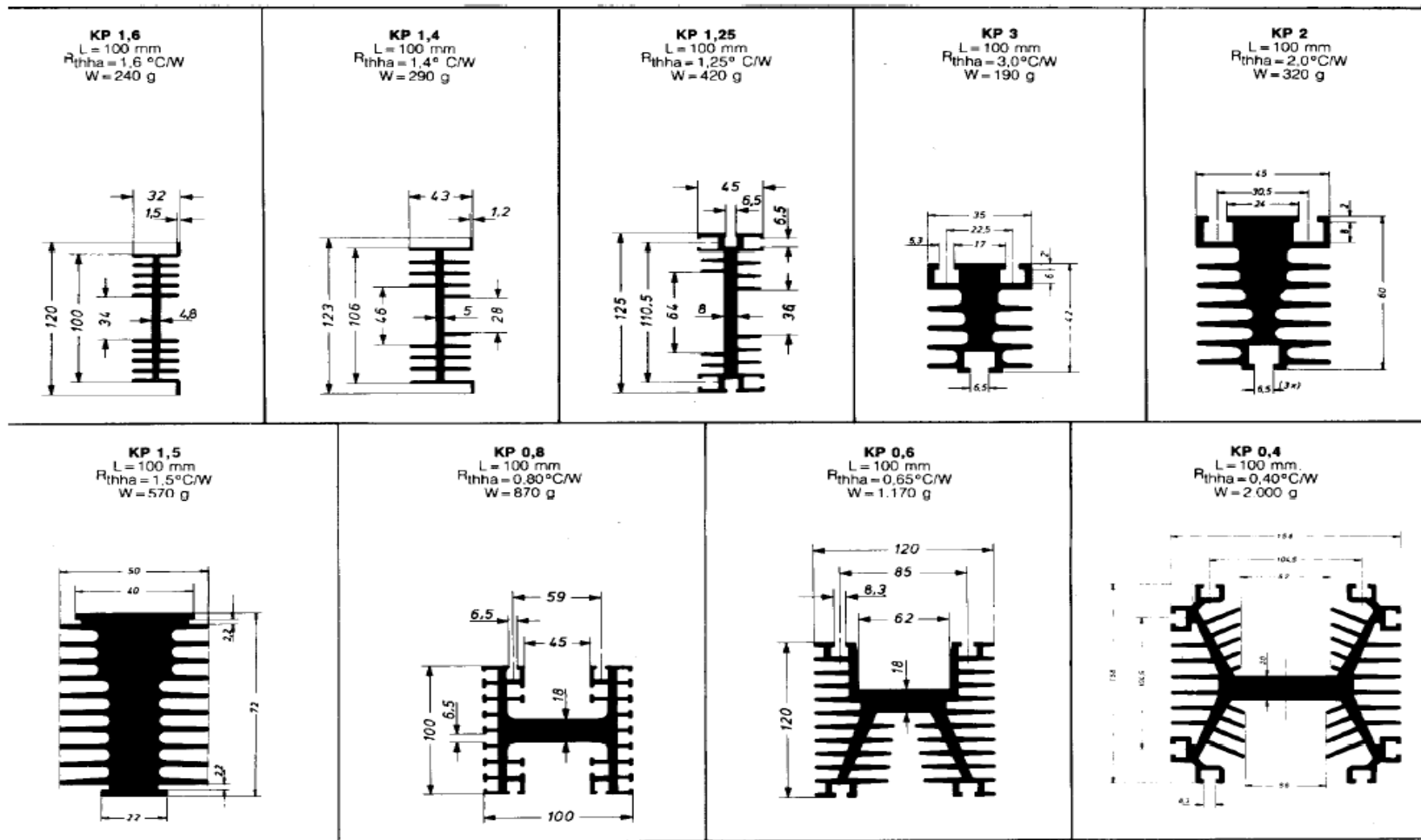
Perímetro: 897 mm

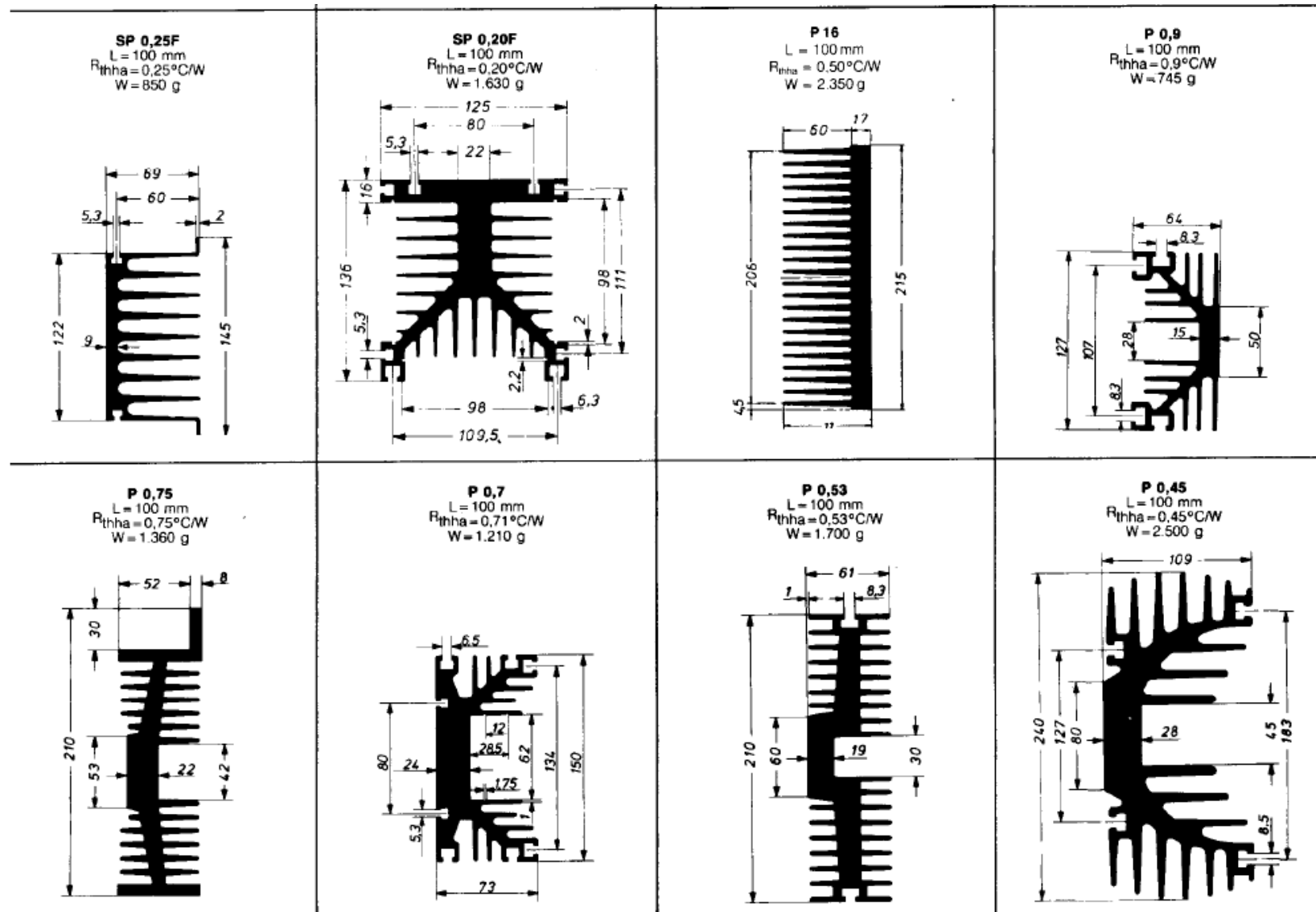
Resistência Térmica: 1,72 °C / W / 4"

Peso Linear: 2,9 kg/m

Capacidade Térmica: 921 J/kg K





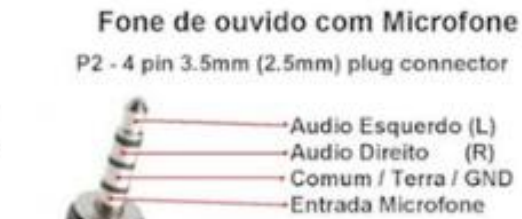


PLUGUES JACKS E CONECTORES

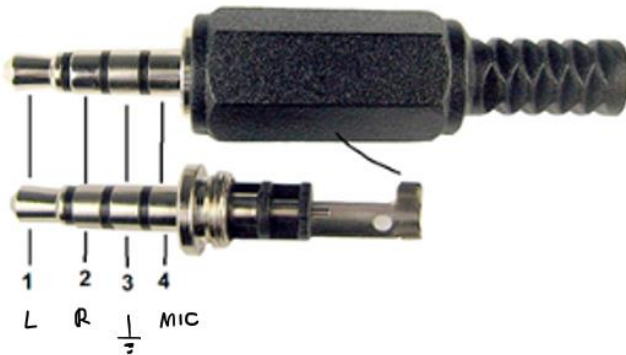


PLUGUE P2 SOM DO PC

## Esquema de Ligação dos Pinos P2



## CONEXÃO PLUGUE P3 NO CELULAR.



Estes plugues de 3,5 mm são usados na maioria dos celulares.

Identificação dos terminais:

- 1 - Saída do canal esquerdo do fone
- 2 - Saída do canal direito do fone
- 3 - Comum (terra)
- 4 - Entrada do microfone

AMPLIFICADOR OPERACIONAL (AMPOP)

TL071 TL072 TL074

**TL071, TL071A, TL071B, TL072  
TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**  
SLOS080J – SEPTEMBER 1978 – REVISED MARCH 2005

- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion . . . 0.003% Typ
- Low Noise  
 $V_n = 18 \text{ nV}/\sqrt{\text{Hz}}$  Typ at  $f = 1 \text{ kHz}$
- High Input Impedance . . . JFET Input Stage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- High Slew Rate . . .  $13 \text{ V}/\mu\text{s}$  Typ
- Common-Mode Input Voltage Range Includes  $V_{CC+}$

#### description/ordering information

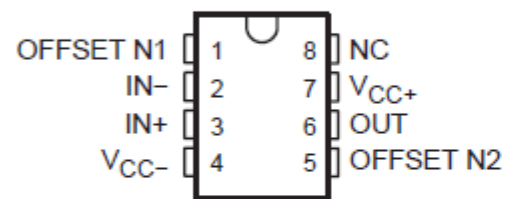
The JFET-input operational amplifiers in the TL07x series are similar to the TL08x series, with low input bias and offset currents and fast slew rate. The low harmonic distortion and low noise make the TL07x series ideally suited for high-fidelity and audio preamplifier applications. Each amplifier features JFET inputs (for high input impedance) coupled with bipolar output stages integrated on a single monolithic chip.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

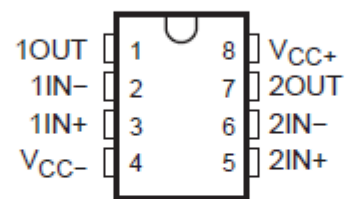
**TL071, TL071A, TL071B, TL072**  
**TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080J – SEPTEMBER 1978 – REVISED MARCH 2005

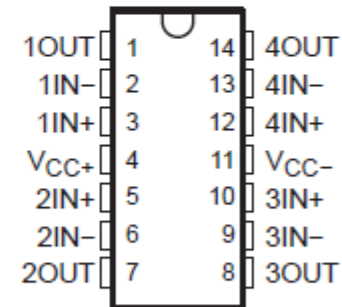
TL071, TL071A, TL071B  
 D, P, OR PS PACKAGE  
 (TOP VIEW)



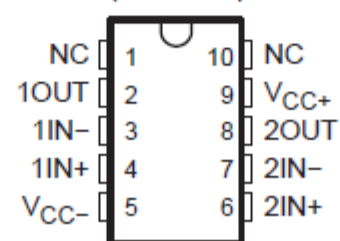
TL072, TL072A, TL072B  
 D, JG, P, PS, OR PW PACKAGE  
 (TOP VIEW)



TL074A, TL074B  
 D, J, N, NS, OR PW PACKAGE  
 TL074... D, J, N, NS, PW,  
 OR W PACKAGE  
 (TOP VIEW)



TL072  
 U PACKAGE  
 (TOP VIEW)



Electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	T <sub>A</sub> ‡	TL071C TL072C TL074C			TL071AC TL072AC TL074AC			TL071BC TL072BC TL074BC			TL071H TL072H TL074H			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C		3	10		3	6		2	3		3	6	mV
		Full range			13		7.5			5			8		
V <sub>IO</sub>	Temperature coefficient of input offset voltage V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	Full range		18		18		18		18		18		18	µV/°C
I <sub>IO</sub>	Input offset current V <sub>O</sub> = 0	25°C		5	100		5	100		5	100		5	100	pA
		Full range			10		2		2		2		2		
I <sub>B</sub>	Input bias current§ V <sub>O</sub> = 0	25°C		65	200		65	200		65	200		65	200	pA
		Full range			7		7		7		20		20		
V <sub>ICR</sub>	Common-mode input voltage range	25°C	±11	-12 to 15		±11	-12 to 15		±11	-12 to 15		±11	-12 to 15	V	
V <sub>OM</sub>	Maximum peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	±12	±13.5		±12	±13.5		±12	±13.5		±12	±13.5	V
		R <sub>L</sub> ≥ 10 kΩ	Full range		±12		±12		±12		±12		±12		
		R <sub>L</sub> ≥ 2 kΩ	Full range		±10		±10		±10		±10		±10		
V <sub>OD</sub>	Large-signal differential voltage amplification V <sub>O</sub> = ±10 V, R <sub>L</sub> ≥ 2 kΩ	25°C		25	200		50	200		50	200		50	200	V/mV
		Full range		15		25		25		25		25			
f <sub>u</sub>	Unity-gain bandwidth	25°C		3		3		3		3		3		MHz	
R <sub>in</sub>	Input resistance	25°C		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		Ω	
CMRR	Common-mode rejection ratio V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C		70	100		75	100		75	100		75	100	dB
PSRR	Supply-voltage rejection ratio V <sub>CC</sub> = ±9 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C		70	100		80	100		80	100		80	100	dB
I <sub>CC</sub>	Supply current (each amplifier) V <sub>O</sub> = 0, No load	25°C		1.4	2.5		1.4	2.5		1.4	2.5		1.4	2.5	mA
CMV <sub>O2</sub>	Crosstalk attenuation A <sub>VD</sub> = 100	25°C		120		120		120		120		120		dB	

† Characteristics are measured under open-loop conditions with zero common-mode voltage, unless otherwise specified.  
 ‡ Full range is: T<sub>A</sub> = 0°C to 70°C for TL07\_C, TL07\_AC, TL07\_BC and is T<sub>A</sub> = -40°C to 85°C for TL07\_J.  
 § Input bias currents of an FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive, as shown in Figure 4. Pulse techniques must be used that maintain the junction temperature as close to the ambient temperature as possible.

TL071, TL071A, TL071B, TL072  
 TL072A, TL072B, TL074, TL074A, TL074B  
 LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS  
 SLOS080U - SEPTEMBER 1978 - REVISED MARCH 2005

PARAMETER MEASUREMENT INFORMATION

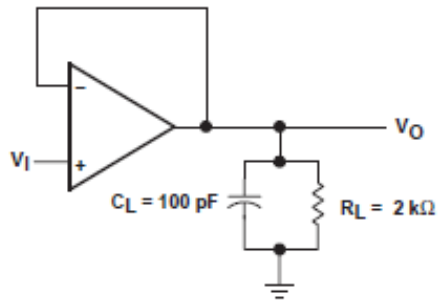


Figure 1. Unity-Gain Amplifier

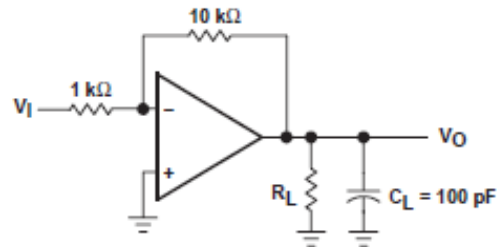


Figure 2. Gain-of-10 Inverting Amplifier

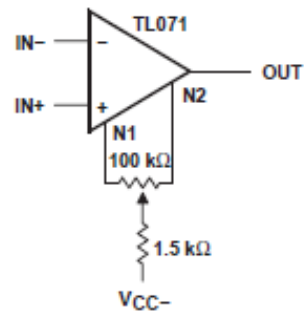


Figure 3. Input Offset-Voltage Null Circuit

TL082 WIDE BANDWITDH DUAL JFET INPUT





August 2000

## TL082

# Wide Bandwidth Dual JFET Input Operational Amplifier

### General Description

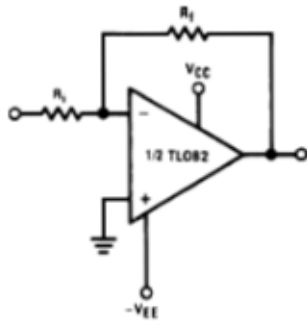
These devices are low cost, high speed, dual JFET input operational amplifiers with an internally trimmed input offset voltage (BI-FET II™ technology). They require low supply current yet maintain a large gain bandwidth product and fast slew rate. In addition, well matched high voltage JFET input devices provide very low input bias and offset currents. The TL082 is pin compatible with the standard LM1558 allowing designers to immediately upgrade the overall performance of existing LM1558 and most LM358 designs.

These amplifiers may be used in applications such as high speed integrators, fast D/A converters, sample and hold circuits and many other circuits requiring low input offset voltage, low input bias current, high input impedance, high slew rate and wide bandwidth. The devices also exhibit low noise and offset voltage drift.

### Features

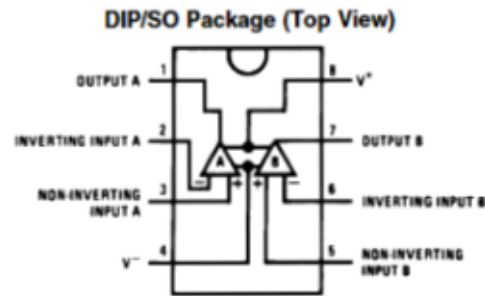
■ Internally trimmed offset voltage:	15 mV
■ Low input bias current:	50 pA
■ Low input noise voltage:	16nV/√Hz
■ Low input noise current:	0.01 pA/√Hz
■ Wide gain bandwidth:	4 MHz
■ High slew rate:	13 V/μs
■ Low supply current:	3.6 mA
■ High input impedance:	10 <sup>12</sup> Ω
■ Low total harmonic distortion:	≤0.02%
■ Low 1/f noise corner:	50 Hz
■ Fast settling time to 0.01%:	2 μs

**Typical Connection**



00835701

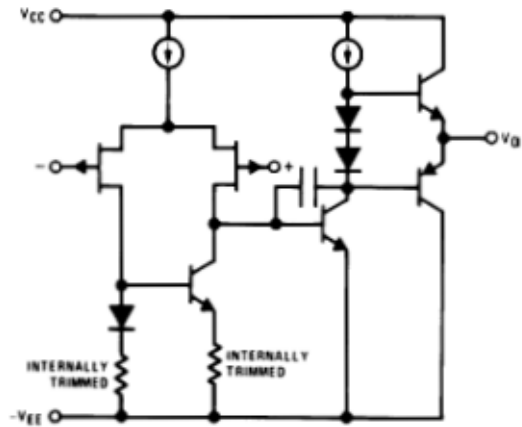
**Connection Diagram**



00835703

Order Number TL082CM or TL082CP  
See NS Package Number M08A or N08E

**Simplified Schematic**



00835702

BI-FET II™ is a trademark of National Semiconductor Corp.

**Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage	±18V
Power Dissipation	(Note 2)
Operating Temperature Range	0°C to +70°C
T <sub>J(MAX)</sub>	150°C
Differential Input Voltage	±30V

Input Voltage Range (Note 3)	±15V
Output Short Circuit Duration	Continuous
Storage Temperature Range	-65°C to +150°C
Lead Temp. (Soldering, 10 seconds)	260°C
ESD rating to be determined.	

**Note 1:** "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits.

**DC Electrical Characteristics** (Note 5)

Symbol	Parameter	Conditions	TL082C			Units
			Min	Typ	Max	
V <sub>OS</sub>	Input Offset Voltage	R <sub>S</sub> = 10 kΩ, T <sub>A</sub> = 25°C		5	15	mV
		Over Temperature			20	mV
ΔV <sub>OS</sub> /ΔT	Average TC of Input Offset Voltage	R <sub>S</sub> = 10 kΩ		10		μV/°C
I <sub>OS</sub>	Input Offset Current	T <sub>J</sub> = 25°C, (Notes 5, 6)		25	200	pA
		T <sub>J</sub> ≤ 70°C			4	nA
I <sub>B</sub>	Input Bias Current	T <sub>J</sub> = 25°C, (Notes 5, 6)		50	400	pA
		T <sub>J</sub> ≤ 70°C			8	nA
R <sub>IN</sub>	Input Resistance	T <sub>J</sub> = 25°C		10 <sup>12</sup>		Ω
A <sub>VOL</sub>	Large Signal Voltage Gain	V <sub>S</sub> = ±15V, T <sub>A</sub> = 25°C	25	100		V/mV
		V <sub>O</sub> = ±10V, R <sub>L</sub> = 2 kΩ Over Temperature	15			V/mV
V <sub>O</sub>	Output Voltage Swing	V <sub>S</sub> = ±15V, R <sub>L</sub> = 10 kΩ	±12	±13.5		V
V <sub>CM</sub>	Input Common-Mode Voltage Range	V <sub>S</sub> = ±15V	±11	+15		V
				-12		V
CMRR	Common-Mode Rejection Ratio	R <sub>S</sub> ≤ 10 kΩ	70	100		dB
PSRR	Supply Voltage Rejection Ratio	(Note 7)	70	100		dB
I <sub>S</sub>	Supply Current			3.6	5.6	mA

**AC Electrical Characteristics** (Note 5)

Symbol	Parameter	Conditions	TL082C			Units
			Min	Typ	Max	
	Amplifier to Amplifier Coupling	$T_A = 25^\circ\text{C}$ , $f = 1\text{Hz}-20\text{kHz}$ (Input Referred)		-120		dB
SR	Slew Rate	$V_S = \pm 15\text{V}$ , $T_A = 25^\circ\text{C}$	8	13		V/ $\mu\text{s}$
GBW	Gain Bandwidth Product	$V_S = \pm 15\text{V}$ , $T_A = 25^\circ\text{C}$		4		MHz
$e_n$	Equivalent Input Noise Voltage	$T_A = 25^\circ\text{C}$ , $R_S = 100\Omega$ , $f = 1000\text{Hz}$		25		nV/ $\sqrt{\text{Hz}}$
$i_n$	Equivalent Input Noise Current	$T_J = 25^\circ\text{C}$ , $f = 1000\text{Hz}$		0.01		pA/ $\sqrt{\text{Hz}}$
THD	Total Harmonic Distortion	$A_V = +10$ , $R_L = 10\text{k}$ , $V_O = 20\text{V}_p - p$ , $\text{BW} = 20\text{Hz}-20\text{kHz}$		<0.02		%

**Note 2:** For operating at elevated temperature, the device must be derated based on a thermal resistance of 115°C/W junction to ambient for the N package.

**Note 3:** Unless otherwise specified the absolute maximum negative input voltage is equal to the negative power supply voltage.

**Note 4:** The power dissipation limit, however, cannot be exceeded.

**Note 5:** These specifications apply for  $V_S = \pm 15\text{V}$  and  $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$ .  $V_{OS}$ ,  $I_B$  and  $I_{OS}$  are measured at  $V_{CM} = 0$ .

**Note 6:** The input bias currents are junction leakage currents which approximately double for every 10°C increase in the junction temperature,  $T_J$ . Due to the limited production test time, the input bias currents measured are correlated to junction temperature. In normal operation the junction temperature rises above the ambient temperature as a result of internal power dissipation,  $P_D$ .  $T_J = T_A + \theta_{JA} P_D$  where  $\theta_{JA}$  is the thermal resistance from junction to ambient. Use of a heat sink is recommended if input bias current is to be kept to a minimum.

**Note 7:** Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously in accordance with common practice.  $V_S = \pm 6\text{V}$  to  $\pm 15\text{V}$ .

Philips Semiconductors

## Low power dual op

### DESCRIPTION

The 532/358/LM2904 consists of two internally frequency-compensated operational amplifiers. It can operate from a single power supply or from dual supplies. Operation from dual power supplies is a power supply current drain is independent of power supply voltage.

### UNIQUE FEATURES

In the linear mode the input common-mode voltage can swing to ground and the output voltage can also swing to ground. The output voltage can also swing to  $V_{CC}$  when operated from only a single power supply. The crossover frequency is temperature-compensated and the output current is also temperature-compensated.

### FEATURES

- Internally frequency-compensated for unity gain
- Large DC voltage gain: 100 dB
- Wide bandwidth (unity gain): 1 MHz (temperature-compensated)
- Wide power supply range single supply:  $3 V_{DC}$  to  $30 V_{DC}$ , or dual supplies:  $\pm 1.5 V_{DC}$  to  $\pm 15 V_{DC}$
- Very low supply current drain ( $400 \mu A$ )—essentially independent of supply voltage ( $1 mW/op\ amp\ at\ +5 V_{DC}$ )
- Low input biasing current:  $45 nA_{DC}$  temperature-compensated
- Low input offset voltage:  $2 mV_{DC}$ , and offset current:  $5 nA_{DC}$
- Large output voltage:  $0 V_{DC}$  to  $V_{+} - 1.5 V_{DC}$  swing

EQUIVALENT CIRCUIT

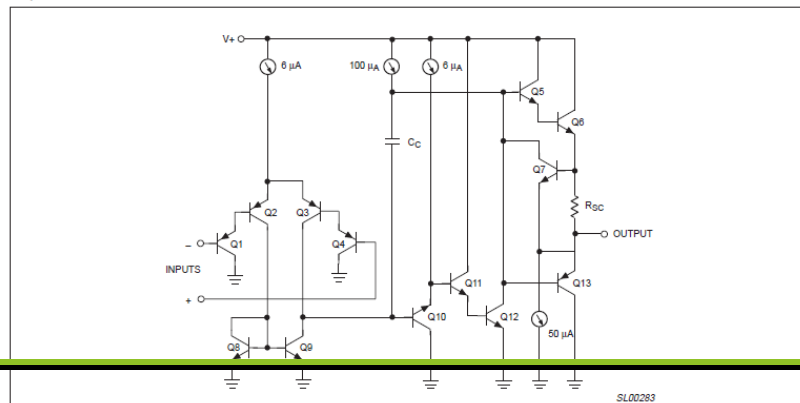


Figure 2. Equivalent circuit.

## ABSOLUTE MAXIMUM RATINGS

SYMBOL	PARAMETER	RATING	UNIT
$V_S$	Supply voltage, $V_+$	32 or $\pm 16$	$V_{DC}$
	Differential input voltage	32	$V_{DC}$
$V_{IN}$	Input voltage	-0.3 to +32	$V_{DC}$
$P_D$	Maximum power dissipation $T_{amb} = 25\text{ }^\circ\text{C}$ (Still air) <sup>1</sup>		
	N package	1160	mW
	D package	780	mW
	DP package	714	mW
	Output short-circuit to GND <sup>2</sup> $V_+ < 15\text{ }V_{DC}$ and $T_{amb} = 25\text{ }^\circ\text{C}$	Continuous	
$T_{amb}$	Operating ambient temperature range		
	NE532/LM358/LM358A	0 to +70	$^\circ\text{C}$
	LM258	-25 to +85	$^\circ\text{C}$
	LM2904	-40 to +125	$^\circ\text{C}$
	SA532	-40 to +85	$^\circ\text{C}$
SE532	-55 to +125	$^\circ\text{C}$	
$T_{stg}$	Storage temperature range	-65 to +150	$^\circ\text{C}$
$T_{slid}$	Lead soldering temperature (10 sec max)	230	$^\circ\text{C}$

Philips Semiconductors

Product data

Low power dual operational amplifiers

NE/SA/SE532/  
LM258/358/A/2904

DC ELECTRICAL CHARACTERISTICS

T<sub>amb</sub> = 25 °C; V<sub>+</sub> = +5 V, unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	SE532, LM258			NE/SA532/ LM358/LM2904			UNIT
			Min	Typ	Max	Min	Typ	Max	
V <sub>OS</sub>	Offset voltage <sup>1</sup>	R <sub>O</sub> = 0 Ω		±2	±5		±2	±7	mV
		R <sub>O</sub> = 0 Ω; over temp.			±7			±9	mV
V <sub>OS</sub>	Drift	R <sub>O</sub> = 0 Ω; over temp.		7			7		μV/°C
I <sub>OS</sub>	Offset current	I <sub>IN(+)</sub> - I <sub>IN(-)</sub>		±3	±30		±5	±50	nA
		Over temp.			±100			±150	nA
I <sub>OS</sub>	Drift	Over temp.		10			10		pA/°C
I <sub>BIAS</sub>	Input current <sup>2</sup>	I <sub>IN(+)</sub> or I <sub>IN(-)</sub>		45	150		45	250	nA
		I <sub>IN(+)</sub> or I <sub>IN(-)</sub> ; Over temp.		40	300		40	500	nA
I <sub>B</sub>	Drift	Over temp.		50			50		pA/°C
V <sub>CM</sub>	Common-mode voltage range <sup>3</sup>	V <sub>+</sub> = 30 V	0		V <sub>+</sub> -1.5	0		V <sub>+</sub> -1.5	V
		V <sub>+</sub> = 30 V; Over temp.	0		V <sub>+</sub> -2.0	0		V <sub>+</sub> -2.0	V
CMRR	Common-mode rejection ratio	V <sub>+</sub> = 30 V	70	85		65	70		dB
V <sub>OH</sub>	Output voltage swing	R <sub>L</sub> ≥ 2 kΩ; V <sub>+</sub> = 30 V; over temp.	26			26			V
		R <sub>L</sub> ≥ 10 kΩ; V <sub>+</sub> = 30 V; over temp.	27	28		27	28		V
V <sub>OL</sub>	Output voltage swing	R <sub>L</sub> ≥ 10 kΩ; over temp.		5	20		5	20	mV
I <sub>CC</sub>	Supply current	R <sub>L</sub> = ∞; V <sub>+</sub> = 30 V		0.5	1.0		0.5	1.0	mA
		R <sub>L</sub> = ∞ on all amplifiers; V <sub>+</sub> = 30 V; over temp.		0.6	1.2		0.6	1.2	mA
A <sub>VOL</sub>	Large-signal voltage gain	R <sub>L</sub> ≥ 2 kΩ; V <sub>OUT</sub> ±10 V	50	100		25	100		V/mV
		V <sub>+</sub> = 15 V (for large V <sub>O</sub> swing); over temp.	25			15			V/mV
PSRR	Supply voltage rejection ratio	R <sub>O</sub> = 0 Ω	65	100		65	100		dB
	Amplifier-to-amplifier coupling <sup>4</sup>	f = 1 kHz to 20 kHz (input referred)		-120			-120		dB
I <sub>OUT</sub>	Output current (Source)	V <sub>IN+</sub> = +1 V <sub>DC</sub> ; V <sub>IN-</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub>	20	40		20	40		mA
		V <sub>IN+</sub> = +1 V <sub>DC</sub> ; V <sub>IN-</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub> ; over temp.	10	20		10	20		mA
		V <sub>IN-</sub> = +1 V <sub>DC</sub> ; V <sub>IN+</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub>	10	20		10	20		mA
	Output current (Sink)	V <sub>IN-</sub> = +1 V <sub>DC</sub> ; V <sub>IN+</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub> ; over temp.	5	8		5	8		mA
		V <sub>IN+</sub> = 0 V; V <sub>IN-</sub> = +1 V <sub>DC</sub> ; V <sub>O</sub> = 200 mV	12	50		12	50		μA
I <sub>SC</sub>	Short circuit current <sup>5</sup>		40	60		40	60		mA
	Differential input voltage <sup>6</sup>								V
GBW	Unity gain bandwidth	T <sub>amb</sub> = 25 °C		1			1		MHz
SR	Slew rate	T <sub>amb</sub> = 25 °C		0.3			0.3		V/μs
V <sub>NOISE</sub>	Input noise voltage	T <sub>amb</sub> = 25 °C; f = 1 kHz		40			40		nV/√Hz

Philips Semiconductors

Product data

Low power dual operational amplifiers

NE/SA/SE532/  
LM258/358/A/2904

DC ELECTRICAL CHARACTERISTICS (continued)

T<sub>amb</sub> = 25 °C; V<sub>+</sub> = +5 V; unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	LM358A			UNIT
			Min	Typ	Max	
V <sub>OS</sub>	Offset voltage <sup>1</sup>	R <sub>O</sub> = 0 Ω		±2	±3	mV
		R <sub>O</sub> = 0 Ω; over temp.			±5	mV
V <sub>OS</sub>	Drift	R <sub>O</sub> = 0 Ω; over temp.		7	20	μV/°C
I <sub>OS</sub>	Offset current	I <sub>IN(+)</sub> - I <sub>IN(-)</sub>		5	±30	nA
		Over temp.			±75	nA
I <sub>OS</sub>	Drift	Over temp.		10	300	pA/°C
I <sub>BIAS</sub>	Input current <sup>2</sup>	I <sub>IN(+)</sub> or I <sub>IN(-)</sub>		45	100	nA
		I <sub>IN(+)</sub> or I <sub>IN(-)</sub> ; Over temp.		40	200	nA
I <sub>B</sub>	Drift	Over temp.		50		pA/°C
V <sub>CM</sub>	Common-mode voltage range <sup>3</sup>	V <sub>+</sub> = 30 V	0		V <sub>+</sub> -1.5	V
		V <sub>+</sub> = 30 V; Over temp.	0		V <sub>+</sub> -2.0	V
CMRR	Common-mode rejection ratio	V <sub>+</sub> = 30 V	65	85		dB
V <sub>OH</sub>	Output voltage swing	R <sub>L</sub> ≥ 2 kΩ; V <sub>+</sub> = 30 V; over temp.	26			V
		R <sub>L</sub> ≥ 10 kΩ; V <sub>+</sub> = 30 V; over temp.	27	28		V
V <sub>OL</sub>	Output voltage swing	R <sub>L</sub> ≥ 10 kΩ; over temp.		5	20	mV
I <sub>CC</sub>	Supply current	R <sub>L</sub> = ∞; V <sub>+</sub> = 30 V		0.5	1.0	mA
		R <sub>L</sub> = ∞ on all amplifiers; V <sub>+</sub> = 30 V; over temp.		0.6	1.2	mA
A <sub>VOL</sub>	Large-signal voltage gain	R <sub>L</sub> ≥ 2 kΩ; V <sub>OUT</sub> ±10 V	25	100		V/mV
		V <sub>+</sub> = 15 V (for large V <sub>O</sub> swing); over temp.	15			V/mV
PSRR	Supply voltage rejection ratio	R <sub>O</sub> = 0 Ω	65	100		dB
	Amplifier-to-amplifier coupling <sup>4</sup>	f = 1 kHz to 20 kHz (input referred)		-120		dB
I <sub>OUT</sub>	Output current (Source)	V <sub>IN+</sub> = +1 V <sub>DC</sub> ; V <sub>IN-</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub>	20	40		mA
		V <sub>IN+</sub> = +1 V <sub>DC</sub> ; V <sub>IN-</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub> ; over temp.	10	20		mA
		V <sub>IN-</sub> = +1 V <sub>DC</sub> ; V <sub>IN+</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub>	10	20		mA
	Output current (Sink)	V <sub>IN-</sub> = +1 V <sub>DC</sub> ; V <sub>IN+</sub> = 0 V <sub>DC</sub> ; V <sub>+</sub> = 15 V <sub>DC</sub> ; over temp.	5	8		mA
		V <sub>IN+</sub> = 0 V; V <sub>IN-</sub> = +1 V <sub>DC</sub> ; V <sub>O</sub> = 200 mV	12	50		μA
I <sub>SC</sub>	Short circuit current <sup>5</sup>		40	60		mA
	Differential input voltage <sup>6</sup>					V
GBW	Unity gain bandwidth	T <sub>amb</sub> = 25 °C		1		MHz
SR	Slew rate	T <sub>amb</sub> = 25 °C		0.3		V/μs
V <sub>NOISE</sub>	Input noise voltage	T <sub>amb</sub> = 25 °C; f = 1 kHz		40		nV/√Hz

LM324 QUADRUPLO



LM224K, LM224KA, LM324, LM324A, LM324K, LM324KA, LM2902  
LM124, LM124A, LM224, LM224A, LM2902V, LM2902K, LM2902KV, LM2902KAV

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LMx24, LMx24x, LMx24xx, LM2902, LM2902x, LM2902xx, LM2902xxx Quadruple  
Operational Amplifiers

1 Features

- 2-kV ESD Protection for:
  - LM224K, LM224KA
  - LM324K, LM324KA
  - LM2902K, LM2902KV, LM2902KAV
- Wide Supply Ranges
  - Single Supply: 3 V to 32 V (26 V for LM2902)
  - Dual Supplies:  $\pm 1.5$  V to  $\pm 16$  V ( $\pm 13$  V for LM2902)
- Low Supply-Current Drain Independent of Supply Voltage: 0.8 mA Typical
- Common-Mode Input Voltage Range Includes Ground, Allowing Direct Sensing Near Ground
- Low Input Bias and Offset Parameters
  - Input Offset Voltage: 3 mV Typical  
A Versions: 2 mV Typical
  - Input Offset Current: 2 nA Typical
  - Input Bias Current: 20 nA Typical  
A Versions: 15 nA Typical
- Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage: 32 V (26 V for LM2902)
- Open-Loop Differential Voltage Amplification: 100 V/mV Typical
- Internal Frequency Compensation
- On Products Compliant to MIL-PRF-38535, All Parameters are Tested Unless Otherwise Noted. On All Other Products, Production Processing Does Not Necessarily Include Testing of All Parameters.

2 Applications

- Blu-ray Players and Home Theaters
- Chemical and Gas Sensors
- DVD Recorders and Players
- Digital Multimeter: Bench and Systems
- Digital Multimeter: Handhelds
- Field Transmitter: Temperature Sensors
- Motor Control: AC Induction, Brushed DC, Brushless DC, High-Voltage, Low-Voltage, Permanent Magnet, and Stepper Motor
- Oscilloscopes
- TV: LCD and Digital
- Temperature Sensors or Controllers Using Modbus
- Weigh Scales

3 Description

These devices consist of four independent high-gain frequency-compensated operational amplifiers that are designed specifically to operate from a single supply or split supply over a wide range of voltages.

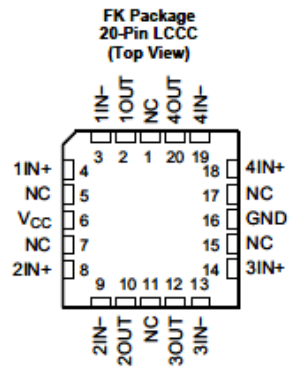
Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LMx24, LMx24x, LMx24xx, LM2902, LM2902x, LM2902xx, LM2902xxx	SOIC (14)	8.65 mm × 3.91 mm
	CDIP (14)	19.56 mm × 6.87 mm
	PDIP (14)	19.30 mm × 6.35 mm
	CFP (14)	9.21 mm × 5.97 mm
	TSSOP (14)	5.00 mm × 4.40 mm
	SO (14)	9.20 mm × 5.30 mm
LM124, LM124A	SSOP (14)	6.20 mm × 5.30 mm
	LCCC (20)	8.90 mm × 8.90 mm

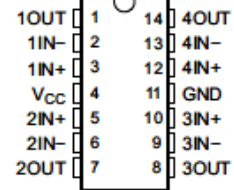
(1) For all available packages, see the orderable addendum at the end of the data sheet.



5 Pin Configuration and Functions



D, DB, J, N, NS, PW, W  
14-Pin SOIC, SSOP, CDIP, PDIP, SO, TSSOP, CFP  
(Top View)



Pin Functions

NAME	PIN		I/O	DESCRIPTION
	LCCC NO.	SOIC, SSOP, CDIP, PDIP, SO, TSSOP, CFP NO.		
1IN-	3	2	I	Negative input
1IN+	4	3	I	Positive input
1OUT	2	1	O	Output
2IN-	9	6	I	Negative input
2IN+	8	5	I	Positive input
2OUT	10	7	O	Output
3IN-	13	9	I	Negative input
3IN+	14	10	I	Positive input
3OUT	12	8	O	Output
4IN-	19	13	I	Negative input
4IN+	18	12	I	Positive input
4OUT	20	14	O	Output
GND	16	11	—	Ground
NC	1	—	—	Do not connect
	5			
	7			
	11			
	15			
V <sub>CC</sub>	6	4	—	Power supply

LM224K, LM224KA, LM324, LM324A, LM324K, LM324KA, LM2902  
 LM124, LM124A, LM224, LM224A, LM2902V, LM2902K, LM2902KV, LM2902KAV



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## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	LM2902		LMx24, LMx24x, LMx24xx, LM2902x, LM2902xx, LM2902xxx		UNIT
	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC}$ <sup>(2)</sup>	±13	28	±16	32	V
Differential input voltage, $V_{ID}$ <sup>(3)</sup>		±26		±32	V
Input voltage, $V_I$ (either input)	-0.3	28	-0.3	to 32	V
Duration of output short circuit (one amplifier) to ground at (or below) $T_A = 25^\circ\text{C}$ , $V_{CC} \leq 15\text{ V}$ <sup>(4)</sup>	Unlimited		Unlimited		
Operating virtual junction temperature, $T_J$	150		150		°C
Case temperature for 60 seconds	FK package		260		°C
Lead temperature 1.6 mm (1/16 inch) from case for 60 seconds	J or W package		300		°C
Storage temperature, $T_{stg}$	-85	150	-85	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values (except differential voltages and  $V_{CC}$  specified for the measurement of  $I_{OQ}$ ) are with respect to the network GND.

(3) Differential voltages are at  $IN+$ , with respect to  $IN-$ .

(4) Short circuits from outputs to VCC can cause excessive heating and eventual destruction.

### 6.2 ESD Ratings

		VALUE	UNIT
<b>LM224K, LM224KA, LM324K, LM324KA, LM2902K, LM2902KV, LM2902KAV</b>			
$V_{(ESD)}$ Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101	±1000	
<b>LM124, LM124A, LM224, LM224A, LM324, LM324A, LM2902, LM2902V</b>			
$V_{(ESD)}$ Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±500	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101	±1000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	LM2902		LMx24, LMx24x, LMx24xx, LM2902x, LM2902xx, LM2902xxx		UNIT
	MIN	MAX	MIN	MAX	
$V_{CC}$ Supply voltage	3	28	3	30	V
$V_{CM}$ Common-mode voltage	0	$V_{CC} - 2$	0	$V_{CC} - 2$	V
$T_A$ Operating free air temperature	LM124		-55	125	°C
	LM2904	-40	125		
	LM324		0	70	
	LM224		-25	85	

LM224K, LM224KA, LM324, LM324A, LM324K, LM324KA, LM2902  
 LM124, LM124A, LM224, LM224A, LM2902V, LM2902K, LM2902KV, LM2902KAV



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**6 Specifications**

**6.1 Absolute Maximum Ratings**

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	LM2902		LMx24, LMx24x, LMx24xx, LM2902x, LM2902xx, LM2902xxx		UNIT
	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC}$ <sup>(2)</sup>	±13	26	±16	32	V
Differential input voltage, $V_{ID}$ <sup>(2)</sup>		±26		±32	V
Input voltage, $V_I$ (either input)	-0.3	26	-0.3	to 32	V
Duration of output short circuit (one amplifier) to ground at (or below) $T_A = 25^\circ\text{C}$ , $V_{CC} \leq 15\text{ V}$ <sup>(4)</sup>	Unlimited		Unlimited		
Operating virtual junction temperature, $T_J$	150		150		°C
Case temperature for 60 seconds	FK package		260		°C
Lead temperature 1.6 mm (1/16 inch) from case for 60 seconds	J or W package		300		°C
Storage temperature, $T_{STG}$	-65	150	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values (except differential voltages and  $V_{CC}$  specified for the measurement of  $I_{CC}$ ) are with respect to the network GND.
- (3) Differential voltages are at IN+, with respect to IN-.
- (4) Short circuits from outputs to VCC can cause excessive heating and eventual destruction.

**6.2 ESD Ratings**

		VALUE	UNIT
<b>LM224K, LM224KA, LM324K, LM324KA, LM2902K, LM2902KV, LM2902KAV</b>			
$V_{ESD}$ Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101	±1000	
<b>LM124, LM124A, LM224, LM224A, LM324, LM324A, LM2902, LM2902V</b>			
$V_{ESD}$ Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±500	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101	±1000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

**6.3 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)

	LM2902		LMx24, LMx24x, LMx24xx, LM2902x, LM2902xx, LM2902xxx		UNIT
	MIN	MAX	MIN	MAX	
$V_{CC}$ Supply voltage	3	26	3	30	V
$V_{CM}$ Common-mode voltage	0	$V_{CC} - 2$	0	$V_{CC} - 2$	V
$T_A$ Operating free air temperature	LM124		-65	125	°C
	LM2904	-40	125		
	LM324		0	70	
	LM224		-25	85	



LM224K, LM224KA, LM324, LM324A, LM324K, LM324KA, LM2902  
 LM124, LM124A, LM224, LM224A, LM2902V, LM2902K, LM2902KV, LM2902KAV

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6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>	LMx24, LM2902					LMx24			UNIT
	D (SOIC)	DB (SSOP)	N (PDIP)	NS (SO)	PW (TSSOP)	FK (LCCC)	J (CDIP)	W (CFP)	
	14 PINS	14 PINS	14 PINS	14 PINS	14 PINS	20 PINS	14 PINS	14 PINS	
R <sub>θJA</sub> <sup>(2)(3)</sup> Junction-to-ambient thermal resistance	86	86	80	76	113	—	—	—	°C/W
R <sub>θJC</sub> <sup>(4)</sup> Junction-to-case (top) thermal resistance	—	—	—	—	—	5.61	15.05	14.65	

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).
- (2) Short circuits from outputs to VCC can cause excessive heating and eventual destruction.
- (3) Maximum power dissipation is a function of T<sub>J(max)</sub>, R<sub>θJA</sub>, and T<sub>A</sub>. The maximum allowable power dissipation at any allowable ambient temperature is P<sub>D</sub> = (T<sub>J(max)</sub> - T<sub>A</sub>)/R<sub>θJA</sub>. Operating at the absolute maximum T<sub>J</sub> of 150°C can affect reliability.
- (4) Maximum power dissipation is a function of T<sub>J(max)</sub>, R<sub>θJA</sub>, and T<sub>C</sub>. The maximum allowable power dissipation at any allowable case temperature is P<sub>D</sub> = (T<sub>J(max)</sub> - T<sub>C</sub>)/R<sub>θJC</sub>. Operating at the absolute maximum T<sub>J</sub> of 150°C can affect reliability.

**6.5 Electrical Characteristics for LMx24 and LM324K**

at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS (1)	$T_A$ (2)	LM124, LM224			LM324, LM324K			UNIT		
			MIN	TYP (3)	MAX	MIN	TYP (3)	MAX			
$V_{IO}$ Input offset voltage	$V_{CC} = 5\text{ V to MAX}$ , $V_{IC} = V_{ICRmin}$ , $V_O = 1.4\text{ V}$	25°C		3	5		3	7	mV		
		Full range			7			9			
$I_{IO}$ Input offset current	$V_O = 1.4\text{ V}$	25°C		2	30		2	50	nA		
		Full range			100			150			
$I_B$ Input bias current	$V_O = 1.4\text{ V}$	25°C		-20	-150		-20	-250	nA		
		Full range			-300			-500			
$V_{ICR}$ Common-mode input voltage range	$V_{CC} = 5\text{ V to MAX}$	25°C		0 to $V_{CC} - 1.5$		0 to $V_{CC} - 1.5$			V		
		Full range		0 to $V_{CC} - 2$		0 to $V_{CC} - 2$					
$V_{OH}$ High-level output voltage	$R_L = 2\text{ k}\Omega$ $R_L = 10\text{ k}\Omega$ $V_{CC} = \text{MAX}$	25°C		$V_{CC} - 1.5$		$V_{CC} - 1.5$			V		
		Full range									
			$R_L = 2\text{ k}\Omega$	Full range	26		26				
			$R_L \geq 10\text{ k}\Omega$	Full range	27	28	27	28			
$V_{OL}$ Low-level output voltage	$R_L \leq 10\text{ k}\Omega$	Full range		5	20		5	20	mV		
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$ , $V_O = 1\text{ V to }11\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	25°C		50	100		25	100	V/mV		
		Full range		25			15				
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C		70	80		65	80	dB		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC}/\Delta V_{IO}$ )		25°C		65	100		65	100	dB		
$V_{O1}/V_{O2}$ Crosstalk attenuation	$f = 1\text{ kHz to }20\text{ kHz}$	25°C		120			120		dB		
$I_O$ Output current	$V_{CC} = 15\text{ V}$ , $V_{IO} = 1\text{ V}$ , $V_O = 0$	Source	25°C		-20	-30	-60	-20	-30	-60	mA
			Full range			-10			-10		
	$V_{CC} = 15\text{ V}$ , $V_{IO} = -1\text{ V}$ , $V_O = 15\text{ V}$	Sink	25°C		10	20		10	20		
			Full range			5			5		
	$V_{IO} = -1\text{ V}$ , $V_O = 200\text{ mV}$	25°C		12	30		12	30	$\mu\text{A}$		
$I_{OS}$ Short-circuit output current	$V_{CC}$ at 5 V, $V_O = 0$ , GND at -5 V	25°C		$\pm 40$	$\pm 60$		$\pm 40$	$\pm 60$	mA		
$I_{CC}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , no load	Full range		0.7	1.2		0.7	1.2	mA		
	$V_{CC} = \text{MAX}$ , $V_O = 0.5 V_{CC}$ , no load	Full range		1.4	3		1.4	3			

- (1) All characteristics are measured under open-loop conditions, with zero common-mode input voltage, unless otherwise specified. MAX  $V_{CC}$  for testing purposes is 26 V for LM2902 and 30 V for the others.
- (2) Full range is -55°C to 125°C for LM124, -25°C to 85°C for LM224, and 0°C to 70°C for LM324.
- (3) All typical values are at  $T_A = 25^\circ\text{C}$





LM741QML

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LM741QML Operational Amplifier

Check for Samples: [LM741QML](#)

FEATURES

The amplifier offers many features which make their application nearly foolproof: overload protection on the input and output, no latch-up when the common mode range is exceeded, as well as freedom from oscillations

DESCRIPTION

The LM741 is a general purpose operational amplifier which features improved performance over industry standards such as the LM709. They are direct, plug-in replacements for the 709C, LM201, MC1439 and 748 in most applications.

Connection Diagrams

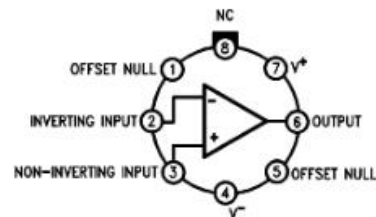


Figure 1. Metal Can Package  
See Package Number LMC0008C

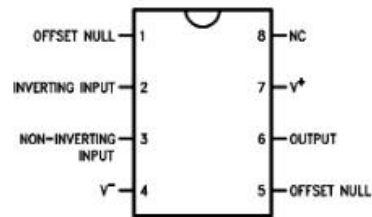


Figure 2. Dual-In-Line Package  
See Package Number NAB0008A

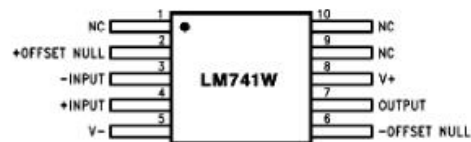


Figure 3. Ceramic Flatpak and SOIC Package  
See Package Number NAD0010A & NAC0010A

COMPARADORES



LM393 DUAS COMPARADOR COLETOR ABERTO IOUT 0,02A



Order this document by LM393/D

**LM393, LM393A,  
LM293, LM2903,  
LM2903V**

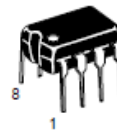
**Low Offset Voltage  
Dual Comparators**

The LM393 series are dual independent precision voltage comparators capable of single or split supply operation. These devices are designed to permit a common mode range-to-ground level with single supply operation. Input offset voltage specifications as low as 2.0 mV make this device an excellent selection for many applications in consumer automotive, and industrial electronics.

- Wide Single-Supply Range: 2.0 Vdc to 36 Vdc
- Split-Supply Range:  $\pm 1.0$  Vdc to  $\pm 18$  Vdc
- Very Low Current Drain Independent of Supply Voltage: 0.4 mA
- Low Input Bias Current: 25 nA
- Low Input Offset Current: 5.0 nA
- Low Input Offset Voltage: 2.0 mV (max) LM393A  
5.0 mV (max) LM293/393
- Input Common Mode Range to Ground Level
- Differential Input Voltage Range Equal to Power Supply Voltage
- Output Voltage Compatible with DTL, ECL, TTL, MOS, and CMOS Logic Levels
- ESD Clamps on the Inputs Increase the Ruggedness of the Device without Affecting Performance

**SINGLE SUPPLY, LOW POWER  
DUAL COMPARATORS**

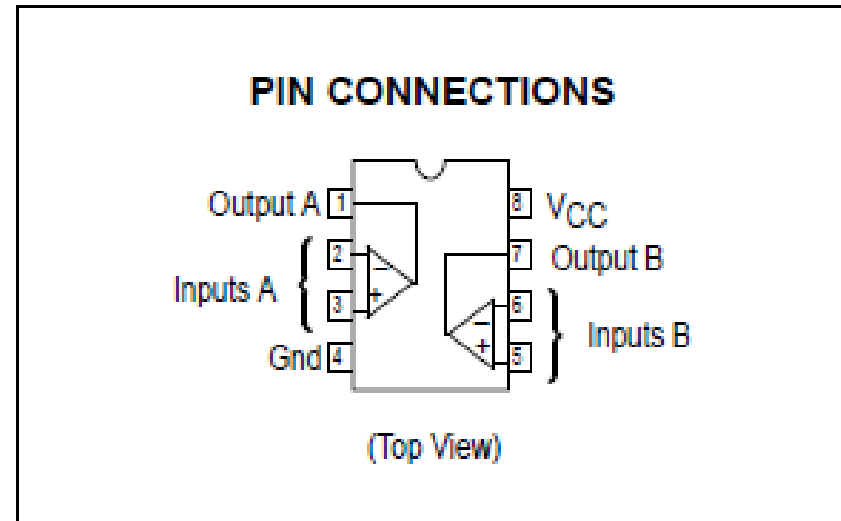
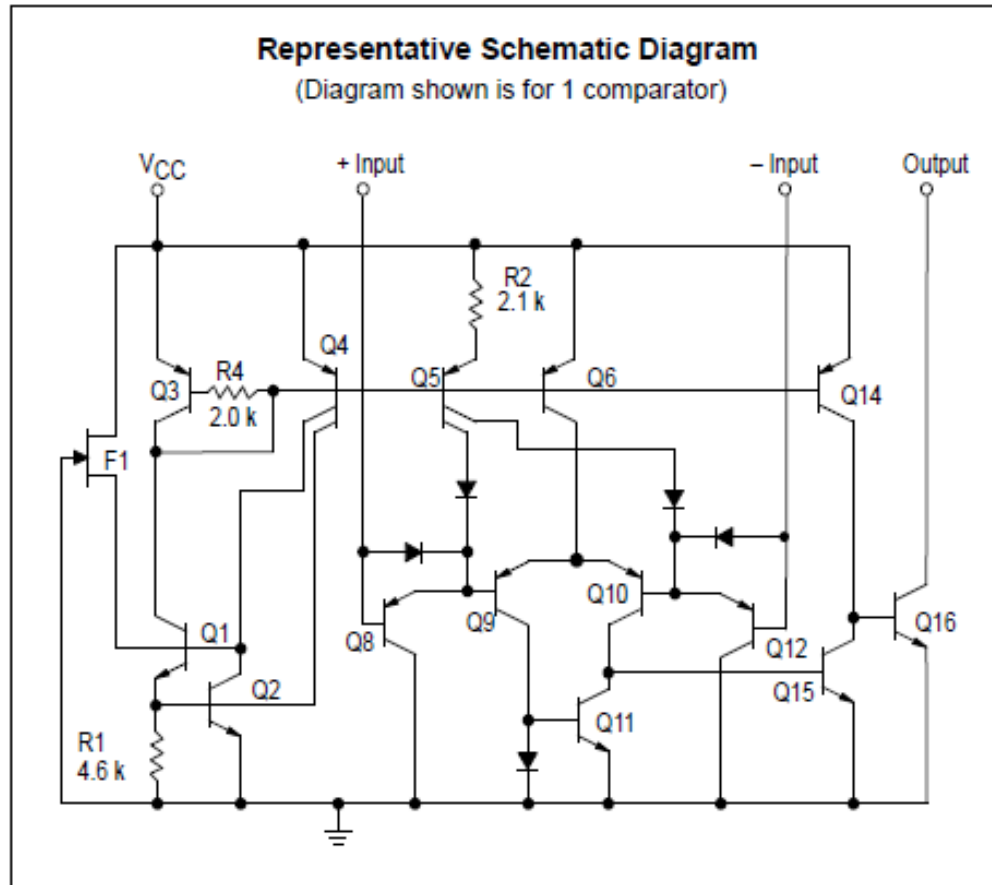
**SEMICONDUCTOR  
TECHNICAL DATA**



**N SUFFIX  
PLASTIC PACKAGE  
CASE 626**



**D SUFFIX  
PLASTIC PACKAGE  
CASE 751  
(SO-8)**



**LM393, LM393A, LM293, LM2903, LM2903V****MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Power Supply Voltage	$V_{CC}$	+36 or $\pm 18$	Vdc
Input Differential Voltage Range	$V_{IDR}$	36	Vdc
Input Common Mode Voltage Range	$V_{ICR}$	-0.3 to +36	Vdc
Output Short Circuit-to-Ground Output Sink Current (Note 1)	$I_{SC}$ $I_{Sink}$	Continuous 20	mA
Power Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$ $1/R_{\theta JA}$	570 5.7	mW mW/ $^\circ\text{C}$
Operating Ambient Temperature Range LM293 LM393, 393A LM2903 LM2903V	$T_A$	-25 to +85 0 to +70 -40 to +105 -40 to +125	$^\circ\text{C}$
Maximum Operating Junction Temperature LM393, 393A, 2903, LM2903V LM293	$T_{J(max)}$	125 150	$^\circ\text{C}$
Storage Temperature Range	$T_{stg}$	-65 to +150	$^\circ\text{C}$

**ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5.0$  Vdc,  $T_{low} \leq T_A \leq T_{high}$ , \* unless otherwise noted.)

Characteristic	Symbol	LM393A			Unit
		Min	Typ	Max	
Input Offset Voltage (Note 2) $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$V_{IO}$	– –	$\pm 1.0$ –	$\pm 2.0$ 4.0	mV
Input Offset Current $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$I_{IO}$	– –	$\pm 50$ –	$\pm 50$ $\pm 150$	nA
Input Bias Current (Note 3) $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$I_{IB}$	– –	25 –	250 400	nA
Input Common Mode Voltage Range (Note 4) $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$V_{ICR}$	0 0	– –	$V_{CC} - 1.5$ $V_{CC} - 2.0$	V
Voltage Gain $R_L \geq 15$ k $\Omega$ , $V_{CC} = 15$ Vdc, $T_A = 25^\circ\text{C}$	$A_{VOL}$	50	200	–	V/mV
Large Signal Response Time $V_{in} = \text{TTL Logic Swing}$ , $V_{ref} = 1.4$ Vdc $V_{RL} = 5.0$ Vdc, $R_L = 5.1$ k $\Omega$ , $T_A = 25^\circ\text{C}$	–	–	300	–	ns
Response Time (Note 5) $V_{RL} = 5.0$ Vdc, $R_L = 5.1$ k $\Omega$ , $T_A = 25^\circ\text{C}$	$t_{TLH}$	–	1.3	–	$\mu\text{s}$
Input Differential Voltage (Note 6) All $V_{in} \geq \text{Gnd}$ or $V^-$ Supply (if used)	$V_{ID}$	–	–	$V_{CC}$	V
Output Sink Current $V_{in} \geq 1.0$ Vdc, $V_{in+} = 0$ Vdc, $V_O \leq 1.5$ Vdc, $T_A = 25^\circ\text{C}$	$I_{Sink}$	6.0	16	–	mA
Output Saturation Voltage $V_{in} \geq 1.0$ Vdc, $V_{in+} = 0$ Vdc, $I_{Sink} \leq 4.0$ mA, $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$V_{OL}$	– –	150 –	400 700	mV

\* $T_{low} = 0^\circ\text{C}$ ,  $T_{high} = +70^\circ\text{C}$  for LM393/393A

- NOTES:**
1. The maximum output current may be as high as 20 mA, independent of the magnitude of  $V_{CC}$ , output short circuits to  $V_{CC}$  can cause excessive heating and eventual destruction.
  2. At output switch point,  $V_O \approx 1.4$  Vdc,  $R_S = 0$   $\Omega$  with  $V_{CC}$  from 5.0 Vdc to 30 Vdc, and over the full input common mode range (0 V to  $V_{CC} = -1.5$  V).
  3. Due to the PNP transistor inputs, bias current will flow out of the inputs. This current is essentially constant, independent of the output state, therefore, no loading changes will exist on the input lines.
  4. Input common mode of either input should not be permitted to go more than 0.3 V negative of ground or minus supply. The upper limit of common mode range is  $V_{CC} - 1.5$  V.
  5. Response time is specified with a 100 mV step and 5.0 mV of overdrive. With larger magnitudes of overdrive faster response times are obtainable.
  6. The comparator will exhibit proper output state if one of the inputs becomes greater than  $V_{CC}$ , the other input must remain within the common mode range. The low input state must not be less than  $-0.3$  V of ground or minus supply.

**LM393, LM393A, LM293, LM2903, LM2903V****ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5.0 \text{ Vdc}$ ,  $T_{low} \leq T_A \leq T_{high}$ , \* unless otherwise noted.)

Characteristic	Symbol	LM393A			Unit
		Min	Typ	Max	
Output Leakage Current $V_{in-} = 0 \text{ V}$ , $V_{in+} \geq 1.0 \text{ Vdc}$ , $V_O = 5.0 \text{ Vdc}$ , $T_A = 25^\circ\text{C}$ $V_{in-} = 0 \text{ V}$ , $V_{in+} \geq 1.0 \text{ Vdc}$ , $V_O = 30 \text{ Vdc}$ , $T_{low} \leq T_A \leq T_{high}$	$I_{OL}$	–	0.1	–	$\mu\text{A}$
Supply Current $R_L = \infty$ Both Comparators, $T_A = 25^\circ\text{C}$ $R_L = \infty$ Both Comparators, $V_{CC} = 30 \text{ V}$	$I_{CC}$	–	0.4	1.0	mA
		–	1.0	2.5	

**ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5.0$  Vdc,  $T_{low} \leq T_A \leq T_{high}$ , unless otherwise noted.)

Characteristic	Symbol	LM392, LM393			LM2903, LM2903V			Unit
		Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage (Note 2) $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$V_{IO}$	-	$\pm 1.0$	$\pm 5.0$ 9.0	-	$\pm 2.0$ 9.0	$\pm 7.0$ 15	mV
Input Offset Current $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$I_{IO}$	-	$\pm 5.0$	$\pm 50$ $\pm 150$	-	$\pm 5.0$ $\pm 50$	$\pm 50$ $\pm 200$	nA
Input Bias Current (Note 3) $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$I_{IB}$	-	25	250 400	-	25 200	250 500	nA
Input Common Mode Voltage Range (Note 3) $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$V_{ICR}$	0 0	-	$V_{CC} - 1.5$ $V_{CC} - 2.0$	0 0	-	$V_{CC} - 1.5$ $V_{CC} - 2.0$	V
Voltage Gain $R_L \geq 15$ k $\Omega$ , $V_{CC} = 15$ Vdc, $T_A = 25^\circ\text{C}$	$A_{VOL}$	50	200	-	25	200	-	V/mV
Large Signal Response Time $V_{in} =$ TTL Logic Swing, $V_{ref} = 1.4$ Vdc $V_{RL} = 5.0$ Vdc, $R_L = 5.1$ k $\Omega$ , $T_A = 25^\circ\text{C}$	-	-	300	-	-	300	-	ns
Response Time (Note 5) $V_{RL} = 5.0$ Vdc, $R_L = 5.1$ k $\Omega$ , $T_A = 25^\circ\text{C}$	$t_{TLH}$	-	1.3	-	-	1.5	-	$\mu\text{s}$
Input Differential Voltage (Note 6) All $V_{in} \geq$ Gnd or V- Supply (if used)	$V_{ID}$	-	-	$V_{CC}$	-	-	$V_{CC}$	V
Output Sink Current $V_{in} \geq 1.0$ Vdc, $V_{in+} = 0$ Vdc, $V_O \leq 1.5$ Vdc, $T_A = 25^\circ\text{C}$	$I_{Sink}$	6.0	16	-	6.0	16	-	mA
Output Saturation Voltage $V_{in} \geq 1.0$ Vdc, $V_{in+} = 0$ , $I_{Sink} \leq 4.0$ mA, $T_A = 25^\circ\text{C}$ $T_{low} \leq T_A \leq T_{high}$	$V_{OL}$	-	150	400 700	-	- 200	400 700	mV
Output Leakage Current $V_{in-} = 0$ V, $V_{in+} \geq 1.0$ Vdc, $V_O = 5.0$ Vdc, $T_A = 25^\circ\text{C}$ $V_{in-} = 0$ V, $V_{in+} \geq 1.0$ Vdc, $V_O = 30$ Vdc, $T_{low} \leq T_A \leq T_{high}$	$I_{OL}$	-	0.1	-	-	0.1	-	nA
Supply Current $R_L = \infty$ Both Comparators, $T_A = 25^\circ\text{C}$ $R_L = \infty$ Both Comparators, $V_{CC} = 30$ V	$I_{CC}$	-	0.4	1.0 2.5	-	0.4 -	1.0 2.5	mA

\*  $T_{low} = 0^\circ\text{C}$ ,  $T_{high} = +70^\circ\text{C}$  for LM393/393A  
 LM293  $T_{low} = -25^\circ\text{C}$ ,  $T_{high} = +85^\circ\text{C}$   
 LM2903  $T_{low} = -40^\circ\text{C}$ ,  $T_{high} = +105^\circ\text{C}$   
 LM2903V  $T_{low} = -40^\circ\text{C}$ ,  $T_{high} = +125^\circ\text{C}$

- NOTES:** 2. At output switch point,  $V_O = 1.4$  Vdc,  $R_S = 0$   $\Omega$  with  $V_{CC}$  from 5.0 Vdc to 30 Vdc, and over the full input common mode range (0 V to  $V_{CC} - 1.5$  V).  
 3. Due to the PNP transistor inputs, bias current will flow out of the inputs. This current is essentially constant, independent of the output state, therefore, no loading changes will exist on the input lines.  
 5. Response time is specified with a 100 mV step and 5.0 mV of overdrive. With larger magnitudes of overdrive faster response times are obtainable.  
 6. The comparator will exhibit proper output state if one of the inputs becomes greater than  $V_{CC}$ , the other input must remain within the common mode range. The low input state must not be less than  $-0.3$  V of ground or minus supply.



REGULADORES DE TENSÃO.



LM317L 100MA REGULADOR AJUSTÁVEL



LM317L

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LM317L 3-Terminal Adjustable Regulator

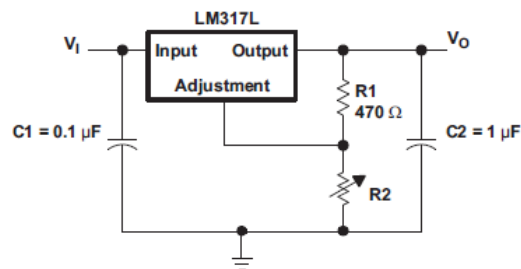
1 Features

- Output Voltage Range Adjustable 1.25 V to 32 V When Used With External Resistor Divider
- Output Current Capability of 100 mA
- Input Regulation Typically 0.01% Per Input-Voltage Change
- Output Regulation Typically 0.5%
- Ripple Rejection Typically 80 dB
- For Higher Output Current Requirements, See LM317M (500 mA) and LM317 (1.5 A)

2 Applications

- Electronic Points of Sale
- Medical, Health, and Fitness Applications
- Printers
- Appliances and White Goods
- TV Set-Top Boxes

4 Simplified Schematic



3 Description

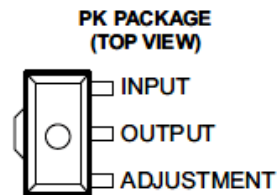
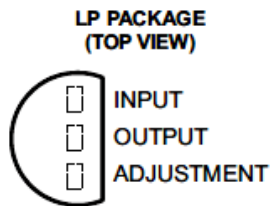
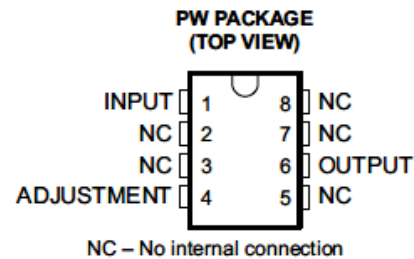
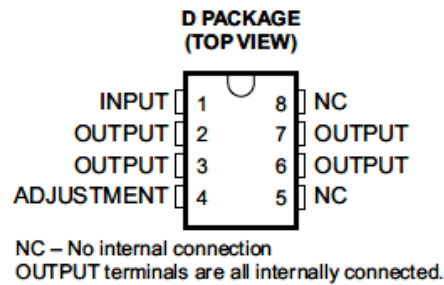
The LM317L device is an adjustable, 3-terminal, positive-voltage regulator capable of supplying 100 mA over an output-voltage range of 1.25 V to 32 V. It is exceptionally easy to use and requires only two external resistors to set the output voltage.

Device Information

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LM317L	SOIC (8)	4.90 mm × 3.91 mm
	TO-92 (3)	4.30 mm × 4.30 mm
	SOT-89 (3)	4.50 mm × 2.50 mm
	TSSOP (8)	3.00 mm × 4.40 mm



6 Pin Configuration and Functions



Pin Functions

NAME	D, PW	LP, PK	TYPE	DESCRIPTION
ADJUSTMENT	4	√	I	Output feedback voltage
INPUT	1	√	I	Input supply voltage
NC	5	√	—	No connect
	8			
OUTPUT	2	√	O	Regulated output voltage
	3			
	6			
	7			



LM317L

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## 7 Specifications

### 7.1 Absolute Maximum Ratings<sup>(1)</sup>

over operating temperature range (unless otherwise noted)

	MIN	MAX	UNIT
$V_I - V_O$ Input-to-output differential voltage		35	V
$T_J$ Operating virtual-junction temperature		150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 7.2 Handling Ratings

	MIN	MAX	UNIT	
$T_{stg}$ Storage temperature range	-65	150	°C	
$V_{(ESD)}$ Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	0	3000	V
	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	0	2000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 7.3 Recommended Operating Conditions

	MIN	MAX	UNIT	
$V_O$ Output voltage	1.25	32	V	
$V_I - V_O$ Input-to-output voltage differential	2.5	32	V	
$I_O$ Output current	2.5	100	mA	
$T_J$ Operating virtual-junction temperature	LM317LC	0	125	°C
	LM317LI	-40	125	

### 7.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>	LM317L				UNIT
	D 8 PINS	LP 3 PINS	PK 3 PINS	PW 8 PINS	
$R\theta_{JA}$ Junction-to-ambient thermal resistance	97.1	139.5	51.5	149.4	°C/W

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report (SPRA953).



LM317L

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7.5 Electrical Characteristics

over recommended operating virtual-junction temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS <sup>(1)</sup>		MIN	TYP	MAX	UNIT
Input voltage regulation <sup>(2)</sup>	$V_I - V_O = 5\text{ V to }35\text{ V}$	$T_J = 25^\circ\text{C}$		0.01	0.02	%V
		$I_O = 2.5\text{ mA to }100\text{ mA}$		0.02	0.05	
Ripple regulation	$V_O = 10\text{ V,}$	$f = 120\text{ Hz}$		65		dB
	$V_O = 10\text{ V,}$ 10- $\mu\text{F}$ capacitor between ADJUSTMENT and ground		66	80		
Output voltage regulation	$V_I = 5\text{ V to }35\text{ V, } T_J = 25^\circ\text{C,}$ $I_O = 2.5\text{ mA to }100\text{ mA,}$	$V_O \leq 5\text{ V}$		25		mV
		$V_O \geq 5\text{ V}$		5		mV/V
	$V_I = 5\text{ V to }35\text{ V,}$ $I_O = 2.5\text{ mA to }100\text{ mA}$	$V_O \leq 5\text{ V}$		50		mV
		$V_O \geq 5\text{ V}$		10		mV/V
Output voltage change with temperature	$T_J = 0^\circ\text{C to }125^\circ\text{C}$			10		mV/V
Output voltage long-term drift	After 1000 hours at $T_J = 125^\circ\text{C}$ and $V_I - V_O = 35\text{ V}$			3	10	mV/V
Output noise voltage	$f = 10\text{ Hz to }10\text{ kHz,}$	$T_J = 25^\circ\text{C}$		30		$\mu\text{V/V}$
Minimum output current to maintain regulation	$V_I - V_O = 35\text{ V}$			1.5	2.5	mA
Peak output current	$V_I - V_O \leq 35\text{ V}$		100	200		mA
ADJUSTMENT current				50	100	$\mu\text{A}$
Change in ADJUSTMENT current	$V_I - V_O = 2.5\text{ V to }35\text{ V,}$	$I_O = 2.5\text{ mA to }100\text{ mA}$		0.2	5	$\mu\text{A}$
Reference voltage (output to ADJUSTMENT)	$V_I - V_O = 5\text{ V to }35\text{ V,}$ $P \leq \text{rated dissipation}$	$I_O = 2.5\text{ mA to }100\text{ mA,}$	1.2	1.25	1.3	V

- (1) Unless otherwise noted, these specifications apply for the following test conditions:  $V_I - V_O = 5\text{ V}$  and  $I_O = 40\text{ mA}$ . Pulse-testing techniques must be used that maintain the junction temperature as close to the ambient temperature as possible. All characteristics are measured with a 0.1- $\mu\text{F}$  capacitor across the input and a 1- $\mu\text{F}$  capacitor across the output.
- (2) Input voltage regulation is expressed here as the percentage change in output voltage per 1-V change at the input.

7.6 Typical Characteristics

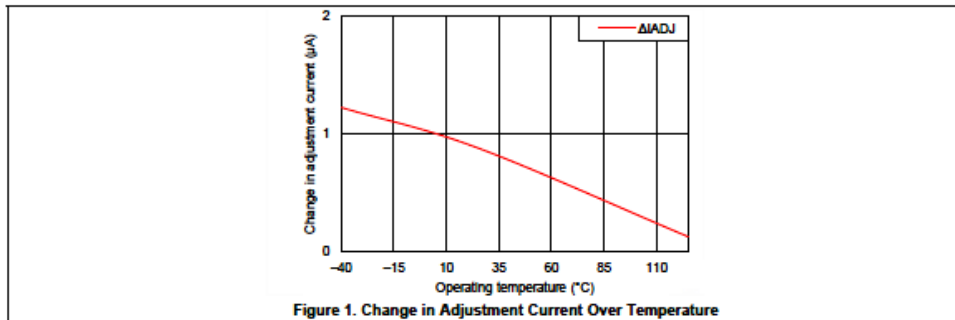


Figure 1. Change in Adjustment Current Over Temperature



LM317L

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9 Application and Implementation

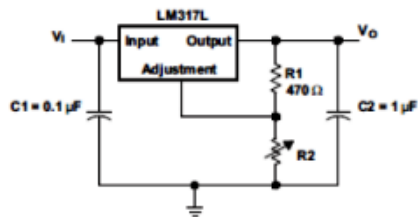
NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The two output resistors are the only components required to adjust  $V_{OUT}$ .

9.2 Typical Application



9.2.1 Design Requirements

1. Use of an input bypass capacitor is recommended if regulator is far from the filter capacitors.
2. For this design example, use the parameters listed in Table 1.
3. Use of an output capacitor improves transient response, but is optional.

Table 1. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Input voltage range	(Output Voltage + 2.5 V) to 32 V
Output voltage	$V_{REF} \times (1 + R_2 / R_1) + I_{ADJ} \times R_2$

9.2.2 Detailed Design Procedure

9.2.2.1 Input Capacitor

An input capacitor is not required, but it is recommended, particularly if the regulator is not in close proximity to the power-supply filter capacitors. A 0.1-µF ceramic or 1-µF tantalum provides sufficient bypassing for most applications, especially when adjustment and output capacitors are used.

9.2.2.2 Output Capacitor

An output capacitor improves transient response, but it not needed for stability.

9.2.2.3 Feedback Resistors

The feedback resistor set the output voltage using Equation 2.

$$V_{REF} \times (1 + R_2 / R_1) + I_{ADJ} \times R_2 \tag{2}$$

## LM-317 REGULADOR DE TENSÃO AJUSTÁVEL 1,5A 1,2V A 37V



[www.fairchildsemi.com](http://www.fairchildsemi.com)

## LM317

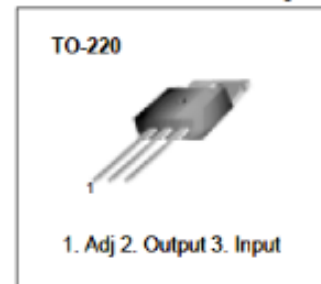
### 3-Terminal Positive Adjustable Regulator

#### Features

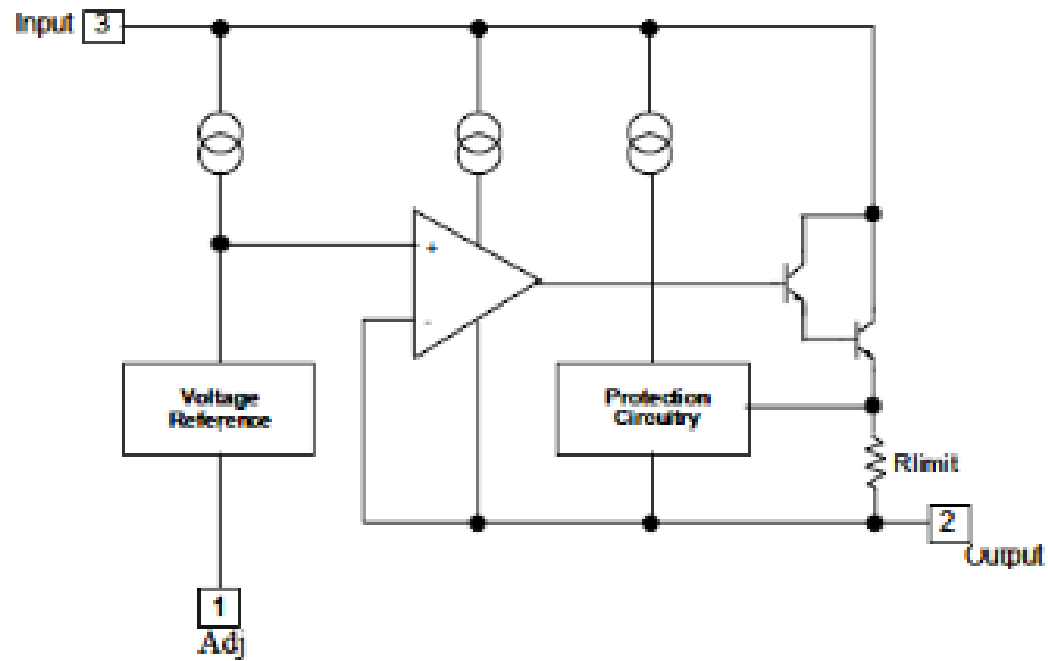
- Output Current In Excess of 1.5A
- Output Adjustable Between 1.2V and 37V
- Internal Thermal Overload Protection
- Internal Short Circuit Current Limiting
- Output Transistor Safe Operating Area Compensation
- TO-220 Package

#### Description

This monolithic integrated circuit is an adjustable 3-terminal positive voltage regulator designed to supply more than 1.5A of load current with an output voltage adjustable over a 1.2 to 37V. It employs internal current limiting, thermal shut-down and safe area compensation.



## Internal Block Diagram



LM317

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**Absolute Maximum Ratings**

Parameter	Symbol	Value	Unit
Input-Output Voltage Differential	$V_I - V_O$	40	V
Lead Temperature	$T_{LEAD}$	230	°C
Power Dissipation	$P_D$	Internally limited	W
Operating Junction Temperature Range	$T_j$	0 ~ +125	°C
Storage Temperature Range	$T_{STG}$	-65 ~ +125	°C
Temperature Coefficient of Output Voltage	$\Delta V_o / \Delta T$	±0.02	%/°C



**Electrical Characteristics**(V<sub>I</sub>-V<sub>O</sub>=5V, I<sub>O</sub>= 0.5A, 0°C ≤ T<sub>J</sub> ≤ + 125°C, I<sub>MAX</sub> = 1.5A, P<sub>D</sub>MAX = 20W, unless otherwise specified)

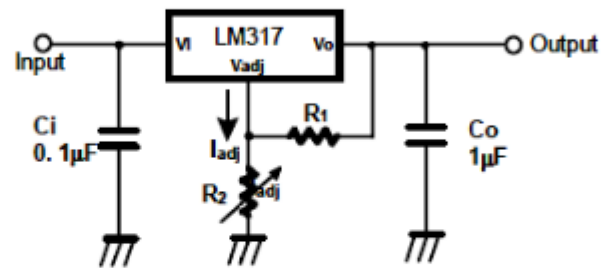
Parameter	Symbol	Conditions	Min	Typ.	Max.	Unit
Line Regulation (Note1)	R <sub>line</sub>	T <sub>A</sub> = +25°C 3V ≤ V <sub>I</sub> - V <sub>O</sub> ≤ 40V	-	0.01	0.04	% / V
		3V ≤ V <sub>I</sub> - V <sub>O</sub> ≤ 40V	-	0.02	0.07	% / V
Load Regulation (Note1)	R <sub>load</sub>	T <sub>A</sub> = +25°C, 10mA ≤ I <sub>O</sub> ≤ I <sub>MAX</sub> V <sub>O</sub> < 5V V <sub>O</sub> ≥ 5V	-	18 0.4	25 0.5	mV% / V <sub>O</sub>
		10mA ≤ I <sub>O</sub> ≤ I <sub>MAX</sub> V <sub>O</sub> < 5V V <sub>O</sub> ≥ 5V	-	40 0.8	70 1.5	mV% / V <sub>O</sub>
Adjustable Pin Current	I <sub>ADJ</sub>	-	-	46	100	μA
Adjustable Pin Current Change	ΔI <sub>ADJ</sub>	3V ≤ V <sub>I</sub> - V <sub>O</sub> ≤ 40V 10mA ≤ I <sub>O</sub> ≤ I <sub>MAX</sub> P <sub>D</sub> ≤ P <sub>MAX</sub>	-	2.0	5	μA
Reference Voltage	V <sub>REF</sub>	3V ≤ V <sub>IN</sub> - V <sub>O</sub> ≤ 40V 10mA ≤ I <sub>O</sub> ≤ I <sub>MAX</sub> P <sub>D</sub> ≤ P <sub>MAX</sub>	1.20	1.25	1.30	V
Temperature Stability	ST <sub>T</sub>	-	-	0.7	-	% / V <sub>O</sub>
Minimum Load Current to Maintain Regulation	I <sub>L(MIN)</sub>	V <sub>I</sub> - V <sub>O</sub> = 40V	-	3.5	12	mA
Maximum Output Current	I <sub>O(MAX)</sub>	V <sub>I</sub> - V <sub>O</sub> ≤ 15V, P <sub>D</sub> ≤ P <sub>MAX</sub> V <sub>I</sub> - V <sub>O</sub> ≤ 40V, P <sub>D</sub> ≤ P <sub>MAX</sub> T <sub>A</sub> =25°C	1.0	2.2 0.3	-	A
RMS Noise, % of V <sub>OUT</sub>	e <sub>N</sub>	T <sub>A</sub> = +25°C, 10Hz ≤ f ≤ 10KHz	-	0.003	0.01	% / V <sub>O</sub>
Ripple Rejection	RR	V <sub>O</sub> = 10V, f = 120Hz without C <sub>ADJ</sub> C <sub>ADJ</sub> = 10μF (Note2)	66	60 75	-	dB
Long-Term Stability, T <sub>J</sub> = T <sub>HIGH</sub>	ST	T <sub>A</sub> = +25°C for end point measurements, 1000HR	-	0.3	1	%
Thermal Resistance Junction to Case	R <sub>θJC</sub>	-	-	5	-	°C / W

**Note:**

1. Load and line regulation are specified at constant junction temperature. Change in V<sub>D</sub> due to heating effects must be taken into account separately. Pulse testing with low duty is used. (P<sub>MAX</sub> = 20W)
2. C<sub>ADJ</sub>, when used, is connected between the adjustment pin and ground.

## LM317

## Typical Application



$$V_O = 1.25V (1 + R_2 / R_1) + I_{adj} R_2$$

Figure 5. Programmable Regulator

$C_i$  is required when regulator is located an appreciable distance from power supply filter.

$C_o$  is not needed for stability, however, it does improve transient response.

Since  $I_{ADJ}$  is controlled to less than  $100\mu A$ , the error associated with this term is negligible in most applications.

## LM150 LM350 Regulador ajustável 3A



LM150, LM350-N, LM350A

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## LM150/LM350A/LM350 3-Amp Adjustable Regulators

Check for Samples: LM150, LM350-N, LM350A

## FEATURES

- Adjustable Output Down to 1.2V
- Guaranteed 3A output Current
- Guaranteed Thermal Regulation
- Output is Short Circuit Protected
- Current Limit Constant with Temperature
- P\* Product Enhancement Tested
- 86 dB Ripple Rejection
- Ensured 1% Output Voltage Tolerance (LM350A)
- Ensured Max. 0.01%/V Line Regulation (LM350A)
- Ensured Max. 0.3% Load Regulation (LM350A)

## APPLICATIONS

- Adjustable Power supplies
- Constant Current Regulators
- Battery Chargers

## DESCRIPTION

The LM150 series of adjustable 3-terminal positive voltage regulators is capable of supplying in excess of 3A over a 1.2V to 33V output range. They are exceptionally easy to use and require only 2 external resistors to set the output voltage. Further, both line and load regulation are comparable to discrete designs. Also, the LM150 is packaged in standard transistor packages which are easily mounted and handled.

In addition to higher performance than fixed regulators, the LM150 series offers full overload protection available only in IC's. Included on the chip are current limit, thermal overload protection and safe area protection. All overload protection circuitry remains fully functional even if the adjustment terminal is accidentally disconnected.

Normally, no capacitors are needed unless the device is situated more than 6 inches from the input filter capacitors in which case an input bypass is needed. An output capacitor can be added to improve transient response, while bypassing the adjustment pin will increase the regulator's ripple rejection.

Besides replacing fixed regulators or discrete designs, the LM150 is useful in a wide variety of other applications. Since the regulator is "floating" and sees only the input-to-output differential voltage, supplies of several hundred volts can be regulated as long as the maximum input to output differential is not exceeded, i.e., avoid short-circuiting the output.

By connecting a fixed resistor between the adjustment pin and output, the LM150 can be used as a precision current regulator. Supplies with electronic shutdown can be achieved by clamping the adjustment terminal to ground which programs the output to 1.2V where most loads draw little current.

The part numbers in the LM150 series which have a NDS suffix are packaged in a standard Steel TO-3 package, while those with a NDE suffix are packaged in a TO-220 plastic package. The LM150 is rated for  $-55^{\circ}\text{C} \leq T_J \leq +150^{\circ}\text{C}$ , while the LM350A is rated for  $-40^{\circ}\text{C} \leq T_J \leq +125^{\circ}\text{C}$ , and the LM350 is rated for  $0^{\circ}\text{C} \leq T_J \leq +125^{\circ}\text{C}$ .

LM150, LM350-N, LM350A



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

**Absolute Maximum Ratings**<sup>(1)(2)(3)</sup>

Power Dissipation	Internally Limited	
Input-Output Voltage Differential	+35V	
Storage Temperature	-65°C to +150°C	
Lead Temperature	Metal Package (Soldering, 10 sec.)	300°C
	Plastic Package (Soldering, 4 sec.)	260°C
ESD Tolerance	TBD	
Operating Temperature Range	LM150	-55°C ≤ T <sub>J</sub> ≤ +150°C
	LM350A	-40°C ≤ T <sub>J</sub> ≤ +125°C
	LM350	0°C ≤ T <sub>J</sub> ≤ +125°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics.
- (2) Refer to RETS150K drawing for military specifications of the LM150K.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.

**Connection Diagram**

Case is Output

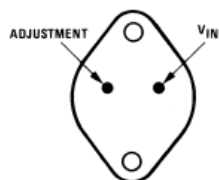


Figure 1. (TO-3 STEEL) Metal Can Package  
Bottom View  
See Package Number NDS0002A

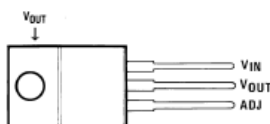


Figure 2. (TO-220) Plastic Package  
Front View  
See Package Number NDE0003B

## Electrical Characteristics

Specifications with standard type face are for  $T_J = 25^\circ\text{C}$ , and those with **boldface type** apply over **full Operating Temperature Range**. Unless otherwise specified,  $V_{IN} - V_{OUT} = 5\text{V}$ , and  $I_{OUT} = 10\text{ mA}$ .<sup>(1)</sup>

Parameter	Conditions	LM150			Units
		Min	Typ	Max	
Reference Voltage	$3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$ , $10\text{ mA} \leq I_{OUT} \leq 3\text{A}$ , $P \leq 30\text{W}$	<b>1.20</b>	1.25	1.30	V
Line Regulation	$3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$ <sup>(2)</sup>		0.005	0.01	%/V
			<b>0.02</b>	<b>0.05</b>	%/V
Load Regulation	$10\text{ mA} \leq I_{OUT} \leq 3\text{A}$ <sup>(2)</sup>		0.1	0.3	%
			<b>0.3</b>	<b>1</b>	%
Thermal Regulation	20 ms Pulse		0.002	0.01	%/W
Adjustment Pin Current			50	100	$\mu\text{A}$
Adjustment Pin Current Change	$10\text{ mA} \leq I_{OUT} \leq 3\text{A}$ , $3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$		0.2	5	$\mu\text{A}$
Temperature Stability	$T_{MIN} \leq T_J \leq T_{MAX}$		1		%
Minimum Load Current	$V_{IN} - V_{OUT} = 35\text{V}$		3.5	5	mA
Current Limit	$V_{IN} - V_{OUT} \leq 10\text{V}$	<b>3.0</b>	4.5		A
	$V_{IN} - V_{OUT} = 30\text{V}$	<b>0.3</b>	1		A
RMS Output Noise, % of $V_{OUT}$	$10\text{ Hz} \leq f \leq 10\text{ kHz}$		0.001		%
Ripple Rejection Ratio	$V_{OUT} = 10\text{V}$ , $f = 120\text{ Hz}$ , $C_{ADJ} = 0\text{ }\mu\text{F}$		65		dB
	$V_{OUT} = 10\text{V}$ , $f = 120\text{ Hz}$ , $C_{ADJ} = 10\text{ }\mu\text{F}$	<b>66</b>	86		dB
Long-Term Stability	$T_J = 125^\circ\text{C}$ , 1000 hrs		0.3	1	%
Thermal Resistance, Junction to Case	NDS Package		1.2	1.5	$^\circ\text{C/W}$
Thermal Resistance, Junction to Ambient (No Heat Sink)	NDS Package		35		$^\circ\text{C/W}$

- (1) These specifications are applicable for power dissipations up to 30W for the TO-3 (NDS) package and 25W for the TO-220 (NDE) package. Power dissipation is ensured at these values up to 15V input-output differential. Above 15V differential, power dissipation will be limited by internal protection circuitry. All limits (i.e., the numbers in the Min. and Max. columns) are ensured to AOQL (Average Outgoing Quality Level).
- (2) Regulation is measured at a constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specifications for thermal regulation.

### Electrical Characteristics

Specifications with standard type face are for  $T_J = 25^\circ\text{C}$ , and those with **boldface type** apply over **full Operating Temperature Range**. Unless otherwise specified,  $V_{IN} - V_{OUT} = 5\text{V}$ , and  $I_{OUT} = 10\text{ mA}$ .<sup>(1)</sup>

Parameter	Conditions	LM350A			LM350			Units
		Min	Typ	Max	Min	Typ	Max	
Reference Voltage	$I_{OUT} = 10\text{ mA}$ , $T_J = 25^\circ\text{C}$ $3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$ , $10\text{ mA} \leq I_{OUT} \leq 3\text{A}$ , $P \leq 30\text{W}$	1.238	1.250	1.262				V
		<b>1.225</b>	<b>1.250</b>	<b>1.270</b>	<b>1.20</b>	<b>1.25</b>	<b>1.30</b>	V
Line Regulation	$3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}^{(2)}$		0.005	0.01		0.005	0.03	%/V
			<b>0.02</b>	<b>0.05</b>		<b>0.02</b>	<b>0.07</b>	%/V
Load Regulation	$10\text{ mA} \leq I_{OUT} \leq 3\text{A}^{(2)}$		0.1	0.3		0.1	0.5	%
			<b>0.3</b>	<b>1</b>		<b>0.3</b>	<b>1.5</b>	%
Thermal Regulation	20 ms Pulse		0.002	0.01		0.002	0.03	%/W
Adjustment Pin Current			50	100		50	100	$\mu\text{A}$
Adjustment Pin Current Change	$10\text{ mA} \leq I_{OUT} \leq 3\text{A}$ , $3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$		0.2	5		0.2	5	$\mu\text{A}$
Temperature Stability	$T_{MIN} \leq T_J \leq T_{MAX}$		1			1		%
Minimum Load Current	$V_{IN} - V_{OUT} = 35\text{V}$		3.5	10		3.5	10	mA
Current Limit	$V_{IN} - V_{OUT} \leq 10\text{V}$	3.0	4.5		3.0	4.5		A
	$V_{IN} - V_{OUT} = 30\text{V}$	0.3	1		0.25	1		A
RMS Output Noise, % of $V_{OUT}$	$10\text{ Hz} \leq f \leq 10\text{ kHz}$		0.001			0.001		%
Ripple Rejection Ratio	$V_{OUT} = 10\text{V}$ , $f = 120\text{ Hz}$ , $C_{ADJ} = 0\text{ }\mu\text{F}$		65			65		dB
	$V_{OUT} = 10\text{V}$ , $f = 120\text{ Hz}$ , $C_{ADJ} = 10\text{ }\mu\text{F}$	66	86		66	86		dB
Long-Term Stability	$T_J = 125^\circ\text{C}$ , 1000 hrs		0.25	1		0.25	1	%
Thermal Resistance, Junction to Case	NDS Package					1.2	1.5	$^\circ\text{C/W}$
	NDE Package		3	4		3	4	$^\circ\text{C/W}$
Thermal Resistance, Junction to Ambient (No Heat Sink)	NDS Package					35		$^\circ\text{C/W}$
	NDE Package		50			50		$^\circ\text{C/W}$

- (1) These specifications are applicable for power dissipations up to 30W for the TO-3 (NDS) package and 25W for the TO-220 (NDE) package. Power dissipation is ensured at these values up to 15V input-output differential. Above 15V differential, power dissipation will be limited by internal protection circuitry. All limits (i.e., the numbers in the Min. and Max. columns) are ensured to AOQL (Average Outgoing Quality Level).
- (2) Regulation is measured at a constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specifications for thermal regulation.

## LM150, LM350-N, LM350A



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## APPLICATION HINTS

In operation, the LM150 develops a nominal 1.25V reference voltage,  $V_{REF}$ , between the output and adjustment terminal. The reference voltage is impressed across program resistor  $R1$  and, since the voltage is constant, a constant current  $I_1$  then flows through the output set resistor  $R2$ , giving an output voltage of

$$V_{OUT} = V_{REF} \left( 1 + \frac{R2}{R1} \right) + I_{ADJ} R2. \quad (1)$$

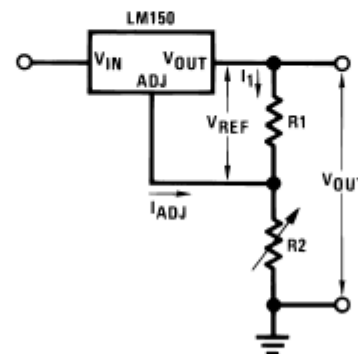


Figure 15.

Since the 50  $\mu$ A current from the adjustment terminal represents an error term, the LM150 was designed to minimize  $I_{ADJ}$  and make it very constant with line and load changes. To do this, all quiescent operating current is returned to the output establishing a minimum load current requirement. If there is insufficient load on the output, the output will rise.

## EXTERNAL CAPACITORS

An input bypass capacitor is recommended. A 0.1  $\mu\text{F}$  disc or 1  $\mu\text{F}$  solid tantalum on the input is suitable input bypassing for almost all applications. The device is more sensitive to the absence of input bypassing when adjustment or output capacitors are used but the above values will eliminate the possibility of problems.

The adjustment terminal can be bypassed to ground on the LM150 to improve ripple rejection. This bypass capacitor prevents ripple from being amplified as the output voltage is increased. With a 10  $\mu\text{F}$  bypass capacitor 86 dB ripple rejection is obtainable at any output level. Increases over 10  $\mu\text{F}$  do not appreciably improve the ripple rejection at frequencies above 120 Hz. If the bypass capacitor is used, it is sometimes necessary to include protection diodes to prevent the capacitor from discharging through internal low current paths and damaging the device.

In general, the best type of capacitors to use is solid tantalum. Solid tantalum capacitors have low impedance even at high frequencies. Depending upon capacitor construction, it takes about 25  $\mu\text{F}$  in aluminum electrolytic to equal 1  $\mu\text{F}$  solid tantalum at high frequencies. Ceramic capacitors are also good at high frequencies, but some types have a large decrease in capacitance at frequencies around 0.5 MHz. For this reason, 0.01  $\mu\text{F}$  disc may seem to work better than a 0.1  $\mu\text{F}$  disc as a bypass.

Although the LM150 is stable with no output capacitors, like any feedback circuit, certain values of external capacitance can cause excessive ringing. This occurs with values between 500 pF and 5000 pF. A 1  $\mu\text{F}$  solid tantalum (or 25  $\mu\text{F}$  aluminum electrolytic) on the output swamps this effect and insures stability.

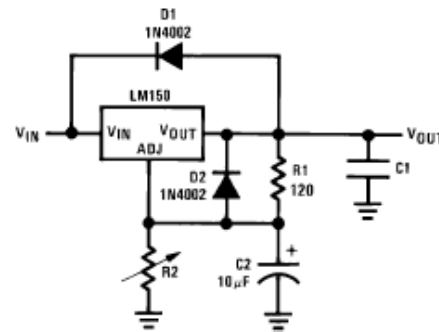


## PROTECTION DIODES

When external capacitors are used with *any* IC regulator it is sometimes necessary to add protection diodes to prevent the capacitors from discharging through low current points into the regulator. Most 10  $\mu\text{F}$  capacitors have low enough internal series resistance to deliver 20A spikes when shorted. Although the surge is short, there is enough energy to damage parts of the IC.

When an output capacitor is connected to a regulator and the input is shorted, the output capacitor will discharge into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage of the regulator, and the rate of decrease of  $V_{\text{IN}}$ . In the LM150, this discharge path is through a large junction that is able to sustain 25A surge with no problem. This is not true of other types of positive regulators. For output capacitors of 25  $\mu\text{F}$  or less, there is no need to use diodes.

The bypass capacitor on the adjustment terminal can discharge through a low current junction. Discharge occurs when *either* the input or output is shorted. Internal to the LM150 is a 50 $\Omega$  resistor which limits the peak discharge current. No protection is needed for output voltages of 25V or less and 10  $\mu\text{F}$  capacitance. Figure 17 shows an LM150 with protection diodes included for use with outputs greater than 25V and high values of output capacitance.



D1 protects against C1  
D2 protects against C2

Figure 17. Regulator with Protection Diodes

$$V_{\text{OUT}} = 1.25\text{V} \left( 1 + \frac{R2}{R1} \right) + I_{\text{ADJ}}R2 \quad (2)$$

## 9.2 Typical Application

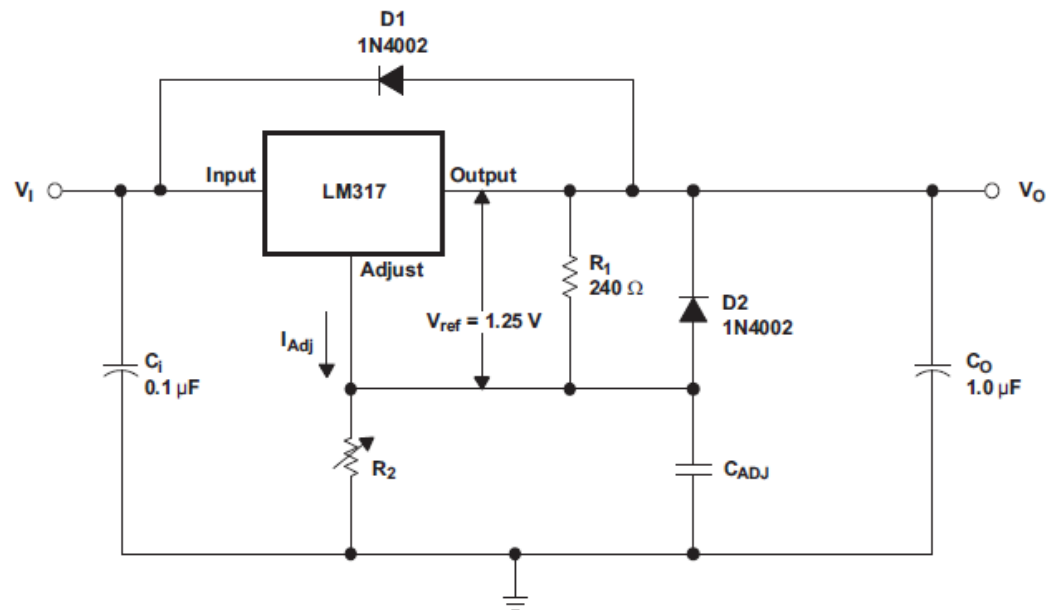


Figure 9. Adjustable Voltage Regulator

## 9.2.2 Detailed Design Procedure

$V_O$  is calculated as shown in Equation 1.  $I_{ADJ}$  is typically 50  $\mu\text{A}$  and negligible in most applications.

$$V_O = V_{REF} (1 + R_2 / R_1) + (I_{ADJ} \times R_2)$$

9.3.1 0-V to 30-V Regulator Circuit

Here, the voltage is determined by  $V_{OUT} = V_{REF} \left( 1 + \frac{R_2 + R_3}{R_1} \right) - 10V$

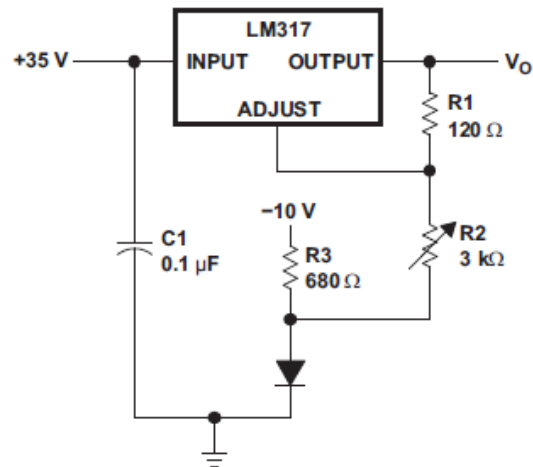


Figure 12. 0-V to 30-V Regulator Circuit

9.3.3 Precision Current-Limiter Circuit

This application limits the output current to the  $I_{LIMIT}$  in the diagram.

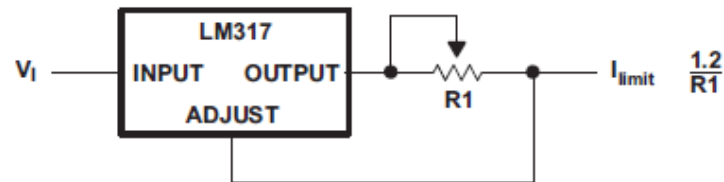


Figure 14. Precision Current-Limiter Circuit

### 9.3.6 Battery-Charger Circuit

The series resistor limits the current output of the LM317, minimizing damage to the battery cell.

$$V_{OUT} = 1.25 \text{ V} \times \left(1 + \frac{R2}{R1}\right)$$

$$I_{OUT(short)} = \frac{1.25\text{V}}{R_S}$$

$$\text{Output Impedance} = R_S \times \left(1 + \frac{R2}{R1}\right)$$

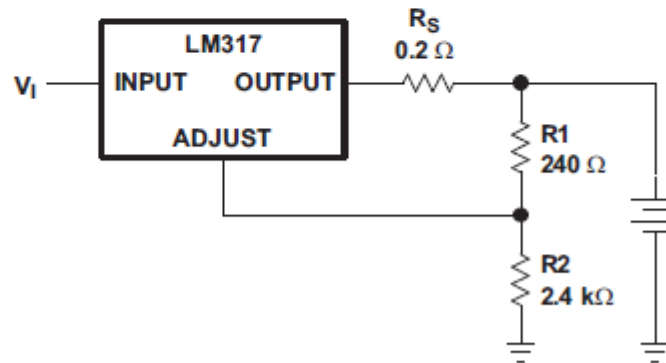


Figure 17. Battery-Charger Circuit

## System Examples (continued)

### 9.3.12 High-Current Adjustable Regulator Circuit

The NPNs at the top of the schematic allow higher currents at  $V_{OUT}$  than the LM317 can provide, while still keeping the output voltage at levels determined by the adjustment pin resistor divider of the LM317.

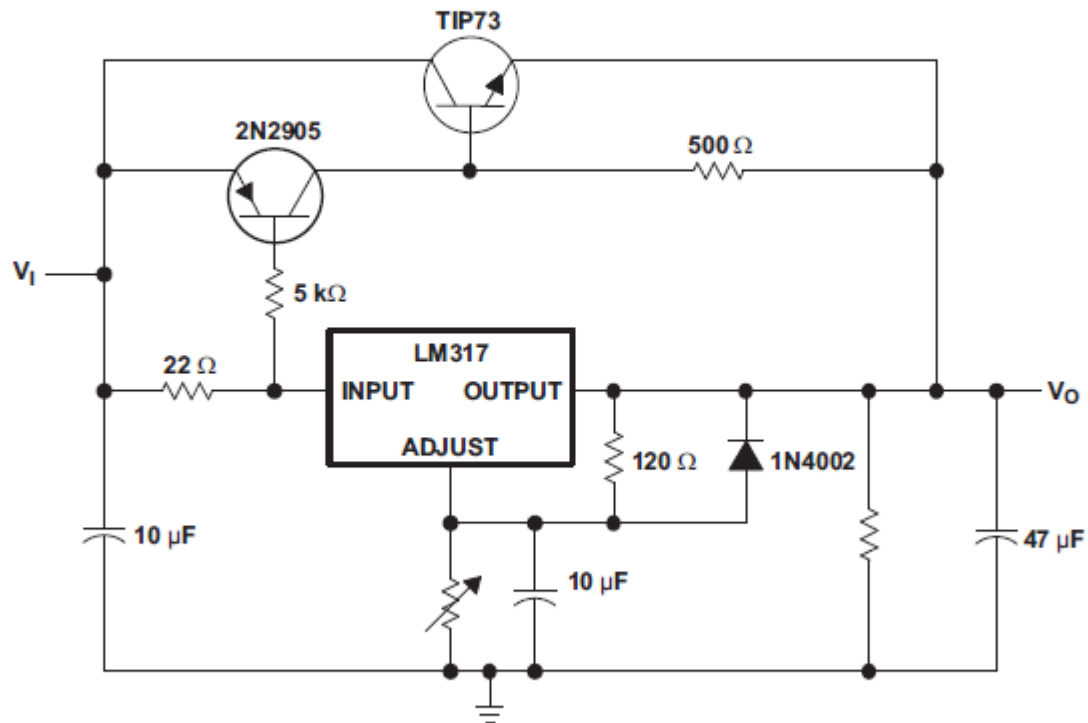


Figure 23. High-Current Adjustable Regulator Circuit

LM138 5A REGULADOR AJUSTÁVEL



LM138/238  
LM338

THREE-TERMINAL 5-A  
ADJUSTABLE VOLTAGE REGULATORS

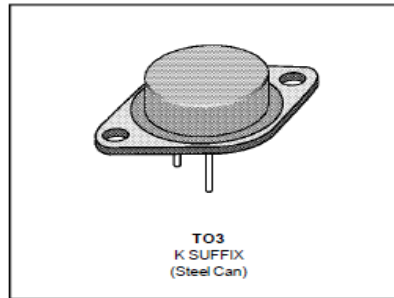
- GUARANTEED 7A PEAK OUTPUT CURRENT
- GUARANTEED 5A OUTPUT CURRENT
- ADJUSTABLE OUTPUT DOWN TO 1.2V
- LINE REGULATION TYPICALLY 0.005% /V
- LOAD REGULATION TYPICALLY 0.1%
- GUARANTEED THERMAL REGULATION
- CURRENT LIMIT CONSTANT WITH TEMPERATURE
- STANDARD 3-LEAD TRANSISTOR PACKAGE

DESCRIPTION

The LM138/LM238/LM338 are adjustable 3-terminal positive voltage regulators capable of supplying in excess of 5A over a 1.2V to 32V output range. They are exceptionally easy to use and require only 2 resistors to set the output voltage. Careful circuit design has resulted in outstanding load and line regulation comparable to many commercial power supplies. The LM138 family is supplied in a standard 3-lead transistor package.

A unique feature of the LM138 family is time-dependent current limiting. The current limit circuitry allows peak currents of up to 12A to be drawn from the regulator for short periods of time. This allows the LM138 to be used with heavy transient loads and speeds start-up under full-load conditions. Under sustained loading conditions, the current limit decreases to a safe value protecting the regulator. Also included on the chip are thermal overload protection and safe area protection for the power transistor. Overload protection remains functional even if the adjustment pin is accidentally disconnected.

Normally, no capacitors are needed unless the device is situated far from the input filter capacitors in which case an input bypass is needed. An optional output capacitor can be added to improve transient response. The adjustment terminal can be bypassed to achieve very high ripple rejection ratios which are difficult to achieve with standard 3-terminal regulators. Besides replacing fixed regulators or discrete designs, the LM238 is useful in a wide variety of other applications. Since the regulator is "floating" and sees only the input-to-output differential voltage, supplies of several hundred volts can be regulated as long as the maximum input to input differential is not exceeded. The LM138/LM238/LM338 are packaged in standard steel TO-3 transistor packages. The LM138 is rated for operation from -55°C to 150°C, the LM238 from -25°C to +150°C and the LM338 from 0°C to +125°C.

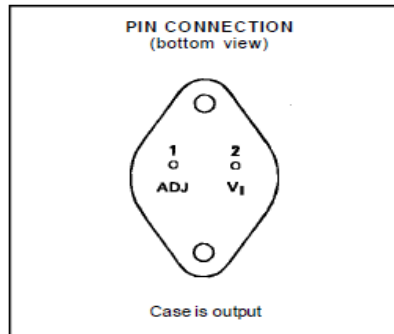


TO3  
K SUFFIX  
(Steel Can)

ORDER CODE

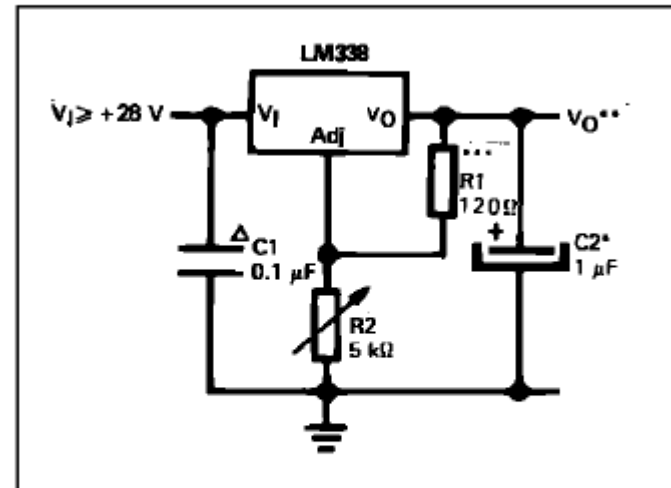
PART NUMBER	TEMPERATURE RANGE	PACKAGE
		K
LM138	-55 °C to + 150 °C	•
LM238	-25 °C to + 150 °C	•
LM338	0 °C to + 125 °C	•

EXAMPLE: LM138K



TYPICAL APPLICATIONS

+ 1.2V to + 25V ADJUSTABLE REGULATOR



- Δ Needed if device is far from filter capacitors.
- \* Optional-improves transient response. Output capacitors in the range of 1μF to 100μF of aluminium or tantalum electrolytic are commonly used to provide improved output impedance and rejection of transients.
- \*\*  $V_0 = 1.25V (1 + \frac{R2}{R1})$
- \*\*\*  $R1 = 240\Omega$  for LM138 and LM238

**LM138-LM238-LM338****ABSOLUTE MAXIMUM RATING**

Symbol	Parameter	Value	Unit	
$P_{tot}$	Power Dissipation	Internally Limited	W	
$V_I - V_O$	Input-Output Voltage Differential	35	V	
$T_{oper}$	Operating Junction Temperature Range	LM138 LM238 LM338	-55 to 150 -25 to 150 0 to 125	°C
$T_{stg}$	Storage Temperature Range	-65 to 150	°C	
$T_{lead}$	Lead Temperature (Soldering, 10 seconds)	300	°C	

**THERMAL CHARACTERISTICS**

Symbol	Parameter	Value	Unit
$R_{th(j-c)}$	Typical Junction-Case Thermal Resistance	1.4	°C/W
$R_{th(j-a)}$	Max Junction-Ambient Thermal Resistance	35	°C/W

## ELECTRICAL CHARACTERISTICS

LM138:  $-55 \leq T_j \leq 150$  °C,  $V_I - V_O = 5V$ ,  $I_O = 2.5A$ LM238:  $-25 \leq T_j \leq 150$  °C,  $V_I - V_O = 5V$ ,  $I_O = 2.5A$ LM338:  $0 \leq T_j \leq 150$  °C,  $V_I - V_O = 5V$ ,  $I_O = 2.5A$ 

Although power dissipation is internally limited, these specifications apply to power dissipation up to 50W (unless otherwise specified).

Symbol	Parameter	LM138-LM238			LM338			Unit
		Min.	Typ.	Max.	Min.	Typ.	Max.	
$K_{VL}$	Line Regulation - (note 1) $T_{amb} = 25$ °C, $3V \leq (V_I - V_O) \leq 35V$		0.005	0.01		0.005	0.03	%/V
$K_{VO}$	Load Regulation $T_{amb} = 25$ °C, $10mA \leq I_O \leq 5A$ $V_O \leq 5V$ - (note 1) $V_O \geq 5V$ - (note 1)		5 0.1	15 0.3		5 0.1	25 0.5	mV %
	Thermal Regulation (pulse = 20 ms)		0.002	0.01		0.002	0.02	%/W
$I_{adj}$	Adjustment Pin Current		45	100		45	100	µA
$\Delta I_{adj}$	Adjustment Pin Current Change $10mA \leq I_L \leq 5A$ , $3V \leq (V_I - V_O) \leq 35V$		0.2	5		0.2	5	µA
$V_{ref}$	Reference Voltage $3V \leq (V_I - V_O) \leq 35V$ , $10mA \leq I_O \leq 5A$ , $P \leq 50W$	1.19	1.24	1.29	1.19	1.24	1.29	V
$K_{VL}$	Line Regulation - (note 1) $3V \leq (V_I - V_O) \leq 35V$		0.02	0.04		0.02	0.06	%/V
$K_{VO}$	Load Regulation $10mA \leq I_O \leq 5A$ $V_O \leq 5V$ - (note 1) $V_O \geq 5V$ - (note 1)		20 0.3	30 0.6		20 0.3	50 1	mV %
$K_{VT}$	Temperature Stability ( $T_{min} \leq T_j \leq T_{max}$ )		1			1		%
$I_{O(min)}$	Minimum Load Current ( $V_I - V_O \leq 35V$ )		3.5	5		3.5	10	mA
$I_{O(max)}$	Current Limit ( $V_I - V_O \leq 10V$ ) DC 0.5 ms Peak $V_I - V_O = 30V$	5 7	8 12 1		5 7	8 12 1		A
	RMS Output Noise, % of $V_O$ ( $T_{amb} = 25$ °C, $10Hz \leq f \leq 10KHz$ )			0.003			0.003	%
$R_{vr}$	Ripple Rejection Ratio $V_O = 10V$ , $f = 120Hz$ $C_{adj} = 10\mu F$	60	60 75		60	60 75		dB
$K_{VH}$	Long Term Stability ( $T_{amb} = 125$ °C)		0.3	1		0.3	1	%

Note 1 : Regulation is measured at constant junction temperature. Changes in output voltage due to heating effects are taken into account separately by thermal rejection.



LM340-5 LM340-12 LM340-15



LM340-N, LM78xx

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## LM340-N/LM78XX Series 3-Terminal Positive Regulators

Check for Samples: [LM340-N](#), [LM78xx](#)

### FEATURES

- Complete Specifications at 1A Load
- Output Voltage Tolerances of  $\pm 2\%$  at  $T_j = 25^\circ\text{C}$  and  $\pm 4\%$  Over the Temperature Range (LM340A)
- Line Regulation of 0.01% of  $V_{\text{OUT}}/V$  of  $\Delta V_{\text{IN}}$  at 1A Load (LM340A)
- Load Regulation of 0.3% of  $V_{\text{OUT}}/A$  (LM340A)
- Internal Thermal Overload Protection
- Internal Short-circuit Current Limit
- Output Transistor Safe Area Protection
- P<sup>+</sup> Product Enhancement Tested

### DESCRIPTION

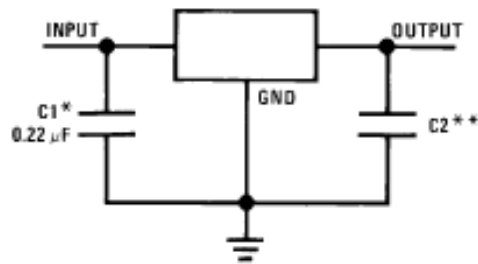
The LM140/LM340A/LM340-N/LM78XXC monolithic 3-terminal positive voltage regulators employ internal current-limiting, thermal shutdown and safe-area compensation, making them essentially indestructible. If adequate heat sinking is provided, they can deliver over 1.0A output current. They are intended as fixed voltage regulators in a wide range of applications including local (on-card) regulation for elimination of noise and distribution problems associated with single-point regulation. In addition to use as fixed voltage regulators, these devices can be used with external components to obtain adjustable output voltages and currents.

Considerable effort was expended to make the entire series of regulators easy to use and minimize the number of external components. It is not necessary to bypass the output, although this does improve transient response. Input bypassing is needed only if the regulator is located far from the filter capacitor of the power supply.

The 5V, 12V, and 15V regulator options are available in the steel TO-3 power package. The LM340A/LM340-N/LM78XXC series is available in the TO-220 plastic power package, and the LM340-N-5.0 is available in the SOT-223 package, as well as the LM340-5.0 and LM340-12 in the surface-mount DPAK/TO-263 package.

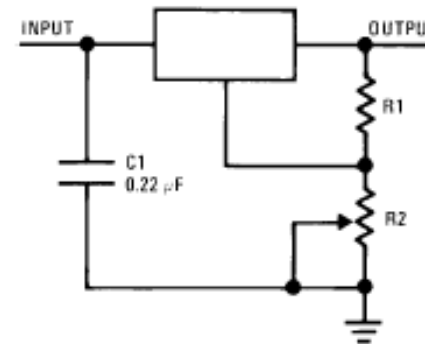


**Typical Applications**



\*Required if the regulator is located far from the power supply filter.  
 \*\*Although no output capacitor is needed for stability, it does help transient response. (If needed, use 0.1 μF, ceramic disc).

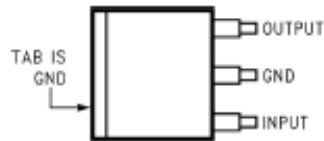
**Figure 1. Fixed Output Regulator**



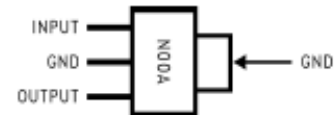
$V_{OUT} = 5V + (5V/R1 + I_Q) R2$   $5V/R1 > 3 I_Q$ ,  
 load regulation ( $L_r$ )  $\approx [(R1 + R2)/R1]$  ( $L_r$  of LM340-5).

**Figure 2. Adjustable Output Regulator**

**Connection Diagrams**



**Figure 5. DDPAK/TO-263 Surface-Mount Package  
Top View  
See Package Number KTT0003B**



**Figure 6. 3-Lead SOT-223  
Top View  
See Package Number DCY**



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

**Absolute Maximum Ratings** <sup>(1)(2)(3)</sup>

DC Input Voltage		35V
Internal Power Dissipation <sup>(4)</sup>		Internally Limited
Maximum Junction Temperature		150°C
Storage Temperature Range		-65°C to +150°C
Lead Temperature (Soldering, 10 sec.)	TO-3 Package (NDS)	300°C
	TO-220 Package (NDE), DDPAK/TO-263 Package (KTT)	230°C
ESD Susceptibility <sup>(5)</sup>		2 kV

LM340-N, LM78xx



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LM340A Electrical Characteristics

$I_{OUT} = 1A$ ,  $0^{\circ}C \leq T_J \leq +125^{\circ}C$  (LM340A) unless otherwise specified<sup>(1)</sup>

Symbol	Output Voltage			5V			12V			15V			Units
	Input Voltage (unless otherwise noted)			10V			19V			23V			
	Parameter	Conditions	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
$V_O$	Output Voltage	$T_J = 25^{\circ}C$	4.9	5	5.1	11.75	12	12.25	14.7	15	15.3	V	
		$P_{D1} \leq 15W$ , $5 mA \leq I_O \leq 1A$	4.8		5.2	11.5		12.5	14.4		15.6	V	
		$V_{MIN} \leq V_{IN} \leq V_{MAX}$	(7.5 $\leq V_{IN} \leq 20$ )				(14.8 $\leq V_{IN} \leq 27$ )				(17.9 $\leq V_{IN} \leq 30$ )	V	
$\Delta V_O$	Line Regulation	$I_O = 500 mA$			10			18			22	mV	
		$\Delta V_{IN}$	(7.5 $\leq V_{IN} \leq 20$ )			(14.8 $\leq V_{IN} \leq 27$ )			(17.9 $\leq V_{IN} \leq 30$ )			V	
		$T_J = 25^{\circ}C$		3	10		4	18		4	22	mV	
		$\Delta V_{IN}$	(7.5 $\leq V_{IN} \leq 20$ )			(14.5 $\leq V_{IN} \leq 27$ )			(17.5 $\leq V_{IN} \leq 30$ )			V	
		$T_J = 25^{\circ}C$ Over Temperature			4		9			10		30	mV
$\Delta V_O$	Load Regulation	$T_J = 25^{\circ}C$		10	25		12	32		12	35	mV	
		5 mA $\leq I_O \leq 1.5A$			15		19				21	mV	
		Over Temperature, 5 mA $\leq I_O \leq 1A$			25		60				75	mV	
$I_Q$	Quiescent Current	$T_J = 25^{\circ}C$			6			6			6	mA	
		Over Temperature			6.5		6.5				6.5	mA	
$\Delta I_Q$	Quiescent Current Change	5 mA $\leq I_O \leq 1A$			0.5			0.5			0.5	mA	
		$T_J = 25^{\circ}C$ , $I_O = 1A$			0.8			0.8			0.8	mA	
		$V_{MIN} \leq V_{IN} \leq V_{MAX}$	(7.5 $\leq V_{IN} \leq 20$ )			(14.8 $\leq V_{IN} \leq 27$ )			(17.9 $\leq V_{IN} \leq 30$ )			V	
$V_N$	Output Noise Voltage	$T_A = 25^{\circ}C$ , 10 Hz $\leq f \leq 100$ kHz			40			75			90	$\mu V$	
		$\Delta V_{IN}$ $\Delta V_{OUT}$			68	80		61	72		60	70	dB
$R_O$	Dropout Voltage	$T_J = 25^{\circ}C$ , $I_O = 1A$			2.0			2.0			2.0	V	
		Output Resistance	$f = 1 kHz$		8		18				19	m $\Omega$	
		Short-Circuit Current	$T_J = 25^{\circ}C$			2.1		1.5				1.2	A
		Peak Output Current	$T_J = 25^{\circ}C$			2.4		2.4				2.4	A
		Average TC of $V_O$	Min, $T_J = 0^{\circ}C$ , $I_O = 5 mA$			-0.6		-1.5				-1.8	mV/ $^{\circ}C$
$V_{IN}$	Input Voltage Required to Maintain Line Regulation	$T_J = 25^{\circ}C$			7.5			14.5			17.5	V	

LM140 Electrical Characteristics<sup>(1)</sup>

$-55^{\circ}C \leq T_J \leq +150^{\circ}C$  unless otherwise specified

Symbol	Output Voltage			5V			12V			15V			Units	
	Input voltage (unless otherwise noted)			10V			19V			23V				
	Parameter	Conditions	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max			
$V_O$	Output Voltage	$T_J = 25^{\circ}C$ , $5 mA \leq I_O \leq 1A$	4.8	5	5.2	11.5	12	12.5	14.4	15	15.6	V		
		$P_{D1} \leq 15W$ , $5 mA \leq I_O \leq 1A$	4.75		5.25	11.4		12.6	14.25		15.75	V		
		$V_{MIN} \leq V_{IN} \leq V_{MAX}$	(8 $\leq V_{IN} \leq 20$ )				(15.5 $\leq V_{IN} \leq 27$ )			(18.5 $\leq V_{IN} \leq 30$ )			V	
$\Delta V_O$	Line Regulation	$I_O = 500 mA$			3		50		4		120	mV		
		$\Delta V_{IN}$	(7 $\leq V_{IN} \leq 25$ )			50			(14.5 $\leq V_{IN} \leq 30$ )			(17.5 $\leq V_{IN} \leq 30$ )	V	
		$-55^{\circ}C \leq T_J \leq +150^{\circ}C$									120	150	mV	
		$\Delta V_{IN}$	(8 $\leq V_{IN} \leq 20$ )							(15 $\leq V_{IN} \leq 27$ )		(18.5 $\leq V_{IN} \leq 30$ )	V	
		$I_O = 1A$					50			120		150	mV	
$\Delta V_O$	Load Regulation	$T_J = 25^{\circ}C$			10		50		12		120	mV		
		5 mA $\leq I_O \leq 1.5A$					25				60	75	mV	
		250 mA $\leq I_O \leq 750 mA$											mV	
		$-55^{\circ}C \leq T_J \leq +150^{\circ}C$ , 5 mA $\leq I_O \leq 1A$									50	120	150	mV
		$\Delta V_{IN}$	(8 $\leq V_{IN} \leq 12$ )							(16 $\leq V_{IN} \leq 22$ )		(20 $\leq V_{IN} \leq 26$ )	V	
$I_Q$	Quiescent Current	$I_O = 1A$			6			6			6	mA		
		$T_J = 25^{\circ}C$			7			7			7	mA		
		$-55^{\circ}C \leq T_J \leq +150^{\circ}C$											mA	
$\Delta I_Q$	Quiescent Current Change	5 mA $\leq I_O \leq 1A$			0.5			0.5			0.5	mA		
		$T_J = 25^{\circ}C$ , $I_O = 1A$			0.8			0.8			0.8	mA		
		$V_{MIN} \leq V_{IN} \leq V_{MAX}$	(8 $\leq V_{IN} \leq 20$ )			(15 $\leq V_{IN} \leq 27$ )			(18.5 $\leq V_{IN} \leq 30$ )			V		
$V_N$	Output Noise Voltage	$T_A = 25^{\circ}C$ , 10 Hz $\leq f \leq 100$ kHz			40			75			90	$\mu V$		
		$\Delta V_{IN}$ $\Delta V_{OUT}$			68	80		61	72		60	70	dB	
$R_O$	Dropout Voltage	$T_J = 25^{\circ}C$ , $I_O = 1A$			2.0			2.0			2.0	V		
		Output Resistance	$f = 1 kHz$		8		18				19	m $\Omega$		
		Short-Circuit Current	$T_J = 25^{\circ}C$			2.1		1.5				1.2	A	
$V_{IN}$	Input Voltage Required to Maintain Line Regulation	$T_J = 25^{\circ}C$			7.5			14.5			17.5	V		
		Output Resistance	$f = 1 kHz$			8		18			19	m $\Omega$		
		Short-Circuit Current	$T_J = 25^{\circ}C$			2.1		1.5				1.2	A	
		Peak Output Current	$T_J = 25^{\circ}C$			2.4		2.4				2.4	A	
		Average TC of $V_{OUT}$	$0^{\circ}C \leq T_J \leq +150^{\circ}C$ , $I_O = 5 mA$			-0.6		-1.5				-1.8	mV/ $^{\circ}C$	

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LM140 Electrical Characteristics<sup>(1)</sup> (continued)

-55°C ≤ T<sub>J</sub> ≤ +150°C unless otherwise specified

Symbol	Output Voltage			5V			12V			15V			Units
	Input Voltage (unless otherwise noted)			10V			19V			23V			
	Parameter	Conditions		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
V <sub>IN</sub>	Input Voltage Required to Maintain Line Regulation	T <sub>J</sub> = 25°C, I <sub>O</sub> = 1A		7.5			14.6			17.7			V

LM340-N Electrical Characteristics<sup>(1)</sup>

0°C ≤ T<sub>J</sub> ≤ +125°C unless otherwise specified

Symbol	Output Voltage			5V			12V			15V			Units	
	Input Voltage (unless otherwise noted)			10V			19V			23V				
	Parameter	Conditions		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
V <sub>O</sub>	Output Voltage	T <sub>J</sub> = 25°C, 5 mA ≤ I <sub>O</sub> ≤ 1A		4.8	5	5.2	11.5	12	12.5	14.4	15	15.6	V	
		P <sub>O</sub> ≤ 15W, 5 mA ≤ I <sub>O</sub> ≤ 1A		4.75		5.25	11.4		12.6	14.25		15.75	V	
		V <sub>MIN</sub> ≤ V <sub>IN</sub> ≤ V <sub>MAX</sub>		(7.5 ≤ V <sub>IN</sub> ≤ 20)			(14.5 ≤ V <sub>IN</sub> ≤ 27)			(17.5 ≤ V <sub>IN</sub> ≤ 30)			V	
ΔV <sub>O</sub>	Line Regulation	I <sub>O</sub> = 500 mA	T <sub>J</sub> = 25°C	3	50		4	120		4	150		mV	
			ΔV <sub>IN</sub>	(7 ≤ V <sub>IN</sub> ≤ 25)			(14.5 ≤ V <sub>IN</sub> ≤ 30)			(17.5 ≤ V <sub>IN</sub> ≤ 30)			V	
			0°C ≤ T <sub>J</sub> ≤ +125°C	50			120			150			mV	
			ΔV <sub>IN</sub>	(8 ≤ V <sub>IN</sub> ≤ 20)			(15 ≤ V <sub>IN</sub> ≤ 27)			(18.5 ≤ V <sub>IN</sub> ≤ 30)			V	
ΔV <sub>O</sub>	Load Regulation	T <sub>J</sub> = 25°C	5 mA ≤ I <sub>O</sub> ≤ 1.5A	10	50		12	120		12	150		mV	
				250 mA ≤ I <sub>O</sub> ≤ 750 mA	25			60			75			mV
					5 mA ≤ I <sub>O</sub> ≤ 1A, 0°C ≤ T <sub>J</sub> ≤ +125°C	50			120			150		
				I <sub>O</sub>	Quiescent Current	I <sub>O</sub> = 1A	T <sub>J</sub> = 25°C	8			8			8
0°C ≤ T <sub>J</sub> ≤ +125°C	8.5						8.5			8.5			mA	
5 mA ≤ I <sub>O</sub> ≤ 1A	0.5						0.5			0.5			mA	
T <sub>J</sub> = 25°C, I <sub>O</sub> = 1A	1.0						1.0			1.0			mA	
ΔI <sub>O</sub>	Quiescent Current Change	I <sub>O</sub> = 1A	V <sub>MIN</sub> ≤ V <sub>IN</sub> ≤ V <sub>MAX</sub>	(7.5 ≤ V <sub>IN</sub> ≤ 20)			(14.8 ≤ V <sub>IN</sub> ≤ 27)			(17.9 ≤ V <sub>IN</sub> ≤ 30)			V	
			I <sub>O</sub> = 500 mA, 0°C ≤ T <sub>J</sub> ≤ +125°C	1.0			1.0			1.0			mA	
			V <sub>MIN</sub> ≤ V <sub>IN</sub> ≤ V <sub>MAX</sub>	(7 ≤ V <sub>IN</sub> ≤ 25)			(14.5 ≤ V <sub>IN</sub> ≤ 30)			(17.5 ≤ V <sub>IN</sub> ≤ 30)			V	
			T <sub>J</sub> = 25°C, 10 Hz ≤ f ≤ 100 kHz	40			75			90			μV	
ΔV <sub>IN</sub> ΔV <sub>OUT</sub>	Ripple Rejection	f = 120 Hz	I <sub>O</sub> = 1A, T <sub>J</sub> = 25°C	62	80		55	72		54	70		dB	
			or I <sub>O</sub> = 500 mA, 0°C ≤ T <sub>J</sub> ≤ +125°C	62			55			54			dB	
			V <sub>MIN</sub> ≤ V <sub>IN</sub> ≤ V <sub>MAX</sub>	(8 ≤ V <sub>IN</sub> ≤ 18)			(15 ≤ V <sub>IN</sub> ≤ 25)			(18.5 ≤ V <sub>IN</sub> ≤ 28.5)			V	

LM340-N Electrical Characteristics<sup>(1)</sup> (continued)

0°C ≤ T<sub>J</sub> ≤ +125°C unless otherwise specified

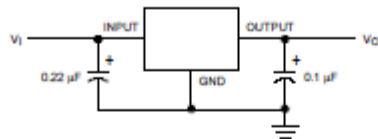
Symbol	Output Voltage			5V			12V			15V			Units
	Input Voltage (unless otherwise noted)			10V			19V			23V			
	Parameter	Conditions		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
R <sub>O</sub>	Dropout Voltage	T <sub>J</sub> = 25°C, I <sub>O</sub> = 1A		2.0			2.0			2.0			V
		f = 1 kHz		8			18			19			mΩ
	Output Resistance	T <sub>J</sub> = 25°C		2.1			1.5			1.2			A
	Short-Circuit Current	T <sub>J</sub> = 25°C		2.4			2.4			2.4			A
	Average TC of V <sub>OUT</sub>	0°C ≤ T <sub>J</sub> ≤ +125°C, I <sub>O</sub> = 5 mA		-0.6			-1.5			-1.8			mV/°C
V <sub>IN</sub>	Input Voltage Required to Maintain Line Regulation	T <sub>J</sub> = 25°C, I <sub>O</sub> = 1A		7.5			14.6			17.7			V

LM7808C Electrical Characteristics

0°C ≤ T<sub>J</sub> ≤ +150°C, V<sub>I</sub> = 14V, I<sub>O</sub> = 500 mA, C<sub>1</sub> = 0.33 μF, C<sub>O</sub> = 0.1 μF, unless otherwise specified

Symbol	Parameter	Conditions <sup>(1)</sup>	LM7808C			Units	
			Min	Typ	Max		
V <sub>O</sub>	Output Voltage	T <sub>J</sub> = 25°C	7.7	8.0	8.3	V	
ΔV <sub>O</sub>	Line Regulation	T <sub>J</sub> = 25°C	10.5V ≤ V <sub>I</sub> ≤ 25V		6.0	160	mV
			11.0V ≤ V <sub>I</sub> ≤ 17V		2.0	80	
ΔV <sub>O</sub>	Load Regulation	T <sub>J</sub> = 25°C	5.0 mA ≤ I <sub>O</sub> ≤ 1.5A		12	160	mV
			250 mA ≤ I <sub>O</sub> ≤ 750 mA		4.0	80	
V <sub>O</sub>	Output Voltage	11.5V ≤ V <sub>I</sub> ≤ 23V, 5.0 mA ≤ I <sub>O</sub> ≤ 1.0A, P ≤ 15W	7.6		8.4	V	
I <sub>O</sub>	Quiescent Current	T <sub>J</sub> = 25°C		4.3	8.0	mA	
ΔI <sub>O</sub>	Quiescent Current Change	With Line	11.5V ≤ V <sub>I</sub> ≤ 25V			1.0	mA
			5.0 mA ≤ I <sub>O</sub> ≤ 1.0A			0.5	
V <sub>N</sub>	Noise	T <sub>A</sub> = 25°C, 10 Hz ≤ f ≤ 100 kHz		52		μV	
ΔV <sub>I</sub> /ΔV <sub>O</sub>	Ripple Rejection	f = 120 Hz, I <sub>O</sub> = 350 mA, T <sub>J</sub> = 25°C	56	72		dB	
V <sub>DO</sub>	Dropout Voltage	I <sub>O</sub> = 1.0A, T <sub>J</sub> = 25°C		2.0		V	
R <sub>O</sub>	Output Resistance	f = 1.0 kHz		16		mΩ	
I <sub>OS</sub>	Output Short Circuit Current	T <sub>J</sub> = 25°C, V <sub>I</sub> = 35V		0.45		A	
I <sub>PK</sub>	Peak Output Current	T <sub>J</sub> = 25°C		2.2		A	
ΔV <sub>O</sub> /ΔT	Average Temperature Coefficient of Output Voltage	I <sub>O</sub> = 5.0 mA		0.8		mV/°C	

Typical Applications



Bypass capacitors are recommended for optimum stability and transient response, and should be located as close as possible to the regulator.

Figure 28. Fixed Output Regulator

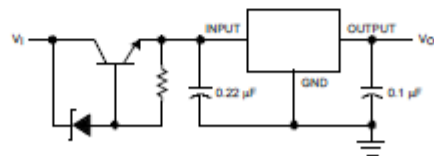
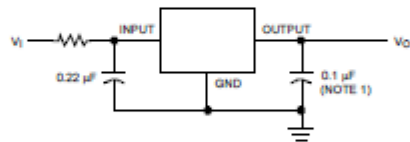
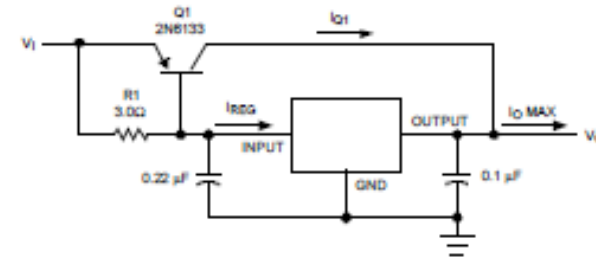


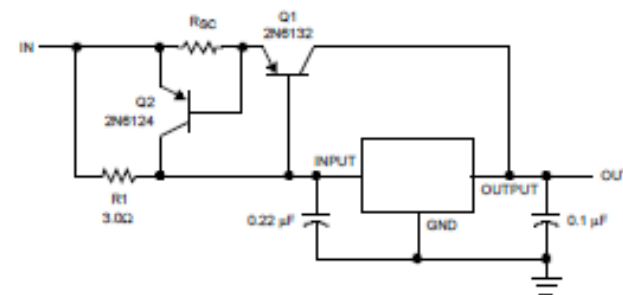
Figure 29. High Input Voltage Circuits



$$\beta(D1) = \frac{I_{O \text{ Max}}}{I_{REG \text{ Max}}}$$

$$R1 = \frac{0.8}{I_{REG}} = \frac{\beta(D1) V_{REG(1)}}{I_{REG \text{ Max}} (\beta + 1) - I_{O \text{ Max}}}$$

Figure 30. High Current Voltage Regulator



$$R_{sc} = \frac{0.8}{I_{sc}}$$

$$R1 = \frac{\beta V_{REG(1)}}{I_{REG \text{ Max}} (\beta + 1) - I_{O \text{ Max}}}$$

Figure 31. High Output Current, Short Circuit Protected



## LM333 3A REGULADOR DE TENSÃO NEGATIVO

## LM333

### 3-Ampere Adjustable Negative Regulator

#### General Description

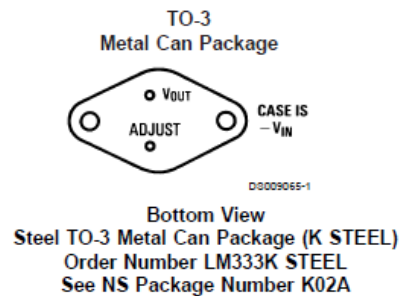
The LM333 is an adjustable 3-terminal negative voltage regulator capable of supplying in excess of  $-3.0A$  over an output voltage range of  $-1.2V$  to  $-32V$ . This regulator is exceptionally easy to apply, requiring only 2 external resistors to set the output voltage and 1 output capacitor for frequency compensation. The circuit design has been optimized for excellent regulation and low thermal transients. Further, the LM333 features internal current limiting, thermal shutdown and safe-area compensation, making them substantially immune to failure from overloads.

The LM333 serves a wide variety of applications including local on-card regulation, programmable-output voltage regulation or precision current regulation. The LM333 is an ideal complement to the LM150/LM350 adjustable positive regulators.

#### Features

- Output voltage adjustable from  $-1.2V$  to  $-32V$
- $3.0A$  output current guaranteed,  $-55^{\circ}C$  to  $+150^{\circ}C$
- Line regulation typically  $0.01\%/V$
- Load regulation typically  $0.2\%$
- Excellent rejection of thermal transients
- $50\text{ ppm}/^{\circ}C$  temperature coefficient
- Temperature-independent current limit
- Internal thermal overload protection
- Standard 3-lead transistor package
- Output is short circuit protected

#### Connection Diagram





**Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Power Dissipation Internally Limited  
Input-Output Voltage Differential 35V

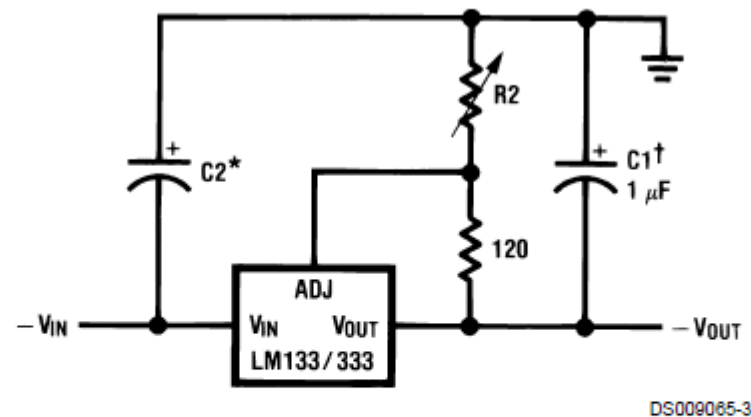
Operating Junction Temperature Range LM333  $T_{MIN}$  to  $T_{MAX}$   
-40°C to +125°C  
Storage Temperature -65°C to +150°C  
Lead Temperature (Soldering, 10 sec.)  
TO-3 Package 300°C  
ESD Susceptibility TBD

**Electrical Characteristics LM333**

Specifications with standard typeface are for  $T_J = 25^\circ\text{C}$ , and those with **boldface type** apply over the full operating temperature range. (Note 3)

Parameter	Conditions	Typical	Min (Note 2)	Max (Note 2)	Units
Reference Voltage	$I_L = 10\text{ mA}$	-1.250	-1.225	-1.275	V
	$3\text{V} \leq  V_{IN} - V_{OUT}  \leq 35\text{V}$ $10\text{ mA} \leq I_L \leq 3\text{A}$ , $P \leq P_{MAX}$	<b>-1.250</b>	<b>-1.213</b>	<b>-1.287</b>	
Line Regulation	$3\text{V} \leq  V_{IN} - V_{OUT}  \leq 35\text{V}$	0.01		0.04	% /V
	$I_{OUT} = 50\text{ mA}$ (Note 4)	<b>0.02</b>		<b>0.07</b>	
Load Regulation	$10\text{ mA} \leq I_L \leq 3\text{A}$ , $P \leq P_{MAX}$ (Notes 4, 5)	0.2		1.0	%
		<b>0.4</b>		<b>1.5</b>	
Thermal Regulation	10 ms Pulse	0.002		0.02	% /W
Temperature Stability	$T_{MIN} \leq T_J \leq T_{MAX}$	<b>0.5</b>			%
Long Term Stability	$T_J = 125^\circ\text{C}$ , 1000 Hours	0.2			%
Adjust Pin Current		65		95	$\mu\text{A}$
		<b>70</b>		<b>100</b>	
Adjust Pin Current Change	$10\text{ mA} \leq I_L \leq 3\text{A}$	2.5		8	$\mu\text{A}$
	$3.0\text{V} \leq  V_{IN} - V_{OUT}  \leq 35\text{V}$				
Minimum Load Current	$ V_{IN} - V_{OUT}  \leq 35\text{V}$	2.5		10	mA
	$ V_{IN} - V_{OUT}  \leq 10\text{V}$	<b>1.5</b>		<b>5.0</b>	
Current Limit (Note 5)	$3\text{V} \leq  V_{IN} - V_{OUT}  \leq 10\text{V}$	3.9	3.0		A
	$ V_{IN} - V_{OUT}  = 20\text{V}$	<b>2.4</b>	1.0		
	$ V_{IN} - V_{OUT}  = 30\text{V}$	<b>0.4</b>	0.20		
Output Noise (% of $V_{OUT}$ )	10 Hz to 10 kHz	0.003			% (rms)
Ripple Rejection	$V_{OUT} = 10\text{V}$ , $f = 120\text{ Hz}$				dB
	$C_{ADJ} = 0\text{ }\mu\text{F}$	60			
	$C_{ADJ} = 10\text{ }\mu\text{F}$	<b>77</b>			
Thermal Resistance Junction to Case	TO-3 Package (K STEEL)	1.2		1.8	$^\circ\text{C/W}$
	TO-220 Package (T)	3		4	
Thermal Shutdown Temperature		163			$^\circ\text{C}$
Thermal Resistance Junction to Ambient (No Heatsink)	K Package	35			$^\circ\text{C/W}$
	T Package	50			

## Adjustable Negative Voltage Regulator



$$-V_{OUT} = -1.25V \left( 1 + \frac{R2}{120\Omega} \right) + \left( -I_{ADJ} \times R2 \right)$$

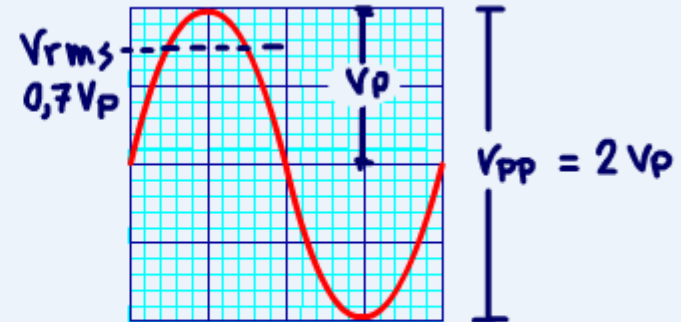
<sup>†</sup>C1 = 1 μF solid tantalum or 10 μF aluminum electrolytic required for stability.

\*C2 = 1 μF solid tantalum is required only if regulator is more than 4" from power supply filter capacitor.

Output capacitors in the range of 1 μF to 1000 μF of aluminum or tantalum electrolytic are commonly used to provide lower output impedance and improved transient response.

AMPLIFICADORES

## EQUAÇÕES DA POTÊNCIA EM AMPLIFICADORES



$$P_{MPD} = \frac{V_{PP}^2}{R_L}$$

$$V_P = \sqrt{2} V_{rms}$$

$$P_{rms} = \frac{V_{rms}^2}{R_L}$$

$$V_{rms} = \frac{V_P}{\sqrt{2}} = \frac{V_{PP}}{2\sqrt{2}}$$

$$P_{rms} = \frac{\left(\frac{V_{PP}}{2\sqrt{2}}\right)^2}{R_L} = \frac{V_{PP}^2}{8 R_L}$$

LM386 LOW VOLTAGE AUDIO POWER AMPLIFIER



LM386

SNAS545C - MAY 2004 - REVISED MAY 2017

**LM386 Low Voltage Audio Power Amplifier**

**1 Features**

- Battery Operation
- Minimum External Parts
- Wide Supply Voltage Range: 4 V–12 V or 5 V–18 V
- Low Quiescent Current Drain: 4 mA
- Voltage Gains from 20 to 200
- Ground-Referenced Input
- Self-Centering Output Quiescent Voltage
- Low Distortion: 0.2% ( $A_V = 20$ ,  $V_S = 6\text{ V}$ ,  $R_L = 8\ \Omega$ ,  $P_O = 125\text{ mW}$ ,  $f = 1\text{ kHz}$ )
- Available in 8-Pin MSOP Package

**2 Applications**

- AM-FM Radio Amplifiers
- Portable Tape Player Amplifiers
- Intercoms
- TV Sound Systems
- Line Drivers
- Ultrasonic Drivers
- Small Servo Drivers
- Power Converters

**3 Description**

The LM386M-1 and LM386MX-1 are power amplifiers designed for use in low voltage consumer applications. The gain is internally set to 20 to keep external part count low, but the addition of an external resistor and capacitor between pins 1 and 8 will increase the gain to any value from 20 to 200.

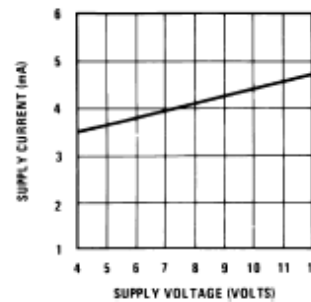
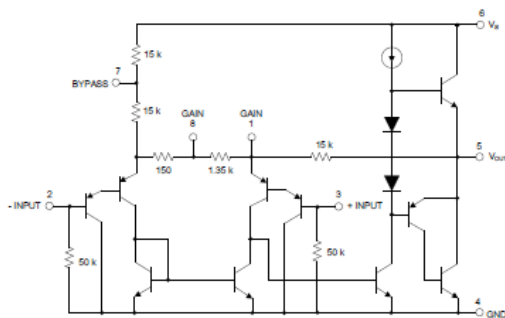
The inputs are ground referenced while the output automatically biases to one-half the supply voltage. The quiescent power drain is only 24 mW when operating from a 6-V supply, making the LM386M-1 and LM386MX-1 ideal for battery operation.

Device Information<sup>(1)</sup>

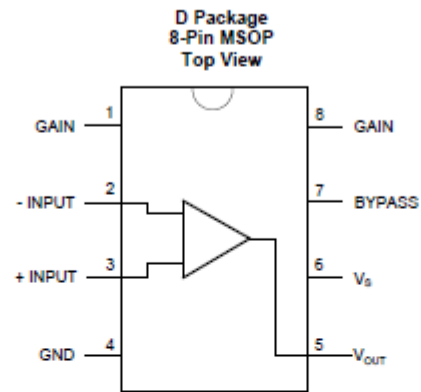
PART NUMBER	PACKAGE	BODY SIZE (NOM)
LM386N-1	PDIP (8)	9.60 mm × 6.35 mm
LM386N-3	PDIP (8)	9.60 mm × 6.35 mm
LM386N-4	PDIP (8)	9.60 mm × 6.35 mm
LM386M-1	SOIC (8)	4.90 mm × 3.90 mm
LM386MX-1	SOIC (8)	4.90 mm × 3.90 mm
LM386MMX-1	VSSOP (8)	3.00 mm × 3.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Schematic



**5 Pin Configuration and Functions**



Pin Functions

PIN		TYPE	DESCRIPTION
NAME	NO.		
GAIN	1	-	Gain setting pin
-INPUT	2	I	Inverting input
+INPUT	3	I	Noninverting input
GND	4	P	Ground reference
V <sub>OUT</sub>	5	O	Output
V <sub>S</sub>	6	P	Power supply voltage
BYPASS	7	O	Bypass decoupling path
GAIN	8	-	Gain setting pin

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
Supply Voltage, $V_{CC}$	LM386N-1/-3, LM386M-1		15	V
	LM386N-4		22	
Package Dissipation	LM386N		1.25	W
	LM386M		0.73	
	LM386MM-1		0.595	
Input Voltage, $V_I$		-0.4	0.4	V
Storage temperature, $T_{STG}$		-65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

		VALUE	UNIT
$V_{(ESD)}$ Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±1000	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



## LM386

SNAS545C –MAY 2004 –REVISED MAY 2017

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## 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V <sub>CC</sub>	Supply Voltage	4		12	V
	LM386N-4	5		18	V
	Speaker Impedance	4			Ω
V <sub>I</sub>	Analog input voltage	-0.4		0.4	V
T <sub>A</sub>	Operating free-air temperature	0		70	°C

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>	LM386	LM386	LM386	UNIT	
	D (SOIC)	DGK (VSSOP)	P (PDIP)		
	8	8	8		
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	115.7	109.3	53.4	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	59.7	73.1	42.1	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	56.2	100.2	30.6	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	12.4	9.2	19.0	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	55.6	99.1	50.5	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.5 Electrical Characteristics

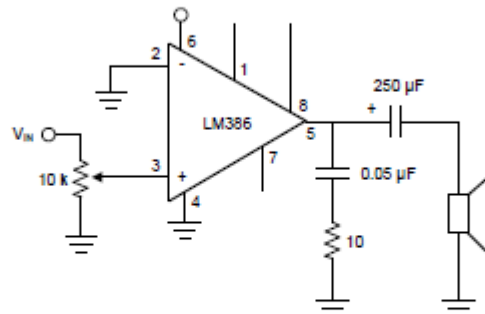
over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_S$ Operating Supply Voltage	LM386N-1, -3, LM386M-1, LM386MM-1	4		12	V
	LM386N-4	5		18	
$I_Q$ Quiescent Current	$V_S = 6\text{ V}$ , $V_{IN} = 0$		4	8	mA
$P_{OUT}$ Output Power	$V_S = 6\text{ V}$ , $R_L = 8\ \Omega$ , THD = 10% (LM386N-1, LM386M-1, LM386MM-1)	250	325		mW
	$V_S = 9\text{ V}$ , $R_L = 8\ \Omega$ , THD = 10% (LM386N-3)	500	700		
	$V_S = 16\text{ V}$ , $R_L = 32\ \Omega$ , THD = 10% (LM386N-4)	700	100		
$A_V$ Voltage Gain	$V_S = 6\text{ V}$ , $f = 1\text{ kHz}$		26		dB
	10 $\mu\text{F}$ from Pin 1 to 8		46		
BW Bandwidth	$V_S = 6\text{ V}$ , Pins 1 and 8 Open		300		kHz
THD Total Harmonic Distortion	$V_S = 6\text{ V}$ , $R_L = 8\ \Omega$ , $P_{OUT} = 125\text{ mW}$ $f = 1\text{ kHz}$ , Pins 1 and 8 Open		0.2%		
PSRR Power Supply Rejection Ratio	$V_S = 6\text{ V}$ , $f = 1\text{ kHz}$ , CBYPASS = 10 $\mu\text{F}$ Pins 1 and 8 Open, Referred to Output		50		dB
$R_{IN}$ Input Resistance			50		k $\Omega$
$I_{BIAS}$ Input Bias Current	$V_S = 6\text{ V}$ , Pins 2 and 3 Open		250		nA

9.2 Typical Application

9.2.1 LM386 with Gain = 20

Figure 10 shows the minimum part count application that can be implemented using LM386. Its gain is internally set to 20.



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Figure 10. LM386 with Gain = 20

9.2.1.1 Design Requirements

Table 1. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.1.3 Application Curve

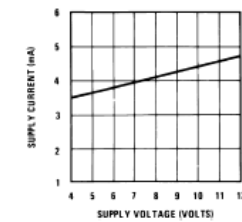
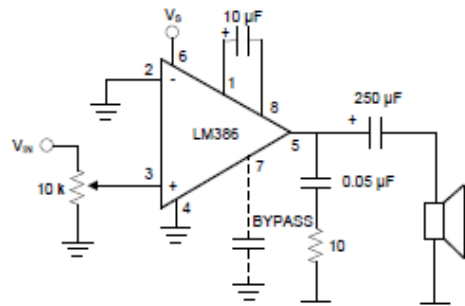


Figure 11. Supply Current vs Supply Voltage

9.2.2 LM386 with Gain = 200



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Figure 12. LM386 with Gain = 200

9.2.2.1 Design Requirements

Table 2. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.2.2 Detailed Design Procedure

The Detailed Design Procedure can be found in the [Detailed Design Procedure](#) section.

9.2.2.3 Application Curve

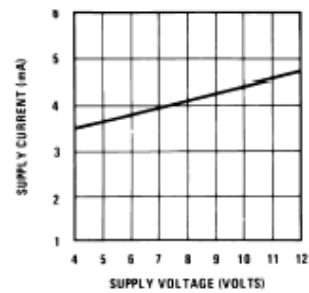
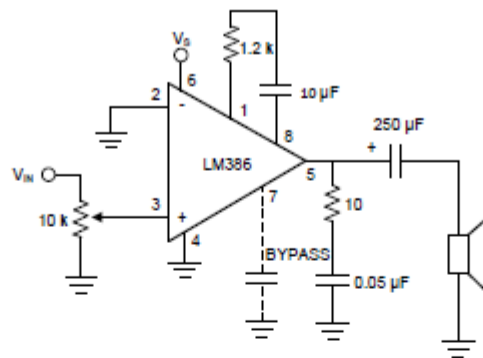


Figure 13. Supply Current vs Supply Voltage

9.2.3 LM386 with Gain = 50



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Figure 14. LM386 with Gain = 50

9.2.3.1 Design Requirements

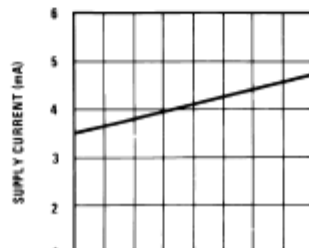
Table 3. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.3.2 Detailed Design Procedure

The Detailed Design Procedure can be found in the [Detailed Design Procedure](#) section.

9.2.3.3 Application Curve



9.2.4 Low Distortion Power Wienbridge Oscillator

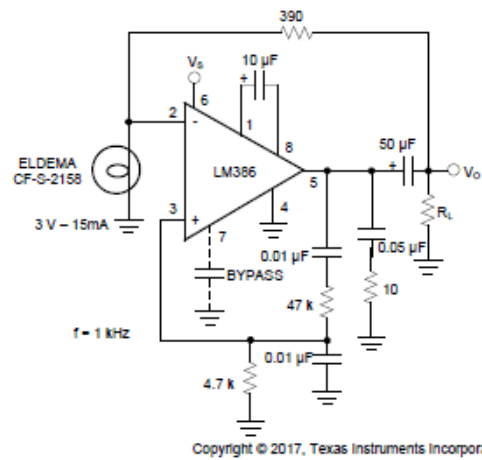


Figure 16. Low Distortion Power Wienbridge Oscillator

9.2.4.1 Design Requirements

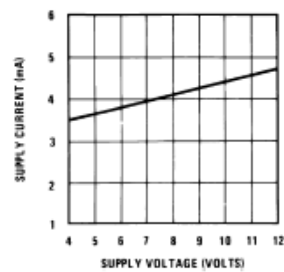
Table 4. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.4.2 Detailed Design Procedure

The Detailed Design Procedure can be found in the [Detailed Design Procedure](#) section.

9.2.4.3 Application Curve



9.2.5 LM386 with Bass Boost

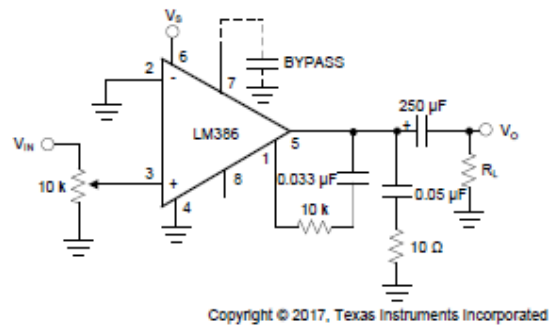


Figure 18. LM386 with Bass Boost

9.2.5.1 Design Requirements

Table 5. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.5.2 Detailed Design Procedure

The Detailed Design Procedure can be found in the [Detailed Design Procedure](#) section.

9.2.5.3 Application Curve

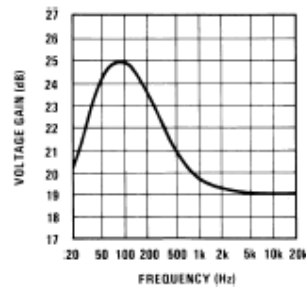


Figure 19. Voltage Gain vs Frequency

9.2.6 Square Wave Oscillator

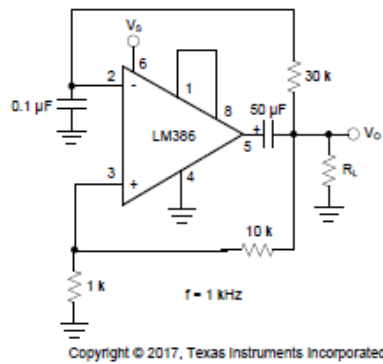


Figure 20. Square Wave Oscillator

Table 6. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.6.1 Detailed Design Procedure

The Detailed Design Procedure can be found in the [Detailed Design Procedure](#) section.

9.2.6.2 Application Curve

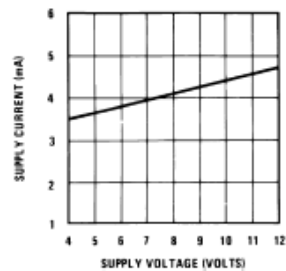
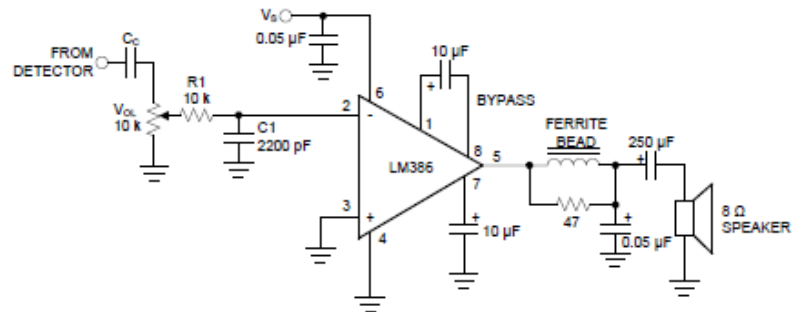


Figure 21. Supply Current vs Supply Voltage



9.2.7 AM Radio Power Amplifier



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Figure 22. AM Radio Power Amplifier

9.2.7.1 Design Requirements

Table 7. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Load Impedance	4 Ω to 32 Ω
Supply Voltage	5 V to 12 V

9.2.7.2 Detailed Design Procedure

The Detailed Design Procedure can be found in the [Detailed Design Procedure](#) section.

9.2.7.3 Application Curve

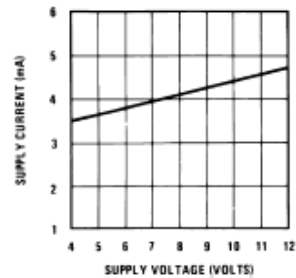


Figure 23. Supply Current vs Supply Voltage

11 Layout

11.1 Layout Guidelines

Place all required components as close as possible to the device. Use short traces for the output to the speaker connection. Route the analog traces far from the digital signal traces and avoid crossing them.

11.2 Layout Examples

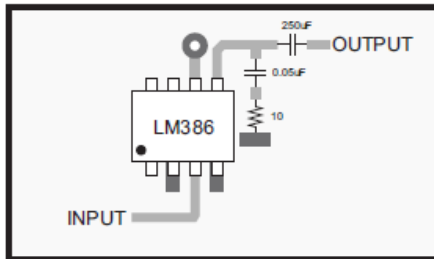


Figure 24. Layout Example for Minimum Parts Gain = 20 dB on PDIP package

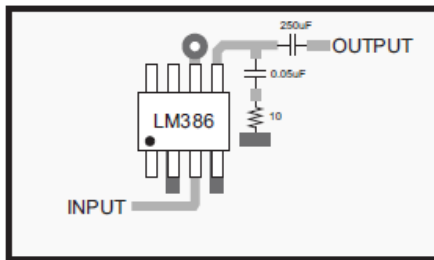


Figure 25. Layout Example for Minimum Parts Gain = 20 dB on SOIC package

LA4282 2 CANAL 10W STEREO COM MUTE PARA HOME TV

Monolithic Linear IC



LA4282

2-Channel 10 W AF Power Amplifier for Use in Home Stereo, TV Applications

Overview

The LA4282 is an IC which seals a high-output power amplifier for TVs and monitors in a compact package.

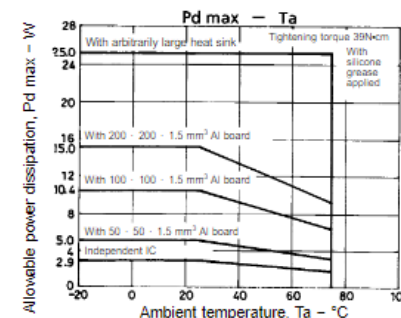
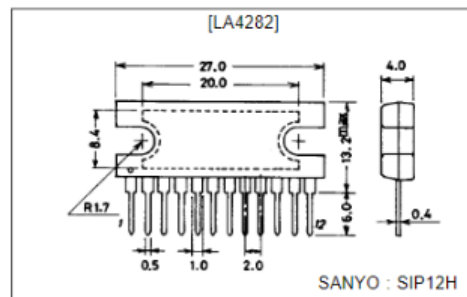
Features

- High-power 2-channel AF power amplifier
- Low distortion
- Minimum number of external parts required (no bootstrap capacitor required)
- Low pop noise at the time of power supply ON/OFF
- Good ripple rejection (58 dB typ)
- Wide operating voltage range
- External muting available
- On-chip protector against abnormality (thermal shutdown, overvoltage)

Package Dimensions

unit : mm

3049A-SIP12H

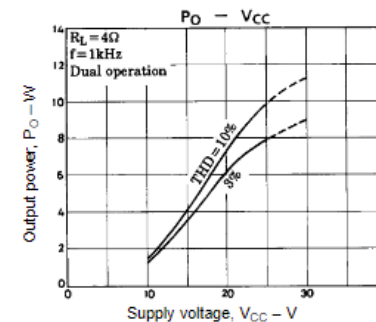
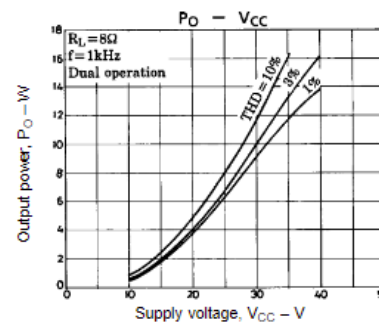


Specifications

Maximum Ratings at Ta = 25 °C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V <sub>CC max</sub>	Quiescent	45	V
Maximum output current	I <sub>O peak</sub>		4	A
Allowable power dissipation	Pd max	With heat sink	25	W
Operating temperature	T <sub>opr</sub>		-20 to +75	°C
Storage temperature	T <sub>stg</sub>		-40 to +150	°C

Operating Conditions at Ta = 25°C



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22896HA(II)/7118TA,TS No.2683-1/6

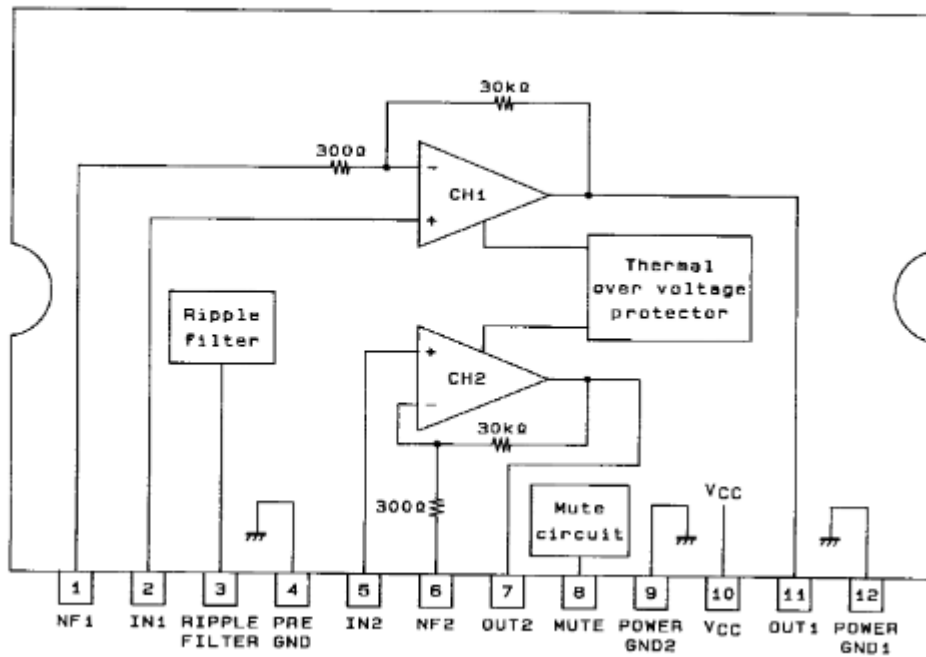
**LA4282**

**Operating Characteristics at  $T_a = 25^\circ\text{C}$ ,  $V_{CC} = 32\text{ V}$ ,  $R_L = 8\ \Omega$ ,  $f = 1\text{ kHz}$ ,  $R_g = 600\ \Omega$ ,  
See Test Circuit.**

Parameter	Symbol	Conditions	min	typ	max	Unit
Quiescent current	$I_{CCO(1)}$	Quiescent	30	60	100	mA
	$I_{CCO(2)}$	Muting switch On	30	56	100	mA
Voltage gain	VG		38	40	42	dB
Voltage gain difference	$\Delta\text{VG}$				1.5	dB
Output power	$P_O(1)$	THD = 1%	9.0	10.0		W
	$P_O(2)$	THD = 3%	10.0	11.5		W
Total harmonic distortion	THD	$P_O = 2\text{ W}$		0.05	0.20	%
Output noise voltage	$V_{NO}$	$R_g = 10\text{ k}\Omega$ , BW = 20 Hz to 20 kHz		0.25	1.0	mV
Ripple rejection	SVRR	$R_g = 10\text{ k}\Omega$ , $f_R = 100\text{ Hz}$ , $V_R = 0\text{ dBm}$	45	58		dB
Crosstalk	CT	$R_g = 10\text{ k}\Omega$	45	60		dB
Muting	$V_{O(MT)}$	Muting switch On, $V_{IN} = -5\text{ dBm}$			-35	dBm

**Equivalent Circuit Block Diagram and Pin Assignment**

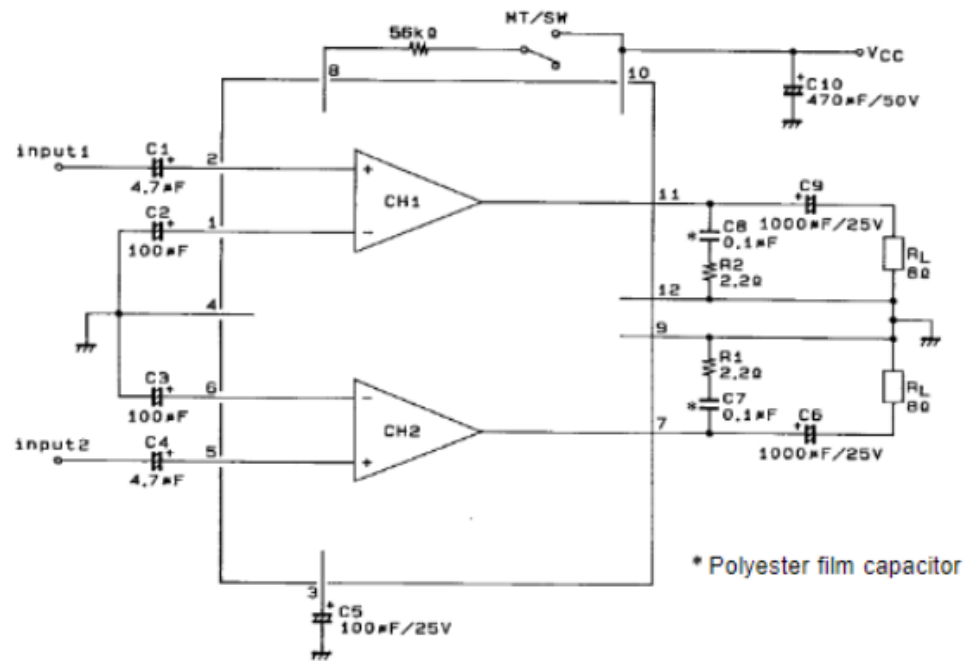
Equivalent Circuit Block Diagram and Pin Assignment



A05005

LA4282

Test Circuit




A05006

FÓRMULAS E GRANDEZAS ELÉTRICAS.

RELAÇÃO E FÓRMULAS DAS PRINCIPAIS GRANDEZAS ELÉTRICAS.

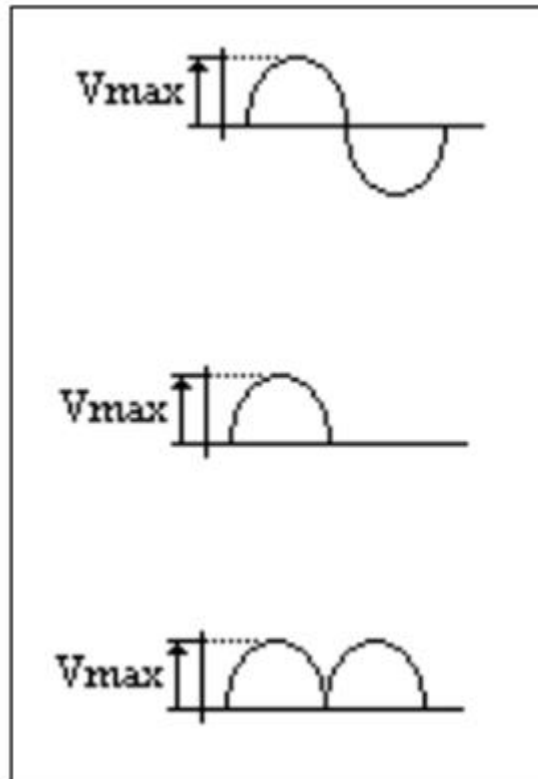
10. RESUMO.



DC	AC
<p><b>Resistência</b></p> $R = \frac{V}{I} [\Omega]$	<p><b>Impedância</b></p> $Z = \frac{I_{AC}}{V_{AC}} [\Omega]$ $Z = R + jX$
	<p><b>Reatância</b></p> $X_L = 2\pi f [\Omega]$ $X_C = \frac{-1}{2\pi f C} [\Omega]$
<p><b>Condutância</b></p> $G = \frac{I}{V} [S]$	<p><b>Admitância</b></p> $Y = \frac{V_{AC}}{I_{AC}} [S]$ $Y = G + jB$
	<p><b>Susceptância</b></p> $B_L = \frac{-1}{2\pi f L} [S]$ $B_C = 2\pi f C [S]$



## FÓRMULA DA TENSÃO RMS E TENSÃO DC EM RETIFICADORES.



Corrente Alternada :

$$V_{rms} = \frac{V_{max}}{\sqrt{2}} \quad V_{med} = 0$$

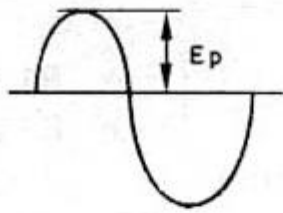
Meia Onda :

$$V_{rms} = \frac{V_{max}}{2} \quad V_{med} = \frac{V_{max}}{\pi}$$

Onda Completa :

$$V_{rms} = \frac{V_{max}}{\sqrt{2}} \quad V_{med} = \frac{2V_{max}}{\pi}$$

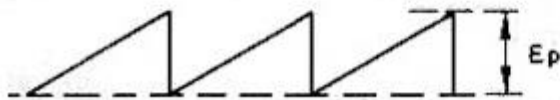
Tensão RMS e média em outras formas de ondas.



$$E_{rms} = \frac{E_p}{\sqrt{2}} \quad E_m = \frac{2E_p}{\pi}$$



$$E_{rms} = E_p \sqrt{\frac{1}{2}} \quad E_m = \frac{E_p}{2}$$

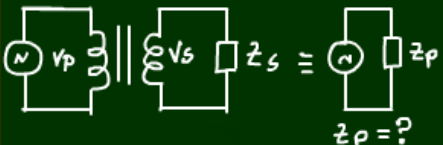


$$E_{rms} = \frac{E_p}{\sqrt{3}} \quad E_m = \frac{E_p}{2}$$

Cálculos para valores médios e RMS

## FÓRMULA IMPEDÂNCIA REFLETIDA NO TRANSFORMADOR.

Deduzindo a Impedância Refletida



$$P_s = P_p$$

$$\frac{V_s^2}{Z_s} = \frac{V_p^2}{Z_p}$$

$$Z_p \cdot V_s^2 = Z_s \cdot V_p^2$$

$$Z_p = \frac{V_p^2}{V_s^2} \cdot Z_s$$

$$Z_p = \left( \frac{V_p}{V_s} \right)^2 \cdot Z_s = \left( \frac{N_p}{N_s} \right)^2 \cdot Z_s$$

Forma rápida de deduzir a impedância refletida

Todo mundo sabe que a impedância do secundário aparece refletida no primário, mas a equação a gente esquece toda hora, vou mostrar uma forma simples de deduzir, para você nunca mais esquecer.

O truque é usar a potência do transformador escrevendo a potência em função da resistência e da tensão, aqui em função da impedância e da tensão.

A impedância do secundário refletida no primário é igual ao quadrado da relação de espiras, número de espira do primário sobre o número de espiras do secundário ao quadrado multiplicado pela impedância do secundário.

YOUTUBE: <https://youtu.be/W3eDnH1vQcE>

TABELA: CORRENTE MÁXIMAS NOS CONDUTORES SÉRIE MM E AWG.

CORRENTES EM CONDUTORES			
PVC/60°C - EB-98 ABNT		PVC/70°C - NBR-6148 - ABNT	
AWG/MCM	AMPÈRES	SÉRIE MÉTRICA	AMPÈRES
22	3,5	0,3	3,5
20	6	0,5	6
18	10	0,75	9
16	13	1	12
14	15	1,5	15,5
12	20	2,5	21
10	30	4	28
8	40	6	36
6	55	10	50
4	70	16	68
2	95	25	89
1	110	35	111
1/0	125	30	134
2/0	145	X	X
3/0	165	70	171
4/0	195	X	X
250	215	95	207
300	240	120	239
350	260	150	272
400	280	X	X
500	320	185	310
600	355	240	364
700	385	X	X
750	400	X	X
800	410	300	419
900	435	X	X
1000	455	X	X
X	X	400	502
X	X	500	578

As correntes indicas aqui são para instalações em eletrônica.

## TABELA COM A RESISTIVIDADE E CONDUTIVIDADE DOS CONDUTORES.

Resistividade e condutividade condutores 20°C		
Substância	condutividade	resistividade
	m Ohm x mm <sup>2</sup>	Ohm x mm <sup>2</sup> m
PRATA	62,50	0,0160
COBRE(PURO)	61,70	0,0162
COBRE(DURO)	56,10	0,0178
OURO	43,50	0,0230
ALUMÍNIO	34,20	0,0292
TUNGSTÊNIO	18,18	0,0550
ZINCO	17,80	0,0562
BRONZE	14,90	0,0671
LATÃO	14,90	0,0671
NÍQUEL	10,41	0,0961
FERRO(PURO)	10,20	0,0980
PLATINA	9,09	0,1100
ESTANHO	8,60	0,1163
CHUMBO	4,80	0,2083
MANGANINA	2,08	0,4808
CONSTANTAN	2,00	0,5000
MECÚRIO	1,00	0,9956
NICROMO	0,91	1,1001
GRAFITE	0,07	14,2857

Quanto maior a condutividade melhor como condutor, quanto maior a resistividade melhor como isolante, veja a lista dos melhores condutores comparando a condutividade em m/OHM x mm<sup>2</sup>.

Prata	62,50
Cobre	58,82
OURO	45,05
Alumínio	35,97
Platina	9,09
Estanho	8,33
Chumbo	4,81

FIM



**CRÉDITOS**

E por favor, se você não é inscrito, se inscreva e marque o sininho para receber as notificações do canal e não esqueça de deixar aquele like e compartilhar para dar uma força ao canal do professor bairros.

**Arthurzinho: E não tem site.**

Tem sim é [www.bairrospd.com](http://www.bairrospd.com) lá você encontra o pdf e tutoriais sobre esse e outros assuntos da eletrônica

E fique atento ao canal do professor bairros para mais tutoriais sobre eletrônica, até lá!



The image shows a screenshot of the website [www.bairrospd.com](http://www.bairrospd.com). The website header includes the logo 'bairrospd' and the text 'BAIROS PROJETOS DIDÁTICOS E ELETRÔNICOS'. A green banner at the top of the page reads 'ESTUDE ELETRÔNICA NO SITE WWW.BAIROSPD.COM'. Below this, there is a section titled 'Um site para pesquisar eletrônica' with three columns of text. A navigation menu includes 'HOME', 'CURSOS', 'BIBLIOTECA', 'TUTORIAIS', 'VOCÊ SABIA?', and 'CONTATO'. A featured article is titled 'APRENDA A LER RESISTORES' and includes an illustration of a man working with a resistor. To the right of the article is a search bar and a snippet of text: 'O QUE SIGNIFICA GASTAR ENERGIA ELÉTRICA: Uma questão de Potência.' At the bottom of the page, there is a blue button that says 'AULAS OU ASSESSORIA COM O ENGENHEIRO E PROFESSOR ROBERTO BAIROS?' and a 'CLIQUE AQUI!!' link.

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**[www.bairrospd.com](http://www.bairrospd.com)  
Professor Bairros**

[www.bairrospd.com](http://www.bairrospd.com)

[https://www.youtube.com/channel/UC\\_tfxnYdBh4lbiR9twtpPA](https://www.youtube.com/channel/UC_tfxnYdBh4lbiR9twtpPA)

Sim, mais um e-book do Professor Bairros especialmente desenvolvido para os amantes da eletrônica. Aqui você encontra uma lista dos componentes mais usados na montagem de circuitos eletrônicos. Os componentes são organizados mostrando os modelos mais comuns e suas principais características o que vai ajudar o técnico a escolher o melhor modelo, ou melhor substituto.

A eletrônica é divertida, divirta-se!

VISITE O SITE DO PROFESSOR BAIROS LÁ TEM O PDF E MUITO MAIS

PARA AULAS ONLINE CONTATE VIA SITE

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SOM: pop alegre Mysteries -30 (fonte YOUTUBE)

SEO: